## 830B-1B 250W FM TRANSMITTER

## SPARE SEMICONDUCTOR AND TUBE LIST

9-1-69

## Quantity

Recommended Set

1
1
2
3
1
1
2
1
1

1

1

1

2

1
1
1
2
2
2
2
4
2
2
1

1

CPN
352-0583-010
352-0695-010
352-0322-000
352-0743-010
352-0713-030
.352-0116-000
352-0349-000
352-0773-030
352-0638-010
352-0373-000
352-0848-020
352-0756-010
352-0630-010
352-0671-010
352-0611-010
352-0747-010
352-0629-030
352-0695-040
353-1721-000
353-3593-010
353-2018-000
353-2857-000
353-3271-000
353-2734-000

353-2710-000

## Description

2N3055
2N3740
2N708
2N4121
2N3643
2N491
2N1613
2N4250
2N3565
S4639
2N4258
2N4416
2N3563
2N3866
2N3375
2N5102
2N3569
2N4235
1N1200
FA2311U
1N270
1N626
FA4000
1N718

| Quantity Complete Set | Quantity <br> Recommended Set |  |  | CPN | Description |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 1 |  | 1 |  | 353-3304-000 | SV3173 |
| $i$ |  | 1 |  | 353-3123-000 | 1N3018B |
| 1 |  | 1 |  | 353-3057-000 | 1N3027B |
| 5 |  | 2 |  | 353-2607-000 | 1N645 |
| 3 |  | 2 |  | 353-2724-000 | 1N758A |
| 2 |  | 1 |  | 353-2712-000 | 1N752A |
| 2 |  | 1 |  | 353-3125-000 | IN3020B. |
| 1 |  | 1 |  | 353-2718-000 | 1N755A |

B830-1 AMPLIFIER

## 2

1
353-1794-000
4JA421EX55

Transistors and diodes, $100 \%$ set including H.V. rectifiers $\$ 273.30$ \$169.20
$\vdots \quad \cdot \quad$ TUBES
$\$ 38.50$
$\$ 37.00$

2

1

Tube set, $100 \%$

Tube set, recommended
1

1

257-0001-000
256-0138-000

OD3 Regulator
4CX250B

830D-1B 1KN FM TRANSMITTER

## SPARE SEMICONDUCTOR AND TUBE LIST

310Z-1 EXCITER
9-1-69
Quantity Complete Set

2

1
4

7

2
1
4
1
1
1
1
2
4

1

1

1

3
5

4

4
10
3

3
-
1

Quantity
Recommended Set
1

1

2

3

1
1
2

1

1

1

1

1

2

1
1
1
2
2
2
2
4

2

2
1

1

## CPN

352-0583-010
352-0695-010
352-0322-000
352-074,3-010
352-0713-030
$\therefore 352-0115-000$
352-0349-000
2N1613
352-0773-030
2N4250
352-0638-010
352-0373-000
352-0848-020
352-0756-010
352-0630-010
352-0671-010
352-0611-010
352-0747-010
352-0629-030
352-0695-040
-353-1721-000
353-3593-010
353-2018-000
353-2857-000
353-3271-000
353-2734-000
353-2710-000
1N751A
Quantity
Complete Set

1
1

1

5
3
2
2
1

D830-1 AMPLIFIER

## + 2 <br> 2

Quantity
Recommended Set
1
1
1

2

2
1
1
1

8
2

2
1
1
1
1
Complete set of transistors and diodes
=.-_ Recommended set of transistors and diodes
1

1 .

1

1

1
1
$=$ Complete set of transistors and diodes
$==$ Recommended set of transistors and diodes

## TUBES

## CPN

| 353-3304-000 | SV3173 |
| :---: | :---: |
| 353-3123-000 | 1N3018B |
| 353-3057-000 | 1N3027B |
| 353-2607-000 | 1N645 |
| 353-2724-000 | 1N758A |
| 353-2712-000 | 1N752A |
| 353-3125-000 | 1 N 3020 B |
| 353-2718-000 | 1N755A |

## Description

SV3173

1N755A

1


Tube set, $100 \%$
Tube set, recommended

353-6015-000
549-2463-004
353-1661-000
353-1736-000
353-3121-000
353-3220-000
353-2016-000
$\$ 512.25$
$\$ 209.00$

50M140ZB5
H.V. Rectifier Stack 1N1492

1N1566
1N3016B
1N963A
1N270

| 1 |  |  |
| :---: | :---: | :---: |
| Tube set, $100 \%$ | $256-0123-000$ | 4CX1000A |
| Tube set, recommended | $\$ 181.00$ |  |
| $\$ 181.00$ |  |  |

## SPARE SEMTCONDUCTOR AND TUBE LIST

Quantity
Complete Set

2

1
4
7
2
$1 \quad \ddots \quad 1$

4

1

1

1
1
2
4
1
1

1

3

5

4

4
10
3

3

1

CPN
352-0583-010
352-0695-010
352-0322-000
352-0743-010
352-0713-030
:352-0116-000
352-0349-000
352-0773-030
352-0638-010
352-0373-000
352-0848-020
352-0756-010
352-0630-010
352-0671-010
352-0611-010
352-0747-010
352-0629-030
352-0695-040
353-1721-000
353-3593-010
353-2018-000
353-2857-000
353-3271-000
353-2734-000
353-2710-000

## Description

2N3055
2N3740
2N708
2N4121
2N3643
2N491
2N1613
2N4250
2N3565
S4639
2N4258
2N4416
2N3563
2N3866
2N3375
2N5102
2N3569
2N4235
1N1200
FA2311U
1N270
1N626
FA4000
1N718

1N751A

Quantity Complete Set

1
1
1
5
3
2
2
1

3830 AMPLIFIER
1
353-1794-000
4JA421EX55

## E830 AMPLIFIER <br> - 2



Transistors and diodes, $100 \%$ set
-.-Transistors and diodes, recommended set
353-1546-000
$\$ 422.00$
\$267.00

TUBE SET

256-0138-000

## Description

SV3173
1N3018B
1N3027B
1N645
1N758A
1N752A
IN3020B
1N755A

1
1
OD3 Regulator
4CX250B driver


## SPARE SEMICONDUCTOR AND TUBE LIST

310Z-1 EXCITER 9-1-69

Quantity Complete Set

2
1
4
7
2
1
4
4
1
1

## 1

1
2
. Quantity
Recommended Set
1.

1

2
3
1

1

2
1
1
1
1
1
2
1
1
1
2
2
4
.4
10
3

3

1

1



3830 AMPLIFIER

## F830 AMPLIEIER

| 12 |  | 4 | $353-1546-000$ |  |
| :--- | :--- | :--- | :--- | :--- |
| 1 | $\vdots$ | 0 |  | $353-6273-000$ |
| 0 |  | 2 | $353-6259-000$ |  |
| 0 | $\vdots$ | 2 |  | $353-6260-000$ |
| 1 | 0 | 3 | $549-2259-004$ |  |
| 0 |  | 0 | $353-1546-000$ |  |
| 1 |  | 1 | $353-6257-000$ |  |
| 0 |  |  | $353-6258-000$ |  |

1N540
Z404 Screen rectifie Rectifier (part of Z404)
Rectifier (part of Z404)
Z405 Bias Rectifier
1N540 (part of 2405)
2406 3ø H.V.Rectific
One leg of 2406
$=$ Transistors and diodes, $100 \%$ set including $3 \emptyset$ H.V. rectifier stack


Wew Capacitor plate 548-7995-003
2 standoff-190.0015-000
2 screws-347-0171-000
2 serew's-34.3-0.330-000
4 washers-302-0026-600
4 washers-310.-0447-000
Figure 4-1. D830-1 FM Power Amplifier (Sheet 1 of 10 )


## SET OF SPARE TRANSISTORS AND DIODES FOR 786M-1




Ref: $\dot{f}$ igsti 8 s30-1 250 Watt FM Power Amplifier, Schematic Diagram Parts Required

| $R-235$ | $C P N$ | $749-4512-000$ | $100 \Omega 10 \%$ |  |
| :--- | :--- | :--- | :--- | :--- |
| $T P-301$ | $C P N$ | $360-0156-000$ | TeST PoINT | Black |
| TP-302 | $C P N$ | $360-0155-000$ | Test Posit yellow |  |

$3 n$


The Eimac 8170/4CX5000A is a compact high-power ceramic an cooled by forced air. It is useful as an oscillator, amplifier, or modulaton up to 110 megacycles and is particularly suited for use as a linca! amplified, class- $\mathrm{AB}_{1}$ audio amplifier, or as a screen-modulated radio-fue

A pair of these tubes will deliver 17.5 kilowatts of audio-frey frequency power with zero driving power. The rated plate dissipation is ti for most classes of services and six kilowatts for class-AB operation

## GENERAE GUADACTERISTICS

## ELECTRICAL

Filament: Thoriated Tungsten


Direct Interelectrede Capaciłances, Grounded Cothode:
al tetrode quencies 'rand
$\qquad$

Faedback
Direct Intarelectrade Capacitanees, Grounded Grid and Screen:
Input
Output
Max.
58 uuf
23 uuf
Faedback
?

uuf


| typical operation | (Fraquancier |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| D.C Scraen Vallaq |  |  |  | 350 | vollt |
| D.C Grid Voltage |  |  |  | -350 | volts |
| D.C Plate Curront |  |  | - | 2.3 | mpar |
| crean Currant |  |  | - | 0.2 | ore |
| D.C Grid Currant |  |  |  | 0.05 | ampor |
| Driving Power |  |  |  | 25 |  |
| Usaful Output Po |  |  |  |  |  |

## PLATE-MODULATED RADIO-FREQUENCY POWER AMPLIFIER

Class-C Telephony (Cartier conditions excopt where noted) maximum ratings


| O.C Plate Voltag | - | - | - | - |  | - |  |  |  |  | - | 5000 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| D.C Sersen Voltag | - |  | - | - |  | - |  | - | - |  | $\cdot$ | 500 |

## SCREEN-MODULATED RADIO-FREQUENCY POWER AMPLIFIER

[Carrier conditions axcept where noted]
$\left.\begin{array}{l}\text { MAXIMUM RATINGS, Class-C Telephony (Per Tube) } \\ \text { D.C PLATE VOLTAGE } \\ \text { D-C SCREEN VOLTAGE }\end{array}\right)$


NOTE: Tvo tubes can be amployed undar conditions listed in the fist column to obtain more than fiva kilowatts plate output power. Likewise, three tubes can be utilized at conditions listed in the second column to obtain better than ten filowatts output pawar.

## AUDIO-FREQUENCY AMPLIFIER OR MODULATOR

 Class-ABiMAXIMUM RATINGS [Per Tube]



## RADIO-FREQUENCY LINEAR AMPLIFIER

Class-AB.
MAXIMUM RATINGS
d-C PLate voltage
d-C screen voltage.
7500 MAX. VOLTS
d-C plate current
1500 MAX. VOLTS

PLATE DISSIP/ TION
4.0 MAX. AMPERES

SCREEN DISSIPATION
6000 MAX. WATTS

GRID DISSIPATION
250 MAX. WATTS
TYPICAL OPERATION. Peak-Envelope or Modulation-Crest Conditions.
(Frequancias below 30 meqacycles)
D.C Piafa Yoltage - - - - - - . - 7500 valis
D.C Screan Yaltaga . . . . . . . . 1250 volts

D-C Grid Yoltage"
-300 volts
Max-Signal Plata Currant
Zara-Siqnal Plate Current
Max.-Signal Screan Currant
1.9 ampera
0.50 dmpere

Peat R-F Grid Voliage
0.20 ampara

Driving Pawer - .
300 volts
. . . . - - - a walis
Plata Dissipation . . . . . . . . 4200 watis
Plate Output Poware . - - . . - . 10,000 watts
-Adiust grid voltage to obtain ipecifiod Zero-Signal plate current. **PEP output or r-f output power at crast of madulation anvalape.

NOTE: In most cases, "TYPICAL OPERATION" data are obtained by calculation from published characteristic curves and confirmed by direct tests. No allowanca for circuit losses, either input or output, ha: boon made. Exceptions are distinguishad by a listing of "Useful" output powar as opposed to "Plate" output power. Values appearing in these groups have been obtained from existing equipmont(s) and the output power is that measured at the load.

# APPLICATION 

## MECHANICAL

Mounting-The 4 CX 5000 A must be operated with its axis vertical. The base of the tube may be down or up at the convenience of the circuit designer.
Socket-The Eimac SK-300A Air-System Socket is designed especially for the concentric base terminals of the $4 \mathrm{CX5000}$. The use of recommended air-flow rates through this socket provides effective forced-air cooling of the tube. Air forced into the bottom of the socket passes over the tube terminals and through an Air Chimney, the SK-306, into the anode cooling fins. The SK-300 socket may be used instead of the SK-300A, but its use will result in a slightly less efficient cooling system at high dissipation levels.
Cooling-The maximum temperature rating for the external surfaces of the 4 CX 5000 A is $250^{\circ} \mathrm{C}$. Sufficient forced-air circulation must be provided to keep the temperature of the anode at the base of the cooling fins and the temperature of the ceramic-metal seals below $250^{\circ} \mathrm{C}$. Sea level air-flow requirements to maintain seal temperatures at $200^{\circ} \mathrm{C}$ in $50^{\circ} \mathrm{C}$ ambient air are tabulated below (for operation below 30 megacycles).

|  | Sx.300A Sackal |  | SK. 100 Sockal |  |
| :---: | :---: | :---: | :---: | :---: |
| $\begin{aligned} & \text { Mate } \\ & \text { Disiioation= } \\ & \text { (Waths] } \end{aligned}$ | Air flow (CFM) | Prossure Drop (Inche: of water) | Aír Flow (CFM) | Pressure <br> Drop (Inchos of woter) |
| 2000 | 75 | 0.4 | 75 | 0.4 |
| 3000 | 105 | 0.7 | 100 | 0.7 |
| 4000 | 145 | 1.1 | 135 | 1.2 |
| 5000 | 190 | 1.5 | 165 | 1.8 |
| 6000 | 230 | 2.0 | 200 | 2.5 |

-Since the power dissipated by the filament represents about 560 watts and since grid-plusscreen dissipation can, under some conditions, represent another 200 to 300 watts, allowance has been made in preparing this tabulation for an additional 1000 watts dissipation.

The blower selected in a given application must be capable of supplying the desired air flow at a back pressure equal to the pressure drop shown above plus any drop encountered in ducts and filters.

At higher altitudes, higher frequencies, or higher ambient temperatures the flow rate must be increased to obtain equivalent cooling. The flow rate and corresponding pressure differential must be determined individually in such cases, using maximum rated temperatures as the criteria for satisfactory cooling.

Filament Operation-The rated filament voltage for the 4 CX 5000 A is 7.5 volts. Filament voltage, as measured at the socket, should be maintained at this value to obtain maximum tube life. In no case should it be allowed to deviate by more than 5 percent from the rated value.

Electrode Dissipation Ratings-The maximum dissipation ratings for the 4 CX 5000 A must be respected to avoid darnage to the tube. An exception is the plate dissipation, which may be permitted to rise above the maximum rating during brief periods, such as may occur dering tuning.

Gontel Grid Operation- The 4CX5000A control grid has a maximum dissipation rating of 75 watts. Precautions should be observed to avoid exceeding this rating. The grid bins and driving power should be kept near the values shown in the "Typical Operation" sections of the data sheet whenever possible.

Screen-Grid Operation-The power dissipated by the screen of the 4 CX 5000 A must not exceed 2.50 watts.

Screen dissipation, in cases where there is no ac applied to the screen, is the simple product of the screen voltage and the screen current. If the screen voltage is modulated, the screen dissipation will depend upon loading, driving power, and carrier screen voltage.

Screen dissipation is likely to rise to excessive values when the plate voltage, bias voltage, or plate load are removed with filament and screen voltages applied. Suitable protective means must be provided to limit the screen dissipation to 250 watts in the event of circuit failure.

Plate Dissipation-The plate-dissipation rating for the 4 CX 5000 A is 5000 watts for most applications but for audio and SSB amplifier applications, the maximum allowable dissipation is 6000 watts.

When the 4 CX 5000 A is operated as a plate-modulated r-f power amplifier, the input power is limited by conditions not connected with the plate efficiency, which is quite high. Therefore, except during tuning there is little possibility that the 3500 -watt maximum plate dissipation rating will be exceeded.

Special Applications-If it is desired to operate this tube under conditions widely different from those given here, write to Power Crid Tube Marketing, Eitel-McCullough, Inc., 301 Industrial Way, San Carlos, California, for infurmation and recommendations.

0

$0$



## 830E， <br> $F, H$

REPORT 36652 FILE III PARTS LIST－TRANSCRIPT

| ASSEM $81 Y$ | PIECE PART | QUANTITY | ADD DEL | MC IT | ITM | Unid ${ }^{\text {P }}$ | SER | ADD R DEL R | PART NABE | DESCRIPTION | AL | $\begin{aligned} & P \\ & C \end{aligned}$ | AGE |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 762－8955－001 | 439－1355－000 | 9.000 | 733 | 00 | 003 | 302 | 7082 | N61464 | WIRE | FC35（FT） |  |  | ＋ |
|  | 762－8955－000 |  | 733 | 02 | 009 | 00 | 7062 | N61464 | CHART | From－TO |  | － | ． |
|  | 762－8955－001 | 1.000 | 733 | 05 | 000 | 07 | 7062 | No1464 | CAELE | FPCT TUEE KIT |  |  |  |
| Walk$762-8957-001$ |  |  | HICrast | ITEM |  | 9 丷⿻大丨日大 | DFTE 9 |  | CACNT | 10 Co．CCEE 03 | 3 | CGNIZ | 2 CD 00 |
|  | 000－0000－601 |  | 733 | 02 | 000 | 00 | 7233 | N61464 | ERACKET | WIRED MIZ Fino |  | OB | 1 |
| 762-8957-001 | 015－3000－010 | 1.000 | 733 | 01 | 008 | 07 | 7062 | 2161464 | OCUPLER | ． |  |  |  |
|  | 150－1542－000 | 2.000 | 733 | 00 | 019 | 07 | 7032 | Na1s64 | Clasp | LOPP |  |  |  |
|  | 230－0517－000 | 1.000 | 733 | 01 | 005 | 07 | 7062 | NS1464 | MOTCR | B403 |  |  |  |
|  | 302－0413－000 | 2.000 | 733 | 03 | 014 | 07 | 7062 | N61464 | WA5F38 | MICA |  |  |  |
|  | 310－0046－000 | 1.000 | 733 | 03 | 013 | 07 | 7062 | N61464 | WASHER | ก． 6 |  |  |  |
|  | 310－0048－000 | 1.000 | 733 | 03 | 015 | 07 | 7062 | N61484 | WASHER | $\text { N. } 8$ |  |  |  |
|  | 310－0282－000 | 8.000 | 733 | 03 | 004 | 07 | 7062 | N61464 | WASTR | NO． 6 LOCX |  |  |  |
| － | 311－0371－000 | 1.000 | 733 | 03 | 007 | 07 | 1062 | N81464 | PIN |  | ： |  |  |
|  | 313－0002－000 | 6.000 | 733 | 03 | 003 | 07 | 7062 | N61464 | NTT | NO． 6 |  |  |  |
|  | 328－0267－000 | 2.000 | 733 | 03 | 009 | 07 | 7082 | N61464 | Swiscsux | N |  |  |  |
|  | 342－0044－000 | 2.000 | 733 | 03 | 020 | 07 | 7062 | N61464 | SCREN | 4－40 X 1／4 PFH |  |  |  |
|  | 343-0167-000 | 5.000 | 733 | 03 | 012 | 07 | 1062 | N61464 | SCPE | $6-32 \times 1 / 4 \text { PFH }$ |  | 1 |  |
|  | 343－0169－000 | 1.000 | 733 | 03 | 017 | 07 | 1062 | N614S4 | SCREM | 6－32 $\times 3 / 8 \mathrm{PPH}$ |  |  |  |
|  | 500－1143－001 | 1.000 | 733803 | 01 | 016 | 07 | 1233 | N61464 T36115 | PCOT | RES NTG |  |  |  |
|  | 500－1143－007 | 1.000 | 803 | 01 | 016 | 07 | 1233 | T66115 | PCST | PE3 MTG |  |  |  |
|  | 540－9474－003 | 2.000 | 733 | 01 | 010 | 07 | 7062 | N61464 | STANDCFF | PEx |  |  |  |
|  | 553－7307－002 | 1.000 | 733 | 01 | 008 | 07 | 7082 | N61464 | Facles | Grime |  |  |  |
| $\cdots$ | 553－7310－003 | 1.000 | 733 | 01 | 018 | ＇07 | 7082 | N61464 | EPACKET | ADIOR NTG |  |  |  |
|  | 746－6131－000 | 1.000 | 737 | 01 | 011 | 07 | 7098 | T32370 | RES | 27 CM Pr194 |  |  | － |
|  | 747－5533－000 | 1.000 | 733137 | 01 | 011 | 07 | 7096 | N61464 T82370 | PESISTCP | 27 Cx RA94 |  |  |  |
| － | 762－8957－000 |  | 733 | 02 | 021 | 00 | 1062 | N81434 | EAACKET | HIRSD MIZ P4TE |  |  | － |
|  | 762－8957－001 | 1.000 | 733 | 05 | 000 | 07 | 7082 | N81464 | ERACKET | WIPED MIE FHEO |  |  |  |
|  | 762－8958－001 | 1.000 | 133 | 09 | 022 | 07 | 1062 | M61464 | CABLE |  |  |  |  |
|  | 762－8959－000 |  | 733 | 02 | 023 | 00 | 7062 | N61464 | SCrisuTl |  |  |  | ＋ |
|  | 970－2304－000 | 2.000. | .733 | 01 | 002 | 07 | 7062 | N61464 | REIAY | AnM K409 K410 |  |  |  |
|  |  |  |  | ITEM |  |  |  |  | CONT | 26 CO CCDC 03 | 3 | Wixiz | 2 CD 00 |



We have been notified by the vendor that after filling the present order for 970-1931-000 relays that he will not build any more. A search has been made for a direct substitute but none was found. It was decided, after consultation with Dick Uhrik, that a new panel be made and the new plug-in relay, which is widely used on other transmitters, used.

A new relay panel has been designed, $P / N$ 793-9713-001, which will mount in the same holes as the old panel. Whenever an order for a replacement 970-1931-000 relay is received, this panel will have to be sent.

A number of these panels should be built up and placed in stock. An addendum to all instruction books should be issued.


LEK: mt

## GENERAL:

This modification connects the Collins remote panel to the remote control relay circuits in the transmitter and adds a fail-safe circuit. The remote panel switch contacts will no longer parallel the control circuits of the transmitter. The filament-off circuit is connected in such a way that it will cause the transmitter to be turned off in the event of an open or short circuit in the remote equipment. This circuit over-rides all other control circuits. A relay is being added which prevents the power change motor from being operated if the motor control wires in the remote cable inadvertently make contact wi th the control voltage in the cable.

MODIFICATION PROCEDURE:
A. Remote panel changes:

1. Disconnect the wires on the PLATE-OFF switch 54 , and reconnect to the nommally open contacts of the same switch.
2. Disconnect the cable wires from $T B 1$, terminals $7,2,3,4,5,6$, 7, and 8. Also disconnect the cable wire from TB3-1.
3. Jumper together terminals $1,3,5$, and 7 on TB1.
4. Disconnect the wire from terminal 4 of 56 which connects to TB3-1.
5. Connect the "hot-side" of switch S2 to terminal 4 of S6.
6. Jumper S6-5 to S6-2 and 7 .
7. Jumper S6-1 to 8 .
8. Reconnect the RA7 wire, which was disconnected in 4 above, to S6-1.
B. B830-1 or D830-1 changes:
9. Move the wire connected to TB331-4 to TB331-6.
10. Remove the cable wires connecting to the remote unit at the following terminals:

$$
\begin{aligned}
& \text { TB302-8 } \\
& \text { TB302-9 } \\
& \text { TB303-1 Connecting to S1 FIL-0N } \\
& \text { TB303-2 Connecting to S1 FIL-0il } \\
& \text { TB303-3 } \\
& \text { TB303-4 } \\
& \text { TB303-5 } \\
& \text { TB303-7 } \\
& \text { TB334-2 or TB410-2 }
\end{aligned}
$$

3. Add a strap betricen TB302-9 and 10 if necessary. Add a strap between TB331-1 and 8.
4. Connect TB303-4 to TB332-2 if necessary.

NOTE: If the above wires were already installed, look for some other wire disconnected between TB302 and TB331; and TB303 and TB332. See schematic.
5. If the power change motor is located in the B830-1 or D830-1 cabinet, continue with Step $\mathcal{C}$; otherwise continue with Section $C$.
6. Add the power change activate relay K 311 and $T B 337$ in the mounting holes above Z301 and Z302.
7. Remove the connections between TB334-4 and TB322-4; and TB334-5 and T3322-5.
8. Connect TB322-4 to TB337-2 (iNew).
9. Connect TB334-4 to TB337-6 (New).
10. Connect TB322-5 to TB337-1 (New).
11. Connect TB334-5 to TB337-5 (Hew).
12. Connect TB333-3 to TB337-8 (ivew).
C. E830-1 or F830-1 changes: (5 or $10 \mathrm{kw} \mathrm{Fif} \mathrm{Transmitters} \mathrm{only)}$

1. Add the power change activate relay $K 411$ and $T 6337$ in the mounting holes to the left of TB410, TB405, and TB403.
2. Remove the connections between TB411-4 and TB410-4; and TB411-5 and TB410-5. (Wires will be re-used below)
3. Connect TB411-4 to TB337-2 (New).
4. Connect TB470-4 to TB337-6 (New).
5. Connect TB411-5 to TB337-1 (New).
6. Connect TB410-5 to TB337-5 (New).
7. Connect TB337-8 (New) to TB470-1.
8. Connect TB410-2 to TB403-8 (TB410-2 may be connected to 115 vac at some other point but it should be comected to 175 vac at Tis403-8.
9. Connect TB410-1 to TB410-3. (These points may already be connected).
D. Cable Connections: (Modification only)

| From | T0 |
| :--- | :--- |
| TB337-7 | TB3-1 |
| TB331-7 | TB1-2 |
| TB33T-10 | TB1-4 |
| TB332-7 | TB1-6 |
| TB32-10 | TB1-8 |
|  | TB3-10 -10 |

E. Principles of operation:
.
The FIL-ON and PLATE-ON switches apply 115 VAC to relays, which close contacts to activate the desired circuit. The PLATE-OFF switch energizes a relay which opens the plate control circuit. The filamentoff relay in the transmitter is held in the energized position continuously by 115 VAC supplied from the remote panel. The filament circuit is opened and the transmitter turned off by removing the 115 VAC from the filament relay. In the event of an open or short circuit in the cable, the filament relay would become de-energized automatically shutting off the transmitter. The filament-off control circuit will "over-ride" a malfunction in any of the other remote circuits. Relay K317 (or K417) disconnects the power control motor from the cable circuits except when the power is being changed by moving switch 56 on the remote panel. Unused contacts are connected by the above modification so that relay K 311 (or K 411 ) is energized, allowing power to be applied to the motor control circuits.
F. Circuit operation check:

- 1. Operate the FIL-ON switch Sl on the remote panel. The transmitter filament circuits should turn on.

2. Operate the FIL-OFF switch $S 2$. The filament circuits should turn off.
3. Operate the PLATE-ON switch S3. The plate and filament circuits should turn on.
4. Operate the PLATE-OFF switch $S 4$. The plate circuits should be turned off.
5. Turn on the plate circuits again. Operate the power change switch S6. Relay K311 (or K411) should energize and the motor run in one direction or the other.


COLINS RADIO COMPANY

DATE: 9-4-62
Page 1 of 3

EQUIPMENT TYPE: E830-1 POWER AMPLIFIER
SYSTEM USE: 830E-1/1A 5-KW FM BROADCAST TRANSMITTERS
SUBJECT: IMPROVEMENT IN OPERATION AND RELJABILITY OF BEAD CHAIN DRIVES (PA TUNING MECHANISM) AND PA CAVITY AIR FLOW VANE

The modification in this bulletin is to improve the operation and reliability of the bead chain drives and to stabilize the air flow around the PA cavity air-flow interlock switch to improve its operation. The performance of the bead chain drive is improved by installing a new set of tested bead chain belts, improving the alignment of bead chain drive, and adding refinements to the chain driving mechanism. Operation of the air-flow interlock switch is improved by adding a bracket around the vane to reduce turbulence of air flow and stabilize vane operation.

These changes are recommended by the manufacturer to reduce the possibility of the chain drive failing and to improve the operation of the air interlock switch. These changes will be factory installed in E830-1 units with serial numbers above 15 .

The estimated time required to perform this modification is 2 hours.
MODIFICATION PROCEDURE:

## Disassembly

1. Remove lower front door by disconnecting the two retaining chains and lifting door up and out.
2. Disconnect fan plug on lower rear cover and remove the rear cover assembly by releasing two side latches and lifting cover up and out.
3. Locate two lead screws at the inside corner of the tuned cavity and remove the stop washers attached to the end of each lead screw. Discard washers.
4. Open cavity door at front of unit and remove the plate resonator center conductor, PA tube, and chimney.
5. Using the knobs, back out each capacitor plate until lead screws are clear of drive sprockets.
6. At rear of unit, locate and remove the sprocket channel which is bolted to the corner of the cavity with four screws.
7. Remove sprocket retainer from channel.
8. Give some slack to the bead chain and push out the two drive sprockets.
9. Remove knobs from front of cavity.
10. Back out the three flat head screws in the sprocket housing, and remove back plate from inside cavity wall.
11. Remove sprocket housing with sprockets and chains.

## Reassemb1y

12. Reassemble sprocket channel with new bead chain belts (015-1888-00). Feed a loop of each belt through openings in one side of channel.
13. Thread the small sprockets through the loop into the sprocket bearings.
14. Center pads of the sprocket retainer over sprocket tapped holes and bolt in place.
15. Bolt the assembled sprocket channel to the cavity with the bead chains extending toward front of cabinet.

NOTE: A misalignment problem may exist in this channel assembly. It may be corrected as follows:

When replacing the channel, slip two $1 / 32$-inch thick washers (310-0055-00) under the left flange at each screw point. This, in effect, rotates the channel in a cuunterclockwise direction looking down on the channel. Sighting through the sprocket at the rectangular holes in the capacitor shield will give a good indication as to correction of the misalignment.
16. With the channel properly a夫tached and aligned, proceed by feeding the bead chains through the slot in the cavity flange to the knob positions.
17. From front of cabinet slip two new sprocket assemblies (549-2352-002) into che large holes in the cavity wall and slip on the respective bead chains.
18. Slip new sprocket housing (549-2353-003) onto sprocket shafts.
19. Temporarily attach knobs and start capacitor lead screws by rotating knobs forward while guiding screw until rectangular section is engaged in rectangular holes.
20. Screw capacitor plates in until approximately $1 / 8$ inch of lead screw protrudes beyond sprocket bearing at rear of cavity.
21. Attach two hex posts (540-9053-003) to new knob backing plate (549-2180-003) and secure with screws (342-0045-00).
22. Without moving knobs from above position, turn outer stops on sprocket assembly inside cavity to extreme counterclockwise position.
23. Now turn knob so that long arms of the outer stop washers point toward each other. A further clockwise turn of 30 degrees (looking at the knobs) of the knobs gives exact positioning desired for attaching back plate.
24. Attach backing plate (549-2180-00) with three flat head screws through the sprocket housing and adjust the chain tensions by pulling out on the sprocket housing and back plate assembly until most of the slack is out of the bead chains. Tighten three screws to secure.
25. Loosen knobs and set the turn counting device for each knob to zero and retighten.
26. Attach new r-f shield (553-5690-003) to backing plate hex posts using two pan head screws (343-0286-00) with lock washers (310-0396-00).
27. Locate air flow switch (vane) at top of PA cavity air stack.
28. Center air baffle bracket (549-1992-003) around air vane and drill two $3 / 16$-inch holes in cavity for mounting bracket. Mount bracket using two 6-32 machine screws (343-0169-00), two no. 6 lock washers (310-0071-00), and two 6-32 nuts (313-0002-00).
29. Check tuning mechanism to ensure that it works properly. Reinstall PA tube, chimney, and plate resonator center conductor in cavity.
30. Reassemble remainder of unit.

PARTS REQUIRED:
Price: \$59.53
Modification kit 553-5769-00 which consists of the following items:

| Qty | Description | Collins Part Number |
| :---: | :---: | :---: |
| 1 | Housing, sprocket-pressed | 549-2353-003 |
| 1 | Plate, backing-drive sprocket | 549-2180-003 |
| 1 | Shield, r-f | 553-5690-003 |
| 2 | Sprocket wheel, pinned assembly | 549-2352-002 |
| 2 | Belt, bead chain | 015-1888-00 |
| 2 | Post, hex | 540-9053-003 |
| 2 | Screw, machine, $4-40 \times 5 / 16$ flat head | 342-0045-00 |
| 2 | Screw, machine, $4-40 \times 5 / 16$ pan head | 343-0286-00 |
| 2 | Washer, lock no. 4 | 310-0396-00 |
| 2 | Screw, machine $6-32 \times 3 / 8$ | 343-0169-00 |
| 2 | Washer, lock no. 6 | 310-0071-00 |
| 2 | Nut, 6-32 hex | 313-0002-00 |
| 1 | Bracket, air baffle | 549-1992-003 |
| 4 | Washer, no. 6 flat | 310-0055-00 |
| The above parts may be obtained from Collins Radio Company, Service Parts Department, Cedar Rapids, Iowa at no charge for six months after the date of this bulletin. All orders should specify modification kit 553-5769-00 and make reference to E830 Service Bulletin No. 1. |  |  |
|  |  |  |

Engineering order
PAGE : OF $z^{7}$ E.O.NO. K-2019
Hee $5=62337$

i. Part Num ber Affected:

5492273000 , Control Panel, Rert th
Hensor for Chauge
(an no longer buy contactor.
( $A 1405$, is 406 )
Effectivity:
Use all old contactors on haud.
Lisit of Changes
$\%$ Changl. 40 , 1170 Soo, Melay,
G405, it $406($ Qty 2) to
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2. Pewort existing urits by using adip trer plate. when new contacfor is uskd, (FOs-o6ン4 0.30.)



List of ryaterais:

Qty Itim PN

| $2^{\prime}$ | 1 | 3420182000 |
| :---: | :---: | :---: | :---: |
| 1 | 2 | 3420186000 |
| 2 | 3 | 3430,185000 |
| 1 | 4 | 3130017000 |
| 3 | 5 | $3,000,72000$ |
| $0-6464(11-63)$ |  |  |

Pescry, Hois
Screw 8-32 $\times \frac{1}{4}$ Pfit
Serew $8-32 \times \frac{1}{2}$ PFIt
Screw 8-32 8 年 PPH
likt s-32
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ENGINEERING ORDER CONTINUATION SHEET
PAGE 3 OF 3 E.O. NO. K-0019
Assembly Details:



Part Numbers Affected:
5492273 oo, Control Panel, Rev L.
5492269 oo, Panel, Rev
Reason for Change:
The metering, resistor for the screen voltage is not close enough to the required value. Meters read out of tolerance.
Effectivity
All units with meters reading out of tolerance.
List of Changes:
$\%$ Add, 147 dian $n_{i m i d}^{i^{k i}}$ way on a line between E420 and E418. Fig!.
2. Install 306097900 Terminal Qty l, Using screw, 6-32x 5/6, 343 $0329.000, Q \times y$, and Washer, N10 6 Lock, Qty 1, 3100077000 . Figs $1+2$
3: A dd Qty 2, Resistor, 402 t. 1\%亿 w, 705 3287000 , Mount in series between E420 and E418. Fig 2 .

D-646 (11-63) !f more space is needed, use form mo. D-64ga, engine bring order continuation sheet.
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insuruction book

Cedar Rapids Division | Collins Radio Company. Cedar Rapids, lowa
A. A OTT, Cocelafs Ramies DAREAS TEX, AAM STM 429.017 830F-1A
10.K w FM Broadcast

Transmitter

## Guarantee

The equipment described herein is sold under the following guarantee:
Collins agrees to repair or replace, without charge, any equipment, parts, or accessories which are defective as to design, workmanship or material, and which are returned to Collins at its factory, transportation prepaid, provided:
(a) Notice of the claimed defect is given Collins within one (1) year from date of delivery and goods are returned in accordance with Collins instructions.
(b) Equipment, accessories, tubes, and batteries not manufactured by Collins or from Collins designs are subject to only such adjustments as Collins may obtatn from the supplier thereof.
(c) Failure of the equipment to operate in a normal and proper manner due to exposure to any environmental condition in excess of those published in the equipment specification shall not be deemed a defect within the meaning of this clause.

Collins further guarantees that any radio transmitter described herein will deliver full radio frequency power output at the antenna lead when connected to a suitable load, but such guarantee shall not be construed as a guarantee of any definite coverage or range of said apparatus.

The guarantee of these paragraphs is void if equipment is altered or repaired by others than Collins or its authorized service center.

No other warranties, expressed or implied, shall be applicable to any equipment sold hereunder, and the foregoing shall constitute the Buyer's sole right and remedy under the agreements in this paragraph contained. In no event shall Collins have any liability for consequential damages, or for loss, damage or expense directly or indirectly arising from the use of the products, or any inability to use them either separately or in combination with other equipment or materials, or from any other cause.

## How to Return Material or Equipment

If, for any reason, you should wish to return material or equipment, whether under the guarantee or otherwise, you should notify us, giving full particulars including the details listed below, insofar as applicable. If the item is thought to be defective, such notice must give full information as to nature of defect and identification (including part number if possible) of part considered defective. (With respect to tubes we suggest that your adjustments can be speeded up if you give notice of defect directly to the tube manufacturer.) Upon receipt of such notice, Collins will promptly advise you respecting the return. Failure to secure our advice prior to the forwarding of the goods or failure to provide full particulars may cause unnecessary delay in the handling of your returned merchandise.

## ADDRESS:

Collins Radio Company
Product Support Division
Cedar Rapids, Iowa

## INFORMATION NEEDED:

(A) Type number, name and serial number of equipment
(B) Date of delivery of equipment
(C) Date placed in service
(D) Number of hours of service
(E) Nature of trouble
(F) Cause of trouble if known
(G) Part number (9 or 10 digit number) and name of part thought to be causing trouble
(H) Item or symbol number of same obtained from parts list or schematic
(I) Collins number (and name) of unit subassemblies involved in trouble
(J) Remarks

## How to Order Replacement Parts

When ordering replacement parts, you should direct your order as indicated below and furnish the following information insofar as applicable. To enable us to give you better replacement service, please be sure to give us complete information.

ADDRESS:
Collins Radio Company Product Support Division
Cedar Rapids, Iowa

## INFORMATION NEEDED:

(A) Quantity required
(B) Collins part number (9 or 10 digit number) and description
(C) Item or symbol number obtained from parts list or schematic
(D) Collins type number, name and serial number of principal equipment
(E) Unit subassembly number (where applicable)

## SERVICE $B U L \mathbb{E} \mathbb{E} T I N$

SERVICE BULLETIN NO. 5

EQUIPMENT SERIES: 830
EQUIPMENT TYPE: 830B-1A through 830F-1A Broadcast Transmitters
SUBJECT: Installation of Striker Plate

This modification is recommended when a replacement door is installed.

## 1. MODIFICATION PROCEDURE

a. Locate and drill two 0.108-diameter holes (No. 36 drill) and tap for 6-32 screws in the cabinet as shown in figure 1.
b. Install magnetic striker plate on the cabinet as shown in figure 1. Use 6-32 X 1/4 PFH screw, 330-2295-000. Be sure the screw head is below the surface of the mounting plate. c. Repeat steps a . and b . for the opposite side of the transmitter cabinet.
2. PARTS REQUIRED

Price: \$2.04
Modification kit 962-9044-002 consists of the following parts:
QUANTITY DESCRIPTION COLLINS PART NUMBER
4 Screw, 6-32 X 1/4 PFH 330-2295-000
2 Plate, Striker 762-9041-002
The above parts may be secured from Service Parts Department, Collins Radio Company, Dallas, Texas 75207, at the indicated price. Orders should be for modification kit (762-9044-002), and the model of the transmitter should be included.


Figure 1. Installation of Striker Plate.

# To: Gart Bowling <br> location: 401-021 <br> - 

Harry Mins

FROM:

Date:
January 31, 1967
subuect:
Modifications made in the A830-2
reference:
The following changes were made in the $\mathrm{A} 830-2$ Exciter to restore its frequency stability and improve operating conditions:

1. The bus wire lead from pin 1 of $V 427$ to $L 429$ was routed with excessive length. It now is routed from V427 pin 6 through the hole in terminal 3 of TB429 directly to terminal 1 of 1429 .
2. An additional ground terminal was added at GRD 432 to shorten lead length from V428-2 to ground and shorten lead length on bypass capacitor C439.
3. The 100 K grid resistor ( R 441 ) in grid circuit of 7428 was deleted. A 1 K resistor ( R 441 ) and bypass capacitor ( $\mathrm{C} 492-4700 \mathrm{pf}$ ) was added to the meter end of R 442 (10K) and ground,making R 442 the grid resistor for V428.
4. V428 screen bypass capacitor (C443) and V428 cathode bypass capacitor (C441) leads were shortened to a maximum length of $1 / 8$ inch.
5. R446 (100K) was deleted from grid circuit of V429 (5763). R446 (1K) and C493 ( 4700 pf ) were added to the meter end of R 447 and ground, making R 447 the grid leak for V429.
6. V429 cathode resistor (R448) was changed from 270 ohms to 470 ohms. mis. 270
7. V429 cathode bypass capacitor ( 4447 ) was changed from 4700 pf to 4770 pf . N: \% \% 1000 F
8. V429 screen dropping resistor (R449) was changed from 10 K ohms to 18 K ohms. $\ldots$..... 27 K
9. V429 screen bypass capacitor (C448) was changed from 4700 pf to 470-pf. It was found that the self-resonant frequency of 470 pf capacitor was approximately,. 94 MHz , making it more tideally suited for bypass use at these frequencies.
10. $C 444$ and $C 446$ ( 4700 pf ) were placed $f 1$ at against the chassis and lead length reduced to $1 / 8$ inch or less.

V11. C 452 was changed from 33 pf to 39 pf and placed between L434-1 and GRD 434 according to the DO-FROM chart, CPN 549-1588-001.
12. $C 453$ and $C 460$ (4700 pf) were placed according to 549-1588-001.

Memo to Gart Bowling
-2-
January 31, 1967
13. $C 494$ and $C 495$ ( 4700 pf ) were added to the metering leads from the grid and the cathode circuits of $V 430,(2 \mathrm{E} 26)$ to keep any RF off these meter leads.
14. All six coils $L 429$ through 4434 had one turn removed and will now cover the tuning range with no additional capacitors.
15. The tube shields (4) were replaced with MS tube shields made by I.E.R.C. and already have good Collins part numbers

Harry Mims

HS; kht
cc: File
C. Dixon
S. Publicover
unit instructions
Stolen frem
A.R.Nett

A830-2
10W Wide-Band FM Broadcast Exciter
©Collins Radio Company 1962. 1964

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## SECTION I GENERAL DESCRIPTION

### 1.1 GENERAL.

This instruction book contains information for operation and maintenance of A830-2 10 W Wide-Band FM Broadcast Exciter. See figure 1-1. The $A 830-2$ is manufactured by Collins Radio Company. Cedar Rapids. Iowa.

### 1.2 PURPOSE OF EQUIPMENT.

The A830-2 10 W Wide-Band FM Broadcast Exciter is a direct FM exciter designed specifically to meet the stringent requirements of stereophonic FMbroadcasting. The A830-2 may be used in monaural broadcasting, Storecasting (SCA), or with Collins $786 \mathrm{M}-1$


Figure 1-1. A830-2 10 W Wide-Band FM Broadcast Exciter. Over-all View

Stereo Generator (optional) for stereophonic broadcasting. The A830-2 is used to drive higher power amplifiers in the FM broadcast service.

### 1.3 EQUIPMENT SUPPLIED.

The A830-2 is normally supplied as a part of a Collins FM transmitter (830B-1A, 830D-1A, 830E-1A, etc.).

The A830-2 mounts in the same cabinet as the first stage of amplification ( 250 or 1000 watts) in the transmitter. A rear view of the A830-2 is shown in figure 1-2.

### 1.4 EQUIPMENT REQUIRED BUT NOT SUPPLIED.

The A830-2 is supplied with all required equipment.

### 1.5 TECHNICAL SUMMARY.

Ambient temperature range . . . . . . . . . . . . . . $+10^{\circ} \mathrm{C}\left(+50^{\circ} \mathrm{F}\right)$ to $+55^{\circ} \mathrm{C}\left(+131^{\circ} \mathrm{F}\right)$.
Ambient humidity range . . . . . . . . . . . . . . . . 0 to 95 percent relative.
Altitude . . . . . . . . . . . . . . . . . . . . . . . 7500 feet, maximum.
Shock and vibration . . . . . . . . . . . . . . . . . . Normal handing and transportation.
Power source . . . . . . . . . . . . . . . . . . . . 117 volts $\pm 5$ percent, $50 / 60 \mathrm{cps}$, single phase.
R-f power output . . . . . . . . . . . . . . . . . . . Adjustable to 10 watts into a $50-$ to 70 -ohm resistive load.
Frequency range . . . . . . . . . . . . . . 88 to 108 mc . Customer frequency is determined by
one crystal in the heterodyning oscillator circuit.
Harmonic and spurious radiation. . . . . . . . . . Any emission appearing on a frequency removed from
the carrier by between 120 kc and 240 kc , inclusive,

Any emission appearing on a frequency removed from the carrier by more than 240 kc up to and including 600 kc is attenuated at least 40 db below the level of the unmodulated carrier.

Any emission appearing on a frequency removed from the carrier by more than 600 kc is attenuated at least 80 db below the level of the unmodulated carrier, with the exception of harmonics of the r-f carrier which complies with the requirements of the particular transmitter in which the A830-2 is installed.

Type of modulation . . . . . . . . . . . . . . . . Frequency modulation. 100 percent modulation is defined as $\pm 75-\mathrm{kc}$ deviation of the main carrier.

Exciter inputs . . . . . . . . . . . . . . . . . Stereophonic channel: 600 ohms, unbalanced. Input of 0.1 volt (approximately) required for 100 percent modulation.

Monophonic channel: 600 ohms, balanced. Input of $10 \mathrm{dbm} \pm 2 \mathrm{db}$ (approximately 2.45 volts) required for 100 percent modulation.

SCA channel: 600 ohms, balanced. Input of 0.35 volt (approximately) required for 10 percent modulation.
Frequency and phase response . . . . . . . . . . . . The frequency and phase response of the A830-2 is
such that when used with a suitable stereophonic
generator such as the $786 \mathrm{M}-1$, stereophonic separation
between left and right stereophonic channels shall be
better than 35 db at audio modulating frequencies
between 30 and $15,000 \mathrm{cps}$.

### 1.6 VACUUM-TUBE, FUSE, AND SEMICONDUCTOR COMPLEMENT.

Table 1-1 lists all of the vacuum tubes, fuses, and semiconductors used in the A830-2.

TABLE 1-1. VACUUM-TUBE, FUSE, AND SEMICONDUCTOR COMPLEMENT

| SYMBOL | TYPE | . FUNCTION |
| :---: | :---: | :---: |
| V426 | 6U8A | Oscillator and buffer |
| V427 | 12 AT 7 | Balanced mixer |
| V428 | 6AU6 | Limiter-amplifier |
| V429 | 5763 | Driver |
| V430 | 2E26 | Power amplifier |
| Q501 | 2N1225 | First afc limiter |
| Q502 | 2N1225 | Second afc limiter |
| Q503 | 2N708 | Afc discriminator driver |
| Q504 | 2N1613 | First error signal amplifier |
| Q505 | 2N1613 | Second error signal amplifier |
| Q506 | 2N1613 | Third error signal amplifier |
| Q507 | 2N1613 | Fourth error signal amplifier |
| Q508 | 2N491 | Keying generator |
| Q509 | 2N1605 | Multivibrator |
| Q510 | 2N1605 | Multivibrator |
| Q511 | 2N1175A | Baseband cancellation amplifier |
| Q601 | 2N1396 | Frequency modulated oscillator |
| Q602 | 2N1225 | First limiter |
| Q603 | 2N1225 | Second limiter |
| Q604 | 2N708 | Discriminator driver |
| Q605 | 2N1225 | Afc buffer |
| Q606 | 2N708 | Modulator output amplifier |
| Q607 | 2N1711 | First baseband amplifier |
| Q608 | 2N1396 | Second baseband amplifier |
| CR401 | 1N1492 | $\mathrm{B}+$ rectifier |
| CR402 | 1N1492 | B+ rectifier |
| CR403 | 1N1492 | $\mathrm{B}+$ rectifier |
| CR404 | 1N1492 | $\mathrm{B}+$ rectifier |
| CR405 | 1N1492 | B+ rectifier |

TABLE 1-1. VACUUM-TUBE, FUSE, AND SEMICONDUCTOR COMPLEMENT (Cont)

| SYMBOL | TYPE | FUNCTION |
| :---: | :---: | :---: |
| CR406 | 1N1492 | B+ rectifier |
| CR407 | 1N1492 | B+ rectifier |
| CR408 | 1N1492 | $\mathrm{B}+$ rectifier |
| CR409 | 1N538 | +20-volt rectifier |
| CR410 | 1N538 | +20-volt rectifier |
| CR411 | 10M10ZB1 | +20-volt regulator |
| CR412 | 1Z10V01 | +10 -volt regulator |
| CR413 | 1N538 | -10-volt rectifier |
| CR414 | 1N538 | -10-volt rectifier |
| CR415 | 1Z10V01 | -10-volt regulator |
| CR426 | 1N977A | Oscillator plate voltage regulator |
| CR501 | 1 N 270 | Gate |
| CR502 | 1N270 | Gate |
| CR503 | 1N270 | Afc limiter |
| CR504 | 1N270 | Afc limiter |
| CR505 | 1N270 | Afc limiter |
| CR506 | 1N270 | Afc limiter |
| CR507 | 1N198 | Afc discriminator |
| CR508 | 1N198 | Afc discriminator |
| CR509 | FA-4000 | Synchronous detector |
| CR510 | FA-4000 | Synchronous detector |
| CR511 | FA-4000 | Gate |
| CR512 | 1N198 | Meter rectifier |
| CR513 | 1N198 | Meter rectifier |
| CR514 | 1N718 | Voltage regulator |
| CR601 | 1N626 | Temperature compensation |
| CR602 | SV3173 | Voltage regulator |
| CR603 | 1N270 | Limiter |
| CR604 | 1N270 | Limiter |
| CR605 | 1N270 | Limiter |
| CR606 | 1N270 | Limiter |
| CR607 | 1N198 | Modulation discriminator |
| CR608 | 1N198 | Modulation discriminator |
| CR609 | 1N751A | Voltage regulator |
| CR610 | 1N198 | Meter rectifier |
| F401 | 1 amp | Protect T401 |
| F402 | 1/4 amp | Protect T402 |



Figure 1-2. A830-2 10 W Wide-Band FM Broadcast Exciter, Rear View

# SECTION II PRINCIPLES OF OPERATION 

### 2.1 GENERAL.

This section describes the principles of operation of A830-2 10 W Wide-Band FM Broadcast Exciter. Figure $2-2$ is a block diagram of the A830-2 and figure 5-1 is the schematic diagram of the A830-2. Refer to these figures for the following discussion.

### 2.2 FREQUENCY MODULATION METHODS.

There are two basic methods used to generate an FM signal, direct FM and phase modulation. There are variations of each of these two methods, but the end results are the same.

### 2.2.1 PHASE MODULATORS.

The phase modulation method consists of phase modulating a CW (continuous wave) signal with audio tones. The audio response is shaped to drop off 6 db per octave from the lowest to the highest frequency. The resultant signal is frequency modulated although produced by a phase modulator. The modulation index of an FM signal is defined as the ratio of the change in carrier frequency (deviation) to the modulating frequency, $\frac{\nabla \mathrm{f}}{f_{\mathrm{m}}}$. The modulation index of present phase modulators is so low that modulation is usually performed at a low frequency (approximately 100 kc ) and then multiplied about 800 times to obtain the output frequency with the desired $\pm 75-\mathrm{kc}$ deviation. The outstanding advantage of this system is that the 100-kc oscillator may be crystal controlled and further frequency stabilization is not required. This
system has been used widely in broadcast FM transmitters in the past.

The arrival of stereophonic FM broadcasting has caused problems in the phase modulator. A composite stereo plus SCA signal (referred to hereafter as the baseband audio signal) occupies a frequency band from 50 cps to 75 kc . The audio response shaping ( 6 db per octave) would require that $50-\mathrm{cps}$ signals be 65.5 db above signals at 75 kc . When a signal-tonoise ratio of 65 db and a dynamic range of approximately 60 db is added to this, it is obvious that baseband amplifiers cannot be built to meet these requirements.

It is possible to split the phase modulation into two steps where one phase modulator accepts only the $\mathrm{L}+\mathrm{R}$ (left and right audio signals) audio spectrum and a subsequent modulator adds the $L-R$ double-sideband suppressed carrier signal. The audio bandwidth for each phase modulator is thereby reduced and the dynamic range of the baseband amplifiers is reduced to acceptable limits. The phase and amplitude relationships must be maintained between the two signals. These requirements are $\pm 0.3-\mathrm{db}$ gain variation and $\pm 3$-degree phase variation to meet the $30-\mathrm{db}$ stereo separation requirement. These requirements would be difficult to obtain without frequent on-the-air adjustment to continually meet the stereo separation requirement.

There are other methods of splitting the signal and using more than one modulator, but all have the phase and gain stability problem.


Figure 2-1. Direct FM Modulation, Simplified Block Diagram

### 2.2.2 DIRECT FREQUENCY MODULATION.

The direct method of generating a frequency modulated signal is shown in figure 2-1. The complete stereo signal (and SCA signal if used) is fed through a baseband amplifier to a frequency-modulated oscillator. The discriminator completes an audio feedback loop which suppresses FM oscillator distortion, incidental $F M$ noise, transient carrier offset, and gain/phase variation in the baseband amplifier and modulator. The center frequency of the oscillator is not sufficiently stable so an automatic frequency control (afc) circuit is required to maintain frequency stability. The output of the modulator is a $14-\mathrm{mc}$ FM signal with $\pm 75-\mathrm{kc}$ peak deviation. The output frequency is obtained by translating this signal with a stable vhf oscillator. The use of the direct FM system removes the requirement for double modulators, phase delay lines, and baseband amplifiers with a response which changes with frequency.

### 2.3 BIOCK DIAGRAM.

Refer to figure 2-2, a block diagram of the A830-2.

### 2.3.1 MODULATOR.

The A830-2 uses the direct FM method of generating an FM signal. The baseband input (and SCA input, if used) is connected to baseband amplifiers Q607 and Q608. The response of these amplifiers is flat.

The gain of the baseband amplifiers is adjustable with AMPL BIAS control R641. Refer to figure 5-1. The emitter voltage on $Q 608$ is regulated to +15 volts by a silicon breakdown diode, CR609. The output of Q608 is coupled to frequency-modulated oscillator Q601. Q601 is an LC oscillator which has a center frequency of 14 mc . The tuned circuit in the base of Q601 contains a voltage-sensitive capacitor, C654. Refer to figure 5-1. The capacitance of C654 varies proportionately with the voltage across it. The change in capacity of C654 makes a corresponding change in the frequency of oscillations in Q601. Thus, the frequency deviation of the output of Q601 is directly proportional to the amplitude of the modulating signal and the peak deviation is $\pm 75 \mathrm{kc}$.

The output of Q601 is coupled to two limiters, Q602 and Q603. The limiters remove any amplitude modulation from the FM signal. This amplitude modulation is caused by variation of the tuned circuit capacity by the baseband signal. The transistors do not do any limiting. The limiting takes place in the diodes connected to the collectors. This method provides symmetrical limiting (positive and negative) which avoids the phase modulation that occurs when unsymmetrical clipping followed by filtering is used. The limiters are set up so that as the input level is raised, the second limiter operates first; just before it becomes nonlinear, the first limiter starts limiting. The limiting range is approximately 31 db .

The output of the second limiter is coupled to discriminator driver Q604. One output of the discriminator driver is connected to modulator discriminator T601 and the other output goes to output amplifier Q606.

Modulator discriminator T601 converts the frequencymodulated $14-\mathrm{mc}$ signal to an AM signal which is detected by diodes CR607 and CR608. The detected audio is mixed with the input baseband audio at the input to the baseband amplifiers. This feedback loop suppresses distortion from the FM oscillator, incidental FM noise, transient carrier offset, and gain/phase variation in the baseband amplifier and modulator.

Output amplifier Q606 provides a signal output of 1.0 volt rms for the balancedmixer in the power amplifier compartment. This output is matched to 50 ohms by an L-section impedance, L611 and C634. A low-pass filter, C632, C633, and L610, attenuates harmonics of the $14-\mathrm{mc}$ signal. A portion of this output is rectified and connected to meter switch S101 for monitoring purposes.

The second output from Q606 is coupled to afc buffer amplifier Q605. This amplifier, as well as the limiters and amplifiers preceding it, reduces oscillator frequency change caused by variation of loading on the output. The output of Q605 is 0.1 volt rms across 50 ohms.

### 2.3.2 AUTOMATIC FREQUENCY CONTROL.

The A830-2 requires automatic frequency control to maintain the center frequency of the modulated oscillator at 14 mc . The error in frequency of this oscillator may be caused by temperature drift, carrier shift due to distortion in the modulator, etc. The afc circuits correct these errors to bring the stability of the output frequency to $\pm 1000$ cycles per second over a temperature range of $+10^{\circ} \mathrm{C}\left(+50^{\circ} \mathrm{F}\right)$ to $+55^{\circ} \mathrm{C}\left(+131^{\circ} \mathrm{F}\right)$ and a line voltage range of $\pm 5$ percent.

The afc correction voltage is obtained by comparing the modulator output signal with the output of a crystal-controlled reference oscillator, and deriving a d-c voltage which is proportional in magnitude and polarity to the magnitude and direction of the difference in frequency of these two signals.

The reference oscillator is a conventional crystalcontrolled oscillator using a fundamental $14-\mathrm{mc}$ series-resonant crystal. The temperature drift of this crystal contributes only $\pm 70$ cycles per second to the output frequency drift over temperature.

The signal from afc buffer Q605 and the output from the reference oscillator are connected to a diode switch, CR501 and CR502. The diode switch is simply two diodes which are alternately switched on and off by the 5 -cps square wave. The diode switch is controlled by a signal from keying generator Q508. This
signal, a square wave with a frequency of approximately 5 cps , alternately couples the reference signal, then the modulated carrier, to the input to first limiter Q501.

The two limiters, Q501 and Q502, and discriminator driver Q503 are identical to the limiters and driver (Q602, Q603, and Q604) used in the modulator. The limiters remove any amplitude difference which might exist between the two signals. The level of the reference signal is adjustable with REF LEVEL control R572. Q503 amplifies the limited signal to a level sufficient to drive the afc discriminator. Assume that there is no modulation applied. In this case, the output from the discriminator will be a $5-\mathrm{cps}$ square wave with an amplitude proportional to the frequency error in the FM oscillator.

The 5 -cps error signal is amplified and applied to the synchronous detector which develops the d-c correction voltage. This d-c voltage is coupled through a low-pass filter to the voltage-sensitive capacitor in the frequency-modulated oscillator to tune the FM oscillator back on frequency.

The operation of the afc circuitry is only slightly different when modulation is applied at $\pm 75-\mathrm{kc}$ deviation. Assume now that modulation is applied and an error of 100 cps exists in the FM oscillator. The output of the afc discriminator due to the $100-\mathrm{cps}$ signal would be $100 \mathrm{~K}_{\mathrm{d}}$ where $\mathrm{K}_{\mathrm{d}}$ is the gain of the discriminator in volts per cps. The output of the discriminator due to the modulation on the carrier would be $150,000 \mathrm{~K}_{\mathrm{d}}$. This means that the undesired signal is 1500 times greater than the desired signal. The undesired signal is removed by the modulation canceling circuit consisting of baseband cancel amplifier Q511 and diode switch CR511. Whenever the modulated carrier is connected to the first limiter diode switch, CR501 and CR502, the baseband audio input is connected to the discriminator output by diode switch CR511. This baseband audio is 180 degrees out of phase with the discriminator output, and when MOD BAL control R652 is properly adjusted, the output of the afc discriminator due to modulation is completely canceled. The 5 -cps error signal due to the frequency error in the FM oscillator is then amplified and detected as if modulation were not applied.

Note that the afc discriminator is used as a comparator rather than as a reference. The exact center frequency of the discriminator is not important since the output voltage need only be proportional to the difference in the two frequencies rather than to the absolute value of these frequencies. Therefore, the center frequency stability of the discriminator does not effect the operation of the afc system.

The last stage of the error signal amplifiers, Q507, is a phase splitter to provide a push-pull output to the synchronous detector. The synchronous detector
is keyed by the 5 -cps square-wave keying signal from the keying generator.

The synchronous detector recovers the information contained in the amplitude and phase of the 5 -cps error signal. The circuit used in the A830-2 is actually two synchronous detectors operating from opposite half cycles of the 5 -cps square-wave keying signal so that the $5-\mathrm{cps}$ square-wave keying signal is balanced out in the output. This is analogous to a double-sideband balanced modulator in which neither input signal is present in the output.

Figures 2-3 through 2-5 illustrate the operation of the two diode switches and the synchronous detector. The electronic circuit and a mechanical analog for each of the circuits is shown. The resistances marked $\mathrm{R}_{\mathrm{f}}$ represent the forward resistance of the diodes.

The output of the synchronous detector may be disabled for test and adjustment by depressing AFC DISABLE switch S102 on the front panel.

### 2.3.3 POWER AMPLIFIER.

The $14-\mathrm{mc}$ FM signal from the modulator is coupled to a balanced mixer, V427. The other input to V427 is the amplified output of a vhf crystal oscillator, V426A. The crystal oscillator operates with a fifthovertone series-resonant crystal in the 74- to $94-\mathrm{mc}$ frequency range. The specific frequency of the crystal is 14 mc below the station's assigned output frequency. The exact frequency is adjustable over a small range by VHF OSC FREQ ADJ control C427. This adjustment is required to compensate for the finishing tolerance and aging in crystals Y426 and Y501. The output of V426A is amplified in V426B and coupled to V427. The two input signals are balanced out of the output of V427 and the sum of the two signals is the operating frequency. The MIX BAL control compensates for unbalance between the sections of V427.

The output of $V 427$ is coupled to limiter amplifier V428. The limiter amplifier removes any amplitude modulation resulting from mixing and couples this signal to driver stage V429. The signal is amplified by V429 to a level sufficient to drive power amplifier stage V430. The power output is adjustable with POWER OUT control R454. The tuning and loading of the output stage is accomplished with C461 and C456.

### 2.3.4 POWER SUPPLY.

The power supply in the A830-2 provides all operating voltages for the $\mathrm{A} 830-2$ and $786 \mathrm{M}-1$ Stereo Generator, if used. The primary power may be 115 or 230 volts, 60 cps . The power supply is of conventional design using a bridge rectifier and a voltage divider for the high voltages. The low voltages are obtained from full-wave rectifiers. Voltage breakdown diodes are used for regulating the +20 -volt, +10 -volt, and $-10-$ volt outputs to $\pm 5$ percent.



MECHANICAL ANALOG

Figure 2-3. Reference Switch, Simplified Schematic and Mechanical Analog Diagram
,



Figure 2-4. Baseband Cancel Switch, Simplified Schematic and Mechanical Analog Diagram


Figure 2-5. Synchronous Detector, Simplified Schematic and Mechanical Analog Diagram

# SECTION III <br> MAINTENANCE 

### 3.1 GENERAL.

This section contains alignment instructions, adjustment procedures, and minimum performance standards for the A830-2.

### 3.2 TEST EQUIPMENT REQUIRED.

The test equipment in table 3-1, or its equivalent, is required to perform the procedures given in this section.

TABLE 3-1
TEST EQUIPMENT REQUIRED

| ITEM | MANUFACTURER'S <br> DESIGNA TION |
| :--- | :--- |
| Audio oscillator <br> Distortion and noise meter <br> A-c vtvm <br> R-f vtvm* <br> Communications receiver | Hewlett-Packard 200AB <br> Hewlett-Packard 330D <br> Hewlett-Packard 410B <br> Bird 91C <br> Capable of receiving <br> 14 mc <br> Microlab AD-10N |
| $10-$ db pad |  |
| Oscilloscope |  |
| FM monitor |  |
| $50-$ ohm load |  |$\quad$| Hewlett-Packard 335B |
| :--- |
| *The 91C is not required if a Tektronix 541 oscillo- |
| scope is available. See paragraph 3.3.10. |

### 3.3 ALIGNMENT AND ADJUSTMENT.

CAUTION
Do not make any adjustment in the modulator or automatic frequency control sections of the A830-2 unless trouble has definitely been traced to misadjustment.

### 3.3.1 PRELIMINARY ADJUSTMENTS.

Perform the following procedure prior to performing any of the alignment procedures.
a. Set the meter switch on the A830-2 to the OFF position.
b. Short AFC DISABLE switch S102 on the A830-2 with a clip lead.
c. Connect the 50 -ohm load to RF OUTPUT jack J402.
d. Operate POWER switch S401 to the ON position. Allow 10 minutes for equipment warm up.

### 3.3.2 MODULATOR LIMITER-DISCRIMINATOR ALIGNMENT.

a. Remove Q601 from its socket.
b. Rotate REF LEVEL control R572 fully counterclockwise.
c. Connect a 0.01 -uf capacitor and clip lead between the movable arm of REF LEVEL control R572 and the emitter pin on the socket for Q601. This supplies an accurate $14-\mathrm{mc}$ signal for alignment of the A830-2.
d. Connect the HP-410B to TP602 and set it to the lowest d-c scale.
e. Rotate R572 clockwise until an indication is observed on the HP-410B.

## NOTE

During this adjustment, maintain the 14 -mc signal at a level below limiting. Limiting causes the tuning peaks to be very broad.
f. Adjust C639, L606, and L603 for maximum indication on the HP-410B.
g. Remove the 0.01 -uf capacitor and clip lead from XQ601 and R572, Replace Q601 into XQ601.
h. Connect the 91C to TP504.

## NOTE

Refer to note in paragraph 3.3.10.
i. Remove Q509 from its socket.
j. Adjust R572 for an indication of 30 millivolts. k. Replace Q509.

### 3.3.3 MODULATOR OUTPUT AMPLIFIER TUNING.

a. Set the meter selector switch on the front panel of the A830-2 to the MOD OUTPUT $B$ position.
b. Tune L611 for maximum indication on the front panel meter.

### 3.3.4 AFC BUFFER TUNING.

a. Connect the 91C (or Tektronix oscilloscope) to TP504.
b. Remove Q510 from its socket.
c. Tune L 608 for maximum indication on the 91 C (or oscilloscope).
d. Replace Q510 into its socket.

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### 3.3.5 FM OSCILLATOR ADJUSTMENT.

a. Loosely couple the communications receiver to FM oscillator Q601 and to the 14 -mc reference oscillator. If the receiver has a bfo, turn it off.
b. Adjust OSC FREQ control C606 for a zero beat on the communication receiver.
c. Remove the communications receiver.

### 3.3.6 MODULATION DISCRIMINATOR.

a. Connect the HP-410B to TP601.
b. Adjust DISCR SEC control C644 for a zero indication on the HP-410B.
c. Check adjustment of DISCR PRI control C639. It should be set for a maximum indication, and C644 set for a minimum indication.

### 3.3.7 AMPLIFIER BIAS ADJUSTMENT.

a. Connect the HP-410B to TP603.
b. Adjust R641 for an indication of +7.5 volts d-c.

### 3.3.8 MODULATOR GAIN ADJUSTMENT.

a. Remove the 50 -ohms load and connect the HP-335B to the output of the A830-2 through the $10-\mathrm{db}$ pad.
b. Connect the HP-200AB to baseband input jack J604.
c. Set the output level of the HP-200AB to 0.1 volt rms at 1000 cps .
d. Adjust MOD GAIN control R635 for an indication of 100 percent modulation ( $\pm 75-\mathrm{kc}$ deviation) on the HP-335B.

### 3.3.9 AFC LIMITER-DISCRIMINATOR ALIGNMENT.

a. Connect the HP-410B to TP501 and set to 10 -volt scale.
b. Remove Q509 from its socket.
c. Adjust REF LEVEL control R572 fully counterclockwise.
d. Adjust L504, L505, and C515 (DISCR PRI control) for maximum indication on the $\mathrm{HP}-410 \mathrm{~B}$.
e. Connect the 91C to TP504.
f. Adjust R572 for an indication of 30 millivolts on the 91C.
g. Connect the HP-410B to TP502 and adjust DISCR SEC control C518 for a minimum indication on the HP-410B when set to its lowest range.
h. Replace Q509 in its socket.
i. Reset R572 as specified in paragraph 3.3.10.

### 3.3.10 REFERENCE LEVEL ADJUSTMENT.

## NOTE

The following procedure may be accomplished with the 91 C or with a Tektronix 541 oscilloscope. Steps a through $f$ describe the procedure for using the 91C and steps $g$ and $h$ describe the procedure for using the 541 oscilloscope.


PROPER ADJUSTMENT

Figure 3-1. Reference Level Adjustment, Oscilloscope Patterns
a. Connect the 91C to TP504.
b. Remove Q510 from its socket.
c. Tune L608 for maximum indication on the 91 C . Record the reading on the 91C.
d. Replace Q510 and remove Q509 from its socket.
e. Adjust R572 for the same indication recorded in step $c$.
f. Replace Q509 in its socket.
g. Connect the Tektronix oscilloscope to TP504.
h. Adjust R572 for alignment of base lines of alternate signals. See figure 3-1.
i. Set meter switch S101 on the A830-2 front panel to the 14 MC REF $B$ position. The meter should indicate in the $B$ range.

### 3.3.11 BASEBAND CANCELING ADJUSTMENT.

a. Remove Q510 from its socket.
b. Make certain that AFC DISABLE switch S102 is still jumpered.
c. Rotate R562 to its maximum counterclockwise position.
d. Connect the oscilloscope to TP503.
e. Connect the HP-200AB to baseband input jack J604.
f. Set the HP-200AB to 50 cps .
g. Set the level of the HP-200AB to produce a 2 -volt peak-to-peak waveform on the oscilloscope.
h. Adjust the oscilloscope to display the $50-\mathrm{cps}$ waveform.
i. Slowly adjust R562 to cancel the signal on the oscilloscope. Gradually increase the input signal from the HP-200AB to 0.1 volt while maintaining the null by adjustment of R562. The waveform on the oscilloscope should be less than 1 volt peak-to-peak when the input signal is 0.1 volt.
j. Replace Q510.

### 3.3.12 AFC LOOP CHECK.

a. Remove the jumper from across AFC DISABLE SWITCH S102.
b. Observe the deviation meter on the HP-335B and depress the AFC DISABLE switch. The frequency should slowly drift off and come back quickly when the AFC DISABLE switch is released.

### 3.3.13 POWER AMPLIFIER ADJUSTMENT AND TUNING.

a. Set meter switch S101 on the A830-2 to the MLXER GRID A position.
b. Adjust C 431 for maximum indication on front panel meter M101.
c. Adjust VHF OSC FREQ ADJ control C427 so that the HP-335B indicates on frequency.
d. Switch S101 to BUFFER GRID A and observe meter. It should indicate approximately 1 unit. e. Switch S101 to V428 B.
f. Adjust L429, L430, and MIX BAL control R438 for maximum indication on the front panel meter. g. Switch S101 to V429 B.
h. Adjust L431 and L432 for maximum indication on the front panel meter.
i. Remove all connections to J402 and connect the 50 -ohm load to J402.
j. Connect the HP-410B across the 50 -ohm load.
k. Switch S101 to V430C B.

1. Adjust PA PLATE control C461 for minimum indication on the front panel.
m. Adjust PA MATCH control C456 for a maximum indication on the HP-410B.
n. Adjust POWER OUT control R454 for an indication of 22.5 volts.

### 3.4 MINIMUM PERFORMANCE STANDARDS.

The A830-2 should be tested in accordance with the following procedures after alignment and adjustment. The following tests may be used to determine if the A830-2 is operating properly.

### 3.4.1 PRELIMINARY ADJUSTMENTS.

a. Connect the $\mathrm{HP}-200 \mathrm{AB}$ to J604 on the A830-2. b. Connect the HP-335B through the $10-\mathrm{db}$ pad to J402.
c. Connect the HP-330D to the modulation output of the HP-335B.

### 3.4.2 FREQUENCY RESPONSE.

a. Perform the preliminary procedures of paragraph 3.4.1.
b. Set the HP-200AB for an output of 0.100 volt on a frequency of 400 cps .
c. Adjust the HP-330D for an indication of 0 db .
d. Set the HP-200AB to 50 cps and reset output level to 0.100 volt. The HP-330D indication should be $0 \pm 0.3 \mathrm{db}$.
e. Repeat step dfor a frequency setting of $15,000 \mathrm{cps}$.

### 3.4.3 HARMONIC DISTORTION.

a. Perform the preliminary procedures of paragraph 3.4.1.
b. Set the HP-200B frequency to 50 cps and the output level to 0.01 volt. Measure the harmonic distortion on the HP-330D. It should be 1.0 percent or less.
c. Repeat step b for frequencies of 400 and $15,000 \mathrm{cps}$.

### 3.4.4 RESIDUAL FM NOISE.

a. Perform the preliminary procedures of paragraph 3.4.1.
b. Set the HP-200AB to 400 cps at an output level of 0.100 volt.
c. Measure the level across terminals 1 and 2 of the HP-335B with the HP-330D. Record the reading.
d. Turn off the HP-200AB and record the indication on the HP-330D. Record this reading.
e. Compute the $s+n / n$ ratio using the readings recorded in steps $c$ and $d$. The ratio should not be less than 60 db .

### 3.4.5 CARRIER FREQUENCY SHIFT.

a. Perform the preliminary procedures of paragraph 3.4.1.
b. Remove the audio input from J604.
c. Connect the output of the HP-200AB to terminals 3 and 4 of TB101.
d. Adjust the output of the HP-200AB to a frequency of 1000 cps and to a level sufficient to modulate the carrier 100 percent.
e. Remove the audio connections from terminals 3 and 4.
f. Adjust the $\mathrm{HP}-335 \mathrm{~B}$ to indicate 0 frequency deviation.
g. Touch the audio connections from the HP-200AB to terminals 3 and 4 of TB101 and note the carrier deviation on the HP-335B. It should be less than 500 cps .

### 3.4.6 AM NOISE MEASUREMENT.

a. Perform the preliminary procedures of paragraph 3.4.1.
b. Set the HP-335B function switch to CARRIER LEVEL and read the carrier output voltage on the modulation meter. An indication of 100 percent modulation equals 10 volts, 90 percent modulation equals 9 volts, etc.
c. Connect the 91 C to J 3 on the HP-335B and measure the noise output. Compute the carrier-toAM noise ratio using the following formula:

$$
\frac{\text { Carrier }}{\text { AM Noise }}=20 \log _{10} \frac{\text { Carrier Voltage }}{\text { AM Noise Voltage }}
$$

The ratio should not be less than 50 db .

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# SECTION IV PARTS LIST 

| ITEM | DESCRIPTION | COLLINS <br> part number | ITEM | DESCRIPTION | COLLINS <br> PART NUMBER |
| :---: | :---: | :---: | :---: | :---: | :---: |
| A830-2 10 W WIDE-BAND FM BROADCAST EXCITER 22.2714 |  |  | C433 | CAPACITOR, FIXED, MICA: same as C426 | 912-2774-00 |
|  |  |  | C434 | CAPACITOR, FIXED. MKCA: same as C426 | 912-2774-00 |
| PANEL |  |  | C435 | CAPACITOR. FIXED. CERAMIC: same as C430 | 913-1187-00 |
|  |  |  | C436 | CAPACITOR, FIXED, MICA: 10 uuf $\pm 5 \% 500$ vdew; | 912-2753-00 |
| FL101 | ATTENUATOR, FIXED: pre-emphasis network for use in FM commercial broadcast equipment; 600 ohns balanced, $w /$ center tap: $\pm 1$ to $\pm 1.5 \mathrm{db}$ frequency response | 378-0426-00 | C437 | CAPACITOR, FIXED, MICA: same as C436 | 912-2753-00 |
|  |  |  | C438 | CAPACITOR, FIXED, CERAMIC: same as C429 | 913-2981-00 |
|  |  |  | C439 | CAPACITOR, FIXED. CERAMIC: same as C430 | 913-1.187-00 |
|  |  |  | ${ }_{\text {chiru }}$ |  |  |
| FL102 |  |  | C445 | CAPACITOR, FIXED, CERAMIC: same as C429 | 913-2981-00 |
|  | FILTER, HIGH PASS: metal encased, hermetically sealed, input 600 ohms. output 600 ohms, 4 solder | 673-0869-00 | C446 | CAPACITOR, FIXED, CERAMIC. same as C430 | 913-1187-00 |
|  | type terminals, continuous duty cycle; A. D. C. part no. D10390 |  | thr |  |  |
| M101 | METER. ARBITRARY SCALE: permanent magnet moving coil d-c microammeter. 500 ua, 100 ohms resistance, 2 scales, A scale, $10-90$ ua. B scale. 175-500 ua; Assembly Products, Inc. part no. 361 | 458-0650-00 | C450 | CAPACITOR, FLXED, CERAMIC: 1000 ư -20\% | 913-1292-00 |
|  |  | 450-0650-00 |  | 180\%. 500 vdcw ; Erle Resistor part no. 327-029X5T0102Z |  |
| R101 |  |  | C451 | CAPACITOR, FIXED. CERAMIC: same as C429 | 913-2981-00 |
|  | RESISTOR, FIXED. COMPOSITION: 1000 ohms $\pm 10 \% .1 / 2 \mathrm{w}$ | 745-1352-00 | C452 | CAPACITOR, FIXED, MICA: 33 uuf $\pm 5 \% .500 \mathrm{vdcw}$; Electro Motive part no. DMI 5E330J01 | 912-2780-00 |
| R102 | RESISTOR. FIXED, FILM: 562 ohms $\pm 1 \% .1 / 4 \mathrm{w}$ | 705-7084-00 | C453 | CAPACITOR, FIXED. CERAMIC: same as C430 | 913-1187-00 |
| R103 | RESISTOR, FIXED, FILM: 261 ohnis $\pm 1$ \% ${ }_{0}$ (1/4 w | 705-7066-00 | C454 | CAPACITOR, FIXED, CERAMIC: same as C430 | 913-1187-00 |
| R104 | RESISTOR. FIXED. FILM: same as R103 | 705-7068-00 | C455 | CAPACITOR, FIXED, CERAMIC: Same as C430 | 913-1187-00 |
| S102 | $\begin{aligned} & 2 \text { section. } 2 \text { moving, } 22 \text { fixed contacts } \\ & \text { SWITCH. PUSH: spst; momentary; } 125 \mathrm{va-c} \text {. } \\ & 0.75 \mathrm{amp}, 250 \mathrm{va} \mathrm{c}, 0.25 \mathrm{amp} ; \text { Cutler-Hammer } \\ & \text { part no. } 8411-\mathrm{K} 6 \end{aligned}$ | 259-1567-00 | C456 | CAPACITOR. VARLABLE, CERAMIC: 4.5 uU $\min$ to 25 uuf max, 500 vdcw | 917-1026-00 |
|  |  | 266-6160-00 | C457 | CAPACITOR, FIXED. CERAMIC: same as C450 | 913-1292-00 |
|  |  |  | C4 | CAPACITOR, FIXE | 3-1282-00 |
| TB101 |  |  | C460 | CAPACITOR, FIXED, CERAMIC: same as C430 | 913-1187-00 |
|  | TERMINAL BOARD: barrier type w/double row front connection of 12 screw terminals; $13 / 32 \mathrm{in}$. by 7/8 in, by $5-11 / 64 \mathrm{in}$.: Howard B. Jones, Div. | 367-0518-00 | C461 | Capacitor. Varlable. air: 3.0 uuf min to 18.7 uuf max; 1250 v a-c; E. F. Johnson Co. part no. 160-110-3 | 922-0033-00 |
| TB102 | Cinch Mrg. Co, part no. 12-140-D TERMINAL BOARD: Bakelite, 4 terminals. 1 grounded. 3 insulated; 21/32 in, w by 1-1/2 in. | 306-2240-00 | C462 | CAPACITOR, FIXED, CERAMIC: same as C450 | 913-1292-00 |
|  |  | 306-2240-00 | C463 | CAPACITOR. FIXED, CERAMIC: same as C450 | 913-1292-00 |
|  | 1g: Cinch Mfg. Corp. part no. 1534-A SOCKET, ELECTRON TUBE: 8 prong actal tube socket w/ steel mtg plate, Amphenol-Borg Electronics part no. 88-8TM |  | C464 | CAPACITOR, FLXED, MICA: same as C428 | 912-2864-00 |
| XFL101 |  | 220-1005-00 | $\begin{aligned} & \text { C465 } \\ & \text { thru } \\ & \text { C468 } \\ & \text { C469 } \end{aligned}$ | CAPACITOR, FIXED, CERAMIC: same as C | 187 |
|  |  |  |  |  |  |
|  |  |  |  | NOT USED |  |
| POWER AMPLIFIER AND POWER SUPPLY |  |  |  | CAPACITOR, FIXED, CERAMIC: same as C4 | 913-1187-00 |
|  |  |  | C471 |  | 913-1187-00 |
| C401 | CAPACITOR, FIXED. CERAMIC: 1000 uU $\pm 20 \%$. 500 vdcw <br> CAPACITOR, FIXED, CERAMIC: same as C401 | 913-1186-00 | C475thru | CAPACITOR, FIXED, CERAMIC: same as C450 | 913-1282-00 |
|  |  |  |  |  |  |
| C402thru |  | 913-1186-00 | C480 |  |  |
|  |  |  | C481 | CAPACITOR, FIXED. CERAMIC: 1.0 uuf $\pm 5 \%$, 500 | 813-2977-00 |
| C408 |  |  |  | vdew; Stackpole Carbon Co. part no. |  |
| C409 | CAPACITOR, FLXED, ELECTROLYTIC: dual section, 40 uf ea section; $-10 \%+50 \%$. 450 vdcw ; Sprague Electric part no. Y27674 | 183-1259-00 | C482 | GA-1.OuutPORM5 CAPACITOR, FIXED, MICA: same as C428 | 912-2864-00 |
|  |  |  | C483 | CAPACITOR, FIXED. MICA: same as C428 | 912-2864-00 |
| C410 | CAPACITOR. FIXED, ELECTROLYTIC: 1000 uf $-10 \%$ r $100 \%$. 50 vdew | 183-1403-00 | C484 | CAPACITOR, FIXED, MICA: same as C428 | 912-2864-00 |
|  |  | 183-1403-00 | C485 | CAPACITOR, FLXED, CERAMIC: same as C450 | 913-1292-00 |
| C411 | CAPACITOR, FIXED. ELECTROLYTIC: same as C410 | 183-1403-00 | C486 | CAPACITOR, FIXED. MICA: same as C426 | 912-2750-00 |
|  |  |  | C487 | CAPACITOR, FIXED, MICA: same as C426 | 912-2750-00 |
| C412 | CAPACITOR, FIXED, ELECTROLYTIC: 500 uf $-10 \%+100 \%$ \% 50 vdew CAPACITOR. FIXED. ELECTROLYTIC: 4 Uf $-10 \%$ $+100 \%$. 50 vdcw NOT USED | 183-1575-00 | C488 | CAPACITOR, FIXED, MICA: same as C426 | 912-2750-00 |
|  |  |  | C489 | CAPACITOR, FIXED, MICA: 150 uU $55 \% .500$ vdcw; Electro Motive part no. DM15F151J01 | 912-2828-00 |
| C413 |  | 183-1389-00 | C490 | CAPACITOR, FLXED. MICA: same as C4B9 | 912-2828-00 |
| C414 |  |  | C491 | CAPACITOR, FIXED. MICA: 20 uuf $\pm 5 \%$ \% 500 vdcw ; | 812-2765-00 |
| thru |  |  | CR401 | Electro Motive part no. DM15C200J01 SEMICONDUCTOR DEVICE. DIODE: sllicon; | 353-1661-00 |
| C425 |  |  | Cr401 | SEMICONDUCTOR DEVICE. DIODE: Silicon; <br> Motorola part no. 1N1492 | 353-1661-00 |
| C426 | CAPACITOR, FIXED, MICA: 5 Uuf $55 \%, 500 \mathrm{vdew}$; Electro Motlve part no. DM15C050J01 | 912-2750-00 | Cr402 | SEMICONDUCTOR DEVICE, DIODE: same as | 353-1661-00 |
| C427 | CAPACITOR. VARIABLE, CERAMIC: 3.0 uuf min to 12.0 urf max, 350 vdew | 917-1072-00 | thru CR408 | CR401 |  |
| C428 | CAPACITOR, FIXED, MICA: 470 uuf $\pm 5 \% 300$ vdew; Electro Motive part no. DM15F471J01 | 912-2864-00 | CR409 | SEMICONDUCTOR DEVICE, DIODE: SUIICOn, single phase. half-wave; General Electrle part | 353-1526-00 |
| C429 | CAPACITOR, FIXED. CERAMIC: 1.5 UUF $\pm 5 \%$, 500 vdcw: Stackpole Carbon Co. part no. GA-1.5uufPORM5 | 913-2981-00 | CRA10 | no. IN538 <br> SEMICONDUCTOR DEVICE, DIODE: same as CR409 | 353-1526-00 |
| C430 | CAPACITOR, FLXED. CERAMIC: 4700 uuf $\pm 20 \%$. 500 vdew | 913-1187-60 | $\begin{aligned} & \text { CR411 } \\ & \text { A\& } \end{aligned}$ | SEMICONDUCTOR DEVICE. SET: two hermetically sealed sillicon voltage reference diodes; | 353-1238-00 |
| C43: | CAPACITOR, VAMLABLE, AIR; 3.0-9.8 uul. 1250 vdcw; E. F. Johnson part no. 160-211 | 922-0046-00 | CR412 | Motorola part no. 10M10ZB1 SEMICONDUCTOR DEVICE. DIODE; silicon, | 353-1208-00 |
| C432 | CAPACITOR, FIXED, CERAMIC: 7.5 u $45 \%$, 500 vdew; Slackpole Carbon Co. part no. GA-7.5uuf PORM5 | 913-2997-C0 |  | hermetically sealed; international Rect. Corp part no. 1210V01 |  |

$\angle 433$ ON64 15
$278-0326-010$

* 12 wire Sil

| ITEM | DESCRIPTION | $\begin{gathered} \text { COLLINS } \\ \text { PART NUMBER } \end{gathered}$ |
| :---: | :---: | :---: |
| CR413 | SEMICONDUCTOR DEVICE, DIODE: same as CR409 | 353-1526-00 |
| CR414 | SEMICONDUCTOR DEVICE, DIODE: same as | 353-1526-00 |
| CR415 | SEMICONDUCTOR DEVICE, DIODE: same as CR412 | 353-1208-00 |
| CR416 thru | NOT USED |  |
| CR425 |  |  |
| Cr426 | SEMICONDUCTOR DEVICE, DIODE: Allicon. hermetically sealed, diffused-junction type; Motorola part no. 1N977A | 353-3237-00 |
| F401 | FUSE, CARTRIDGE: 1.00 mp current rating, 250 v . glass body, ferrule terminals; Bussmann part no. MDL 1 | 264-4280-00 |
| F402 | FUSE. CARTRIDGE: 0.250 amp current rating, 250 v d-c, glass body, ferrule terminals | 264-4240-00 |
| J401 | JACK, TELEPHONE: steel, miniature, panel mtg; Switeheraft, Inc. part no. 3501FP | 360-0148-00 |
| J402 | CONNECTOR, RECEPTACLE, ELECTRICAL: single round female contact, right angle shape; Amplienol part no. 31-213 | 357-8258-00 |
| LA01 | REACTOR: 7.2 henrys min. $0.300 \mathrm{amp} \mathrm{d}-\mathrm{c} ; 60$ ohms; 4-37/64 in. by 5-5/16 in. overall; Stancor Elec. Inc. part no. RS-8300 | 668-0015-00 |
| L402 | NOT USED |  |
| $\begin{aligned} & \text { thru } \\ & \text { LA25 } \end{aligned}$ |  |  |
| L426 | COIL, RADIO FREQUENCY: $0.68 \mathrm{uh} \pm 3 \%, 250 \mathrm{mc}$. $0.12 \mathrm{ohm}, 1750 \mathrm{ma} ; 3 / 16 \mathrm{in}$. dia by $7 / 16 \mathrm{in}$. 1 g ; Delevon part no. 1840 | 240-1844-00 |
| L427 | COIL, RADIO FREQUENCY: $0.25 \mathrm{uh} \pm 3 \% .400 \mathrm{mc}$ $0.04 \mathrm{ohm}, 2850 \mathrm{ma} ; 3 / 16 \mathrm{in}$. dia by $7 / 16 \mathrm{ln} . \mathrm{lg}$ | 240-1843-00 |
| LA28 | COIL, RADIO FREQUENCY, NO. 1: single layer wound $144 \mathrm{wire}, 1 / 2 \mathrm{in}$. ID of coll, $7 / 8 \mathrm{in}$. Ig overall | 548-1605-003 |
| L429 | COIL. RADIO FREQUENCY: variable; 88 to 108 $\mathrm{mc},-15^{\circ} \mathrm{C}$ to $+55^{\circ} \mathrm{C}$ temp range; $850 \mathrm{v} \mathrm{d-c}$ dielectric strength | 278-0730-00 |
| L430 | COIL, Radio Frequency: same as l429 | 278-0730-00 |
| thru |  |  |
| L435 | COIL, RADIO FREQUENCY, NO. 2: single layer wound in 16 wire; $3 / 4 \mathrm{in}$. 1 D of coll. $2-7 / 8 \mathrm{in} .1 \mathrm{~g}$ overall | 549-1606-003 |
| 4436 | COIL, RADIO FREQUENCY: single layer wound; 5.6 uh. 860 ma current. 0.95 ohm; Jeffers Electronics Div. of Speer Carbon Co. part no. 10402-34 | 240-0179-00 |
| L437 | COIL. RADIO FREQUENCY: same as La36 | 240-0179-00 |
| L438 | COIL, RADIO FREQUENCY: single layer wound, 0.47 uh nom inductance. 0.09 ohm max de resistance, 1600 ma max current rating, Jeffers Electronics, Div. of Speer Carbon Co. part no. 10100-126 | 240-0060-00 |
| L439 | COIL, RADIO FREQUENCY, NO. 3: single layer wound 114 wire, $3 / 4 \mathrm{in}$. ID of coll, $1-3 / 8 \mathrm{in}$. h ; approx 1-11/16 in. gg overall | 549-1607-003 |
| L440 | COIL, RADIO FREQUENCY: 1.00 uh $\pm 10 \%, 0.30$ ohm d-c reslstance; $850 \mathrm{ma} \mathrm{d}-\mathrm{c}_{\text {; }}$ Jeffers Electronics part no. 10100-128 | 240-0062-00 |
| L441 | COIL, RADIO FREQUENCY, NO. 4: single layer wound 18 wire, $3 / 18 \mathrm{in}$. ID of coll, $5 / 16 \mathrm{in}$. ig | 553-5671-002 |
| R401 | RESISTOR, FLXED, WIREIVOUND: 100 ohms $\pm 10 \frac{\pi}{6}, 10 \mathrm{w}$ | 710-9053-00 |
| R402 | RESISTOR, FDED, WIREWOUND: 16,000 ohms $\pm 5 \%$, 25 w | 710-0369-00 |
| $\mathrm{R}^{\mathrm{R}} \mathbf{4} \mathbf{3}$ | NOT USED |  |
| R404 | RESISTOR, FDKED, WIREWOUND: 600 ohms $\pm 10 \%$, 10 w | 710-9081-00 |
| R405 | RESISTOR, FLXED, WIREWOUND: 12,000 ohms $\pm 10 \%, 10 w$ | 710-8070-00 |
| R406 | RESISTOR, FTXED, WIREWOUND: 25,000 ohms $\pm 10 \%, 10 w$ | 710-8068-00 |
| R407 | RESISTOR, FLXED, WIREWOUND: 5.0 ohms $\pm 10 \%$, 5 w | 710-9105-00 |
| R40日 | RESISTOR, FIXED, WIREWOUND: 25 ohms $\pm 10 \%$, 7 w | 710-9019-00 |
| R409 | RESISTOR, FIXED, WIREWOUND: same as R4C8 | 710-9019-00 |
| R410 | RESISTOR, FDED, COMPOSITION: 160.0 ohms $\pm 5 \%, 5 \mathrm{w}$ | 747-5444-00 |
| R411 | RESISTOR, FLXED, COMPOSITION: same as 4 d07 | 710-9105-00 |
| R412 | RESISTOR, FIXED, COMPOSITION: game as R410 | 747-5444-00 |
| R413 | RESISTOR, FLXED, WIREWOUND: 100 ohms $\pm 10 \%, 7$ w | 710-9005-00 |
| $\begin{aligned} & \text { R414 } \\ & \text { thru } \\ & \text { R425 } \end{aligned}$ | NOT USED |  |


| TTEM | DESCRIPTION | COLLINS PART NUMBER |
| :---: | :---: | :---: |
| R426 | RESISTOR, FIXED, COMPOSITION: $10,000 \mathrm{ohms}$ $\pm 10 \%, 1 / 2$ w | 745-1394-00 |
| R427 | RESISTOR, FIXED, COMPOSITION: 1000 ohms $\pm 10 \%, 1 / 2$ w | 745-1352-00 |
| H428 | RESISTOR, FIXED, COMPOSITION: 220 ohms $\pm 10 \%$. $1 / 2 \mathrm{w}$ | 745-1324-00 |
| R428 | RESISTOR, FIXED, COMPOSITION: 2700 ohms $\pm 10 \%, 1 / 2$ w | 745-1370-00 |
| R430 | RESISTOR, FIXED, COMPOSITION: 47,000 ohms $\pm 10 \%, 1 / 2$ w | 745-1422-00 |
| R431 | RESISTOR, FIXED, COMPOSITION: 1500 ohms $\pm 10 \%, 1 / 2 \mathrm{w}$ | 745-1359-00 |
| R432 | RESISTOR, FIXED, COMPOSITION; 39,000 ohms $\pm 10 \%$. $1 / 2 \mathrm{w}$ | 745-1419-00 |
| R433 | RESISTOR, FLXED, COMPOSITION: 2200 ohme $\pm 10 \%, 1 / 2$ w | 745-1366-00 |
| R434 | RESISTOR, FIXED, COMPOSITION: 0.10 megohm $\pm 10 \%, 1 / 2 \mathrm{w}$ | 745-1436-00 |
| R435 | RESISTOR, FIXED, COMPOSITION: same as R434 | 745-1436-00 |
| R436 | RESISTOR, FLXED, COMPOSITION: same as R431 | 745-1359-00 |
| R437 | RESISTOR, FLXED, COMPOSITION: same as R433 | 745-1366-00 |
| R438 | RESISTOR. VARIABLE: COMPOSITION; 500 ohms $220 \%$. 0.2 w | 376-0202-00 |
| R438 | RESISTOR, FIXED, COMPOSITION: 3300 ohms $\pm 10 \% .2$ w | 745-5673-00 |
| R440 | RESISTOR, FIXED, COMPOSITION: game as R439 | 745-5673-00 |
| R441 | RESISTOR, FLXED, COMPOSITION: same as R434 | 745-1436-00 |
| R442 | RESISTOR. FLEED, COMPOSITION: same as R426 | 745-1394-00 |
| R443 | RESISTOR, FIXED, COMPOSITION: 68 ohms $\pm 10 \%, 1 / 2$ w | 745-1303-00 |
| 7844 | RESISTOR, FIXED, COMPOSITION: 39,000 chms $\pm 10 \%, 1 \mathrm{w}$ | 745-3418-00 |
| R445 | RESISTOR, FIXED, COMPOSITION: 4700 ohms $\pm 10 \%$, 1 w | 745-3380-00 |
| R446 | RESISTOR. FIXED. COMPOSITION: same as R434 | 745-1436-00 |
| R447 | RESISTOR, FIXED, COMPOSITION: same as R426 | 745-1394-00 |
| R448 | RESISTOR, FIXED, COMPOSTION: 270 ohms $\pm 10 \%, 1 \mathrm{w}$ | 745-3328-00 |
| R449 | RESISTOR, FIXED, COMPOSITION: 10,000 ohms $\pm 10 \%, 1 \mathrm{w}$ | 745-3384-00 |
| R450 | RESISTOR, FLXED, COMPOSITION: 820 ohms $\pm 10 \% .2 \mathrm{w}$ | 745-5649-00 |
| R451 | RESISTOR, FIKED. COMPOSITION: 10 ohms $\pm 100,1 / 2 \mathrm{w}$ | 745-1288-00 |
| R452 | RESISTOR, FIXED, COMPOSITION: 3300 ohms $\pm 10 \%, 1 / 2 \mathrm{w}$ | 745-1373-00 |
| R453 | RESISTOR, FIXED, COMPOSITION: same as R432 | 745-1419-00 |
| R454 | RESISTOR, VARLABLE, WIREWOUND: 250 ohms | 377-0621-00 |
| R455 | RESISTOR, FIXED, COMPOSITION: 180 ohms $\pm 10 \mathrm{G}, 2 \mathrm{w}$ | 745-5621-00 |
| R456 | RESISTOR, FIXED, COMPOSITION: 8200 ohms $\pm 10 \%, 1 / 2 \mathrm{w}$ | 745-1391-00 |
| R457 | RESISTOR, FLXED, COMPOSITION: 33 ohms $\pm 10 \% 1 / 2 \mathrm{w}$ | 745-1299-00 |
| R459 | RESISTOR, FIXED, COMPOSITION: 100 ohms $\pm 10 \%, 1 / 2 \omega$ | 745-1310-00 |
| R459 | RESLSTOR, FLXED, COMPOSITION: same as R451 | 745-1268-00 |
| R460 | NOT USED |  |
| R461 | RESISTOR. FIXED. FILM: 51,000 chms $\pm 10 \%$, 5 w | 714-2973-00 |
| R462 | RESLSTOR, FLXED, COMPOSITION: same as R443 | 745-1303-00 |
| R463 | RESISTOR, FLXED, COMPOSITION: 22.000 ohms $\pm 10 \%, 1 / 2 \mathrm{w}$ | 745-1408-00 |
| R464 | RESISTOR, FIXED, COMPOSITION: same as R458 | 745-1310-00 |
| R465 | RESISTOR, FLXED, COMPOSITION: 27.000 ohms $\pm 10 \%, 1 / 2$ w | 745-1412-00 |
| R466 | RESISTOR, FIXED, WIREWOUND: 20,000 ohms $\pm 10 \%, 10 \mathrm{w}$ | 710-9067-00 |
| R487 | RESISTOR, FIXED, COMPOSITION: $\mathbf{2 2 , 0 0 0}$ ohms $\pm 10 \%$. 2 w | 745-5708-00 |
| S401 | SWITCH, TOGGLE: dpst; 125 v a-c, $15 \mathrm{amp}, 250$ v a-c, 10 amp ; Cutler-Hammer. Inc. part no. 7561 K 4 | 266-0099-00 |
| T401 | TRANSFORMER, POWER. STEP-UP, STEP-DOWN: <br>  $50 / 60 \mathrm{cps}$; continuous duty cycle; Stancor Electric part no. 31215 | 662-0046-00 |
| T402 | TRANSFORMER, POWER, STEP-DOWN: pri 120 v rms; sec. $11,77 \mathrm{v}, \mathrm{ct}$; sec. it2. 41.5 ct ; 50/60 cps; continuous duty cycle; Stancor Electric part no. 31214 | 662-0048-00 |
| T403 thry T425 | NOT USED |  |


| ITEM | DESCRIPTION | COLLINS PART NUMBER |
| :---: | :---: | :---: |
| T428 | TRANSFORMER，RADIO FREQUENCY：pri 14 turns 户 26 wire，close wound：sec． 13 turns 26 wire，close wound | 549－1580－00 |
| TB401 | TERMINAL BOARD：phenolic $w / 3$ solder－lug terminals； $11 / 16 \mathrm{in}$, w by $1-1 / 8 \mathrm{in}$ ．Ig；Cinch Mg．Corp．part no．1520－A | 306－9033－00 |
| TB402 | TERMINAL BOARD：Baklite， 2 terminals； $21 / 32$ In．by $3 / 4 \mathrm{in} . \mathrm{lg}_{1}$ Cinch Mrg．Co．part no．1513－A | 306－2220－00 |
| TB403 | TERMINAL BOARD：laminated phenolic w／4 solder－lug terminals；27／32 in．w by 1－1／2 in．Ig overall；Cinclı Mfg．Co．part no． 1909 | 308－0638－00 |
| TB404 | TERMINAL BOARD：same as TB401 | 306－9033－00 |
| TB405 | TERMINAL BOARD：phenolle；steel mounting base， brass lugs， 12 terminals，H．B．Jones part no． 2012 | 367－0905－00 |
| TB406 | TERMINAL BOARD：same as TB405 | 367－0905－00 |
| TBA07 thru | NOT USED |  |
| TB425 |  |  |
| TB426 | TERMINAL BOARD：phenolic， 4 brass solder－lug terminals； $\mathrm{I} / 16 \mathrm{in}$ ，by $3 / 8 \mathrm{in}$ ．by $1-1 / 2 \mathrm{in}$ ．； Cinch Mig．Corp．part no． 1532 －A | 306－9032－00 |
| TE427 | TERMINAL BOARD；same as TB403 | 306－0838－00 |
| TB428 | TERMINAL BOARD：phenollc， 5 brass solder－lug terminals； $1 / 16 \mathrm{in}$ ．by $3 / 8 \mathrm{in}$ ．by $1-7 / 8 \mathrm{in}$ ．； Cinch M1．g．Corp．part no．1542－A－FV | 306－0951－00 |
| TB429 | TERMINAL BOARD：same as TB428 | 306－0951－00 |
| TB430 | TERMINAL BOARD：samo as TB－402 | 306－2220－00 |
| TBA31 | TERMINAL BOARD：same as TB428 | 306－0951－00 |
| V401 | NOT USED |  |
| thru |  |  |
| V425 |  |  |
| V426 | of America part no．6UBA | 255－0328－00 |
| V427 | ELECTRON TUBE：glass envelope；twin triode； Radlo Corp．of America part no．12AT7 | 255－0205－00 |
| V428 | ELECTRON TUBE：pentode，Radio Corp．of America part no．6AUG | 255－0202－00 |
| V429 | ELECTRON TUBE ；glass envelope；whf beam power；Radio Corp．of America part no． 5763 | 257－0059－00 |
| V430 | ELECTRON TUBE：glass envelope，Radio Corp． of America part no．2E26 | 256－0084－00 |
| XF401 | FUSE HOLDER：extractor post type，for use $w /$ 3 AG fuses；0－20 amp，100－125 v；clear knob； neon lamp type | 265－1072－00 |
| XF402 | FUSE HOLDER：same as XF401 | 265－1072－00 |
| XV401 thru | NOT USED |  |
| XV425 |  |  |
| XV426 | SOCKET，ELECTRON TUBE： 9 contact minfature； copper nonmagnetic alloy contacts；phenolic insulation；Sylvania Electric Products，inc． part no．7490－0100 | 220－1244－00 |
| XV427 | SOCKET，ELECTRON TUBE：same as XV426 | 220－1244－00 |
| XV428 | SOCKET，ELECTRON TUBE： 7 contact miniature for uhf application；phenolic insulation；Sylvania Electric Products，Jnc．part no．7470－0125 | 220－1203－00 |
| XV430 | SOCKET，ELECTRON TUBE： 8 prong octal tube socket w／steel mitg plate；Amphenol－Borg Electronics part no．88－8TM | $220-1005-00$ |
| AUTOMATIC FREQUENCY CONTROL |  |  |
| C501 | CAPACITOR．FIXED．CERAMIC： 1000 uUI $\pm 20 \%$ ． 500 vdcw | 913－1186－00 |
| C502 | CAPACITOR，FIXED，CERAMIC：same as C501 | 013－1186－00 |
| C503 | CAPACITOR，FIXED，CERAMIC：same as C501 | 913－1186－00 |
| C504 | CAPACITOR，FIXED，CERAMIC： $0.01 \mathrm{u}-0 \%$ $+100 \%$ temp range； 100 vdew；Erie Resistor Corp． part no．855－502－X550－103P | 813－3680－00 |
| C505 | CAPACITOR，FIXED，CERAMIC：same as C504 | 913－3680－00 |
| C50日 | CAPACITOR．FIXED，CERAMIC：same as C504 | 813－3680－00 |
| C507 | CAPACITOR，FLXED，CERAMIC：same as C504 | 913－3680－00 |
| C508 | CAPACITOR，FLXED，MICA： 10 uuf $\pm 5 \%, 500 \mathrm{vdc} w$ ； Electro Motive part no．DM15C100J01 | 912－2753－00 |
| C509 | CAPACITOR，FIXED，CERAMIC：same as C504 | 913－3680－00 |
| C510 | CAPACITOR，FIXED，MICA： 82 uul $\pm 5 \%, 500 \mathrm{vdew}$ ； Electro Motive part no．DM15E820．J01 | 912－2810－00 |
| C511 | CAPACITOR，FIXED，CERAMIC： $0.1 \mathrm{uf}-20 \%$ $+80 \%$ ． 50 vdew；Sprague Electric part no．33C41 | 913－3886－00 |
| C512 | CAPACITOR，FIXED，CERAMIC；same as C504 | 913－3680－00 |
| C 513 | CAPACITOR，FLXED．CERAMIC：same as C504 | 819－3680－00 |
| C514 | CAPACITOR．FIXED．MICA： 100 uUf $\pm 5 \%$ 500 wvde | 812－2816－00 |


| ITEM | DESCRIPTION | COLLINS PART NUMBER |
| :---: | :---: | :---: |
| C515 | CAPACITOR，VARIABLE，CERAMIC： 5.0 uul min to 37.5 uuf max， 350 vdcw；Eric Resistor part no． 557018COP039R | 917－1073－00 |
| C516 | CAPACITOR，FIXED，MICA： 220 UUK $\mathbf{~ 5 \%} \% 500$ vdcw，Electro Motive part no．DM15F221J01 | 912－2840－00 |
| C517 | CAPACITOR．FIXED．MICA： 30 uuf $\pm 2 \% .500$ vdcw； Electro Motive part no．DM15E300G01 | 912－2776－00 |
| C518 | CAPACITOR，VARIABLE，CERAMIC： 3.0 uff min to 12.0 uuf max， 350 vdcw | 917－1072－00 |
| C518 | CAPACITOR，FIXED．MICA： 470 UUF $\pm 5 \%, 300$ vdew；Electro Motive part no．DM15F471J01 | 912－2864－00 |
| C520 | CAPACITOR．FIXED，MICA．same as C519 | 912－2864－00 |
| C521 | CAPACITOR，FIXED，ELECTROLYTIC： 100 uf $-10 \%+100 \% .10 \mathrm{vdcw}$ ；Sprague Electric part no． S13891 | 183－2151－00 |
| C522 | CAPACITOR，FIXED，ELECTROLYTIC： 100 uf $-10^{\circ},+100^{\circ} \%, 25$ vdew；Sprague Electric part no． 30D188A1 | 183－1192－00 |
| C523 | CAPACITOR，FIXED，CERAMIC： 0.68 uf $-20 \%$ $\div 80$ 符， 25 ydcw；Sprague Electric part no．5C12A | 913－3809－00 |
| C524 | CAPACITOR，FIXED，ELECTROLYTIC：same as C522 | 183－1192－00 |
| C525 | CAPACITOR．FLXED．CERAMIC：same as C523 | 913－3808－00 |
| C526 | CAPACITOR，FIXED，ELECTROLYTIC：same as C522 | 183－1192－00 |
| C 527 | CAPACITOR，FLXED，CERAMIC：same as C523 | 913－3809－00 |
| C528 | CAPACITOR，FIXED，ELECTROLYTIC：same as C522 | 183－1182－00 |
| C52日 | CAPACITOR，FIXED，PAPER： 5.0 uf $\pm 20 \% .150$ vdew；Sprague Electric part no．121P50501R5S2 | 931－2585－00 |
| C530 | CAPACITOR，FDED，PAPER：Spme as C529 | 931－2585－00 |
| C531 | CAPACITOR，FIXED，PAPER： $2.0 \mathrm{ul} \pm 20 \% 1200$ vdew；Aerovox Corp．part no．P82922N14 | 951－0670－00 |
| C532 | CAPACITOR，FIXED，PAPER： 20 uf $\pm 20 \%, 150$ vdew；Sprague Electric part no．143P101M | 951－2004－00 |
| C533 | CAPACITOR，FLXED，PAPER：same as C531 | 951－0670－00 |
| C534 | CAPACITOR，FIXED．ELECTROLYTIC： 250 uf $-10 \%+100 \%, 30$ valew | 183－1565－00 |
| C535 | CAPACITOR，FIXED，ELECTROLYTIC： 1000 uf $-10 \%+100 \%$ ． 50 vdcw | 183－1403－00 |
| ${ }^{C} 536$ | CAPACITOR，FLXED，ELECTROLYTIC：same as C535 | 183－1403－00 |
| C537 | CAPACITOR，FIXED，PAPER： 35 uf $\pm 20 \%, 150$ vdcw：Sprague Electric part no．143P4M | 851－2003－00 |
| C538 | CAPACITOR，FIXED，PAPER：same as C537 | 951－2003－00 |
| C539 | CAPACITOR，FLXED，ELECTROLYTIC： 250 u $-10 \%+100 \% 12 \mathrm{vdcw}$ ；Sjrague Electric Co．part no．30D157A1 | 183－1190－00 |
| C540 | CAPACITOR，FIXED，CERAMIC：same as C5Il | 913－3886－00 |
| C541 | CAPACITOR，FIXED，MICA： 180 ưf $\pm 5 \%, 500$ vdew，Electro Motive part no．DM15F181J01 | 912－2834－00 |
| C542 | CAPACITOR，FIXED．CERAMIC：same as C511 | 913－3886－00 |
| C543 | NOT USED |  |
| C544 | CAPACITOR，FIXED，MICA： 68 uui $\pm 5 \%, 500$ vdcw；Electro Motive part no．DM15E680J01 | 912－2804－00 |
| C545 | CAPACITOR，FIXED，MICA： 510 unf $\pm 5 \%, 300$ vdcw；Electro Motive part no．DM15F511J01 | 912－2867－00 |
| C54日 | CAPACITOR．FIXED，CERAMIC：same as C501 | 913－1186－00 |
| C547 | CAPACITOR．FIXED，CERAMIC：same as C501 | 913－1186－00 |
| C548 | CAPACITOR．FLXED，CERAMIC：Same as C501 | 913－1186－00 |
| C549 | CAPACITOR，FIXED，CERAMIC： 3300 uU $\pm 20$ 带． 500 vdcw | 913－1193－00 |
| C550 | CAPACITOR．FIXED，MICA： 22 uuf $\pm 5 \% .500$ vdew，Electro Motive part no．DM15C220J01 | 912－2768－00 |
| C551 | CAPACITOR，FLXED，ELECTROLYTIC：same as C534 | 183－1565－00 |
| C552 | CAPACITOR．FLXED，CERAMIC：same as C54日 | 日13－1193－00 |
| C553 | CAPACITOR，FIXED，CERAMIC：game as C501 | 913－1186－00 |
| CR501 | SEMICONDUCTOR DEVICE，DIODE：germanium； Transitron part no．1N270 | 353－2018－00 |
| CR502 thru | SEMICONDUCTOR DEVICE．DIODE：same as CR501 | 353－2018－00 |
| CR506 |  |  |
| CR507 | SEMICONDUCTOR DEVICE，DIODE：germanium； Erie Resistor part no．1N198 | 353－0160－00 |
| CR508 | SEMICONDUCTOR DEVICE，DIODE：same as CR507 | 353－0160－00 |
| CR509 | SEMICONDUCTOR DEVICE，SET：four matched sllicon dlodes；encapsulated；Fairchild Semicon－ ductor Corp．part no．FA－4000 | 353－3271－00 |
| CR510 | SEMICONDUC TOR DEVICE，SET：same as CR50日 | 353－3271－00 |
| CR511 | SEMICONDUCTOR DEVICE，SET：same as Criog | 353－3271－00 |
| $\begin{aligned} & \text { CR5 } 12 \\ & \text { CR513 } \end{aligned}$ | NOT USED <br> SEMICONDUCTOR DEVICE，DIODE：snme as CR507 | 353－0160－00 |


| ITEM | DESCRIPTION | COLLINS PART NUMBER | ITEM | DESCRIPTION | COLLINS PART NUMBER |
| :---: | :---: | :---: | :---: | :---: | :---: |
| CR514 | SEMICONDUCTOR DEVICE，DIODE：hermetically sealed，silicon：Motorola，Inc．part no．1N718 | 353－2734－00 | R525 | SISTOR，FIXED．FILM： 7500 ohms $\pm 1 \%$ | 705－7138－00 |
|  |  |  |  | RESISTOR，FIXED，FILM： 422 ohms $\pm 19,1 / 4 \mathrm{w}$ | 705－7078－00 |
| CR515 | SEMICONDUCTOR DEVICE，DIODE：quick recovery silicon Junction diode；Hughes Atrcraft part no．1N626 | 353－2857－00 | R527 | RESISTOR，FIXED．FILM： $196,000 \mathrm{ohms} \pm 1$ \％， | 705－7206－00 |
|  |  |  | R528 | RESISTOR，FIXED，FILM：same as R524 | 705－7152－00 |
| CR516 | SEMICONDUCTOR DEVICE．DIODE：same as CR515 | 353－2857－00 | R529 | RESISTOR，FIXED，FILM：same as R525 | 705－7138－00 |
|  |  |  | R530 | RESISTOR，FIXED，FILM：same as R526 | 705－7078－00 |
| J501 | JACK，TIP：insulated tip u／w standard 0.080 in ． test probes；browi；E．F．Johnson Co．part no． 105－208－200 | 360－0152－00 | R531 | RESISTOR，FIXED，FILM：same as R527 | 705－7206－00 |
|  |  |  | R532 | RESISTOR，FIXED，FILM：same as R524 | 705－7152－00 |
|  |  |  | R533 | RESISTOR，FLXED，FILM：same as R525 | 705－7138－00 |
| J502 | JACK，TIP：insulated tip u／w standard 0.080 in ． test probes；red；E．F．Johnson Co．part no． 105－202－200 | 360－0150－00 | R534 | RESLSTOR，FIXED，FILM：same as R526 | 705－7078－00 |
|  |  |  | R535 | RESISTOR．FIXED，FILM： 38.300 ohms $\pm 1 \%$ ， $1 / 4 w$ | 705－7172－00 |
| 5503 | JACK，TIP：insulated $\mathrm{tip} \mathrm{u} / \mathrm{w}$ standard 0.080 in ． test probes；orange：E．F．Johnson Co．part no．$105-206-200$ | 360－0154－00 | R536 | RESISTOR，FIXED，FILM： 19,600 ohms $\pm 1 \%$ ． 1／4 w | 705－7158－00 |
|  |  |  | R537 | RESISTOR，FLXED，FILM： 1470 ohmis $41 \%$ ，1／4 | 705－7104－00 |
| J504 | JACK，TIP insulated tip $\mathbf{u} / \mathrm{w}$ standard 0.080 in ． test probes；yellow；E．F．Johnson Co．part no．105-207-200 | 360－0156－00 | R538 | RESISTOR，FIXED，FILM；same as R537 | 705－7104－00 |
|  |  |  | $R 539$ | RESISTOR．FIXED，FILM： 2870 ohms $\pm 1 \% 1 / 4$ | 705－7118－00 |
|  |  |  | R540 | RESISTOR．FIXED，FILM：same as R539 | 705－7118－00 |
| L501 | COIL，RADIO FREQUENCY：single layer wound． 100 uh nom inductance， 3.2 ohms d－c resistance， 530 ma current rating；Jeffers Electronics，Div． of Speer Carbon Co．part no．10404－34 | 240－0193－00 | R541 | RESISTOR，FIXED，FILM： 100,000 ohms $\pm 1$ g， $1 / 4 w$ | 705－7192－00 |
|  |  |  | R542 | RESISTOR．FLXED，FILM：same as R541 | 705－7192－00 |
|  |  |  | R543 | RESISTOR．FLXED，COMPOSITION： 0.12 megohm | 745－1440－00 |
| L502 | COIL，RADIO FREQUENCY：single layer wound， 3.30 uh nom inductance， $0.15 \mathrm{ohm} \mathrm{d}-\mathrm{c}$ resistance， 1150 ma current rating；Jeffers Electronics．Div． of Speer Carbon Co．part no．10102－110 | 240－0065－00 |  | $\pm 10 \% 1 / 2 \mathrm{w}$ |  |
|  |  |  | R544 | RESISTOR，FIXED，COMPOSITION： 27,000 ohms $\pm 10^{C}, 1 / 2 \mathrm{w}$ | 745－1412－00 |
|  |  | 240－0145－00 | R545 | RESISTOR，FIXED，COMPOSITION： 0.18 megohm | 745－1447－00 |
| L503 | COIL，RADIO FREQUENCY：single layer mound， 4.7 uh inductance； 0.22 ohm max d－c resistance， 950 ma current rating；Jeffers Electronlcs．Div． of Speer Carbon Co．part no．10102－115 |  | R546 | RESISTOR，FIXED，FILM： 5620 ohms $=1 \%$ ， $1 / 4 \mathrm{w}$ | 705－7132－00 |
|  |  |  | R5 | RESISTOR，FIXED．FILM： $9090 \mathrm{ohms} \pm 1 \%, 1 / 4 \mathrm{w}$ | 705－7142－00 |
|  |  |  | R5 | RESISTOR，FIXED，FILM：same as R547 | 705－7142－00 |
| L504 |  | 278－0733－00 | R5 | RESISTOR，FIXED，FILM： 8250 ohms $\pm 1 \% .1 / 4 \mathrm{w}$ | 705－7140－00 |
|  |  |  | R550 | RESISTOR，FIXED，FILM： 1330 ohms $\pm 1 \%$ ， $1 / 4 \mathrm{w}$ | 705－7102－00 |
| L505 | $+55^{\circ} \mathrm{C}$ temp range； 100 v d－c dielectric strength COIL．RADIO FREQUENCY：same as L504 | 278－0733－00 | R551 | RESISTOR，FIXED，COMPOSITION： 15,000 ohms | 745－1401－00 |
| Q501 | TRANSISTOR：germanium，RCA part no．2N1225 | 352－0135－00 |  | $\pm 10 \% 1 / 2 \mathrm{w}$ |  |
|  | TRANSISTOR：same as Q501 | 352－0135－00 | R552 | RESISTOR．FIXED，COMPOSITION． 3300 ohms | 745－1373－00 |
| Q503 | TRANSISTOR：hermelically sealed，NPN sillicon； Fairchild Semi Conductor Co．part no，2N708 | 352－0322－00 | R553 | $\pm 10$ ¢0． $1 / 2 \mathrm{w}$ RESISTOR，FIXED，COMPOSITION： 1000 ohms | 745－1352－c0 |
| Q504 | TRANSISTOR：hermetically sealed，NPN diffued siltcon planar translstor；Fairchild Seniconductor Corp．part no．2N1613 | 352－0349－00 |  | $\pm 10 \mathrm{C}, 1 / 2 \mathrm{w}$ |  |
|  |  |  | R554 | RESISTOR．FLXED，COMPOSITION：same as R551 | 745－1401－00 |
|  |  |  | R555 | RESISTOR．FIXED，FILM：75，000 ohms $\pm 1 \%$ ， | 705－7186－00 |
| Q505 | Corp．part no．2N1613 <br> TRANSISTOR：same as Q504 | 352－0349－00 |  | 1／4 |  |
| Q506 | TRANSISTOR：same as Q504 | 352－0349－00 | R556 | RESISTOR．FIXED，COMPOSITION： 10 ohms | 745－1268－00 |
| Q507 | TRANSISTOR：same as Q504 | 352－0349－00 |  | $\pm 10 \%$ ． $1 / 2 \mathrm{w}$ |  |
| Q508 | TRANSLSTOR：silicon；General Electric part no． 2N491 | 352－0116－00 | R557 | RESISTOR，FLXED，COMPOSITION： 220 ohms $\pm 10 \% 1 / 2 \mathrm{w}$ | 745－1324－00 |
| Q509 | TRANSISTOR：gernmanium；hermetically sealed； Sylvania Electric part no．2N1605 | 352－0348－00 | R558 | RESISTOR．FLXED，FILM： 56.200 ohms $\pm 1 \%$ ． 1／4 w | 705－7180－00 |
| Q510 | TRANSISTOR：same as Q509 | 352－0348－00 | R559 | RESISTOR，FLXED，FILM： 2610 ohms $\pm 1 \% 1 / 4 \mathrm{w}$ | 705－7116－00 |
| Q511 | TRANSISTOR：hermetically sealed；PNP germanium；General Electric part no．2N1175A | 352－0315－00 | R560 | RESISTOR，FIXED，FILM． 3160 ohms $\pm 1 \%, 1 / 4 \mathrm{w}$ | 705－7120－00 |
|  |  |  | R561 | RESISTOR．FIXED，COMPOSITION：same as R553 | 745－1352－00 |
| $\begin{aligned} & \text { Q512 } \\ & \text { R50! } \end{aligned}$ |  | 352－0135－00 | R562 | RESISTOR．VARIABLE：COMPOSITION； 1000 | 376－4727－00 |
|  | RESISTOR，FIXED，COMPOSITION： 68 ohns $\pm 10$ 最 $1 / 2 \mathrm{w}$ | 745－1303－00 | R5 | olims $\pm 20 \% .1 / 4 \mathrm{w}$ <br> RESISTOR．FIXED．COMPOSITION：same as R506 | 745－1394－00 |
| R502 | RESISTOR，FIXED，COMPOSITION： 2700 olms $\pm 10$ 名， $1 / 2 \mathrm{w}$ | 745－1370－00 | R5 | RESISTOR，FLXED，FILM： 3480 olims $\pm 1 \% .1 / 4 \mathrm{w}$ | 705－7122－00 |
|  |  |  | R565 | RESLSTOR，FLXED，FILM： 4640 ohms $\pm 1$ \％， $1 / 4 \mathrm{w}$ | 705－7128－00 |
| R503 | RESISTOR，FIXED，COMPOSITION：same as R502 | 745－1370－00 | R566 | RESISTOR，FIXED，FILM：same as R521 | 705－7170－00 |
| R504 | RESISTOR．FIXED，COMPOSITION： 680 ohms $\pm 10^{\circ} \mathrm{o}$ ． $1 / 2 \mathrm{w}$ | 745－1345－00 | R567 | RESISTOR，FIXED．FILM：same as R521 | 705－7170－00 |
|  |  |  | R568 | RESISTOR，FLXED，COMPOSITION： 100 ôms | 745－1310－00 |
| R505 | RESISTOR，FIXED，COMPOSITION： 4700 ohms $\pm 10 \%$ ． $1 / 2 \mathrm{w}$ | 745－1380－00 | R569 | $\pm 109,1 / 2 \mathrm{w}$ RESISTOR，FIXED，COMPOSITION： 6800 ohms | 745－1387－00 |
| R506 | RESISTOR，FIXED，COMPOSITION： 16.000 ohms $\pm 10$ \％ $1 / 2 \mathrm{w}$ | 745－1394－00 | R570 | $\pm 10 \% .1 / 2 \mathrm{w}$ RESISTOR，FIXED，COMPOSTION： 8200 ohms | 745－1381－00 |
| R507 | RESISTOR，FIXED，COMPOSITION：same as R505 | 745－1380－00 |  | $\pm 10 \% .1 / 2 \mathrm{w}$ |  |
| R508R509 | RESISTOR，FIXED．FILM： 42.2 ohms $\pm 1 \% 1 / 4 \mathrm{w}$ | 705－7030－00 | R571 | RESISTOR，FIXED．COMPOSITION：10，000 ohms | 745－1404－00 |
|  | RESISTOR，FIXED，FILM： 51.1 ohms $\pm 1 \%$ ， $1 / 4 \mathrm{w}$ | 705－7034－00 |  | ＋5\％\％1／2 ${ }^{\text {w }}$ |  |
| $\begin{aligned} & \text { R509 } \\ & \text { R510 } \end{aligned}$ | RESISTOR，FIXED，COMPOSITION：same as R506 | 745－1304－00 | R572 | RESISTOR．VARIABLE：COMPOSITION； 500 | 376－4726－00 |
| R511 | RESISTOR，FIXED．COMPOSITION：same as R506 | 745－1394－00 |  | ohms $\pm 20 \%$ 1／4 w |  |
| R512R513 | RESISTOR．FIXED，COMPOSITION：same as R505 | 745－1380－00 | R573 | RESISTOR．FIXED，COMPOSITION：same as R502 | 745－1370－00 |
|  | RESISTOR．FIXED，FILM： 261 ohms $\pm 1 \%$ \％ $1 / 4 \mathrm{w}$ | 705－7068－00 | R574 | RESISTOR，FIXED，COMPOSITION：same as R502 | 745－1370－00 |
| R514 | RESISTOR，FIXED，COMPOSITION： 1800 ohms | 745－1363－00 | R575 | RESISTOR，FIXED，COMPOSITION：same as R505 | $\begin{aligned} & 745-1380-00 \\ & 745-1384-00 \end{aligned}$ |
|  | $\pm 10 \%, 1 / 2 \mathrm{w}$ RESISTOR，FIXED，COMPOSITION：same as R514 |  | R5\％6 | RESISTOR，FIXED．COMPOSITION： 5600 ohms | 745-1384-00 |
| $\begin{aligned} & \text { R515 } \\ & \text { R516 } \end{aligned}$ | RESISTOR，FIXED，COMPOSITION： 150 olams $\pm 10$ 䈭， $1 / 2 \mathrm{w}$ | 745－1317－00 | R577 | RESISTOR，FEXED，COMPOSITION： 39,000 ohms $\pm 10 \%, 1 / 2 \mathrm{w}$ | 745－1418－00 |
| R517 | RESISTOR，FIXED，FILM： 110 ohms $\pm 1 \%$ ． $1 / 4 \mathrm{w}$ | 705－7050－00 | R578 | RESISTOR，FIXED．COMPOSITION：47，000 ohms | 745－1422－00 |
| R518R519 | RESISTOR，FIXED，FILM： 6810 ohms $\pm 1 \%, 1 / 4 \mathrm{w}$ | 705－7136－00 |  | $\pm 10 \%$ \％ $1 / 2 \mathrm{w}$ |  |
|  | RESISTOR，FIXED，FILM：same as R518 | 705－7138－00 | R579 | RESISTOR，FIXED，COMPOSITION： 2150 ohms | 705－7112－00 |
| RS19 R520 | RESISTOR，FIXED，FILM：same as R517 | 705－7050－00 |  | $\pm 1 \%, 1 / 4$ w |  |
| R521 | RESISTOR，FIXED，FILM： 34,800 ohms $\pm 1 \%$ ， 1／4 w | 705－7170－00 | $\mathrm{R580}$ | RESISTOR．FIXED，COMPOSITION： 820 ohms $\pm 10 \% 1 / 2$ w | 745－1349－00 |
| R522 | RESISTOR，FIXED．FLLM： 10,000 ohms $\pm 1 \%$ ． $1 / 4 \mathrm{w}$ | 705－7144－00 | $\begin{aligned} & \text { R581 } \\ & \text { T501 } \end{aligned}$ | RESISTOR，FIXED，COMPOSITION：same 39 R552 NOT USED | 745－1373－00 |
| R523 | RESISTOR，FLXED，FILM： 178.000 ohms $\pm 1 \%$ ． 1／4 w | 705－7204－00 | $\begin{aligned} & \text { T502 } \\ & \text { T503 } \end{aligned}$ | NOT USED TRANSFORMER，RADIO FREQUENCY： 20 turns | 549－1589－00 |
| R524 | RESISTOR，FIXED．FILM： 14,700 ohms $\pm 1 \%$ ， $1 / 4$ w | 705－7152－00 |  | 430 AWG，close wound tapped at 10 turns； 43.5 uh inductance；ferrite core； 0.250 in ．w by 0.500 in．dia |  |


| ITEM | DESCRIPTION | COLLINS <br> part number | ITEM | DESCRIPTION | COLLINS PAET NUMBER |
| :---: | :---: | :---: | :---: | :---: | :---: |
| T504 | TRANSFORMER, RADIO FREQUENCY: | 549-1617-003 | C626 | CAPACITOR, FDXED, CERAMIC: same as C615 | 913-3886-00 |
|  | terminals. primary et; 5/8 in. h by 1-1/8 in. w |  | C627 | CAPACITOR, FXED, CERAMIC: same as C605 | 913-2680-00 |
|  | loy $1-1 / 2 \mathrm{in} . \mathrm{lg}$ |  | C628 | CAPACITOR. FIXED. CERAMIC: same as C605 | 913-2680-00 |
| TB501 | TERMINAL BOARD: phenolic. $1-7 / 8 \mathrm{~mm}$ le by $3 / 8$ | 306-0951-00 | C629 | CAPACITOR, FRXED, MICA: same as C625 | 912-2768-00 |
|  | in. w by $1 / 16$ in. thk: 5 brass solder-lug |  | C630 | CAPACITOR, FLXED, CERAMIC: same as C615 | 913-3886-00 |
|  | terminals; Cinch Mrg. Corp. part no. 1542-A-FV |  | C631 | CAPACITOR, FLXED. CERAMIC: same as C615 | 913-3886-00 |
| TB502 | TERMINAL BOARD: same as TB501 | 306-0951-00 | C632 | CAPACITOR, FLXED, MICA: 150 uuI $5 \% \%$, 500 | 912-2828-00 |
| TB503 | TERMINAL BOARD: same as TB501 | 306-0951-00 |  | vdew: Electro Motive part no. DM15F151J01 |  |
| TB504 | TERMINAL BOARD: Bakelite, 4 terminals. 1 | 306-2240-00 | C633 | CAPACITOR, FLXED, MICA: same as C632 | $912-2828-00$ |
|  | grounded, 3 insulated; 21/32 in. w by 1-1/2 in. ig; Cinch Mig, Corp. part no. 1534-A |  | C634 | CAPACITOR. FLXED, MICA: 39 uuf $\pm 5 \% .500$ vdcw; Electro Mative part no. DM15E390J01 | $912-2786-00$ |
| TB505 | TERMINAL BOARD same as TB501 | 306-0951-00 | C635 | CAPACITOR, FDXED, CERAMIC: same as C605 | 913-3680-00 |
| TB506 | TERMINAL BOARD: same as TE501 | 306-0951-00 | C636 | CAPACITOR. FIXED. CERAMIC: same as C605 | 913-3680-00 |
| TB507 | TERMINAL BOARD: Phenolic w/ 3 solder-lug | 306-9033-00 | C637 | CAPACITOR. FIXED, CERAMIC: same as C605 | -913-3680-00 |
|  | terminals; $11 / 16 \mathrm{in}$. w by 1-1/8 in. Igi Cinch |  | C638 | CAPACITOR, FIXED. CERAMIC: same as C605 | 913-3680-00 |
| TB508 | Mrg. Corp. part no. $1520-\mathrm{A}$ TERMINAL BOARD: same as TB501 |  | C639 | CAPACITOR, VARIABLE, CERAMIC: same as | 917-1.073-00 |
| TB509 | TERMINAL BOARD: phenolic. $1 / 16$ in | 306-9032-00 |  | C606 | 913-1186-00 |
|  | by $1-1 / 2$ in.; 4 brass solder-lug terminals: Cinch Mig. Corp. part no. 1532-A |  | C641 | CAPACITOR. FLXED. MICA: 68 uUf $\pm 5^{\circ} \%, 500$ | 912-2804-00 |
| TB510 | TERMINAL board : same as tb507 | 306-9033-00 | C642 | CAPACITOR. FLXED, MICA: 220 uuf $55 \%$ \%, 500 | 912-2840-00 |
| TB511 | TERMINAL BOARD: same as TB501 | 306-0951-00 |  | vdew, Electro Motive part no. DM15F221J01 |  |
| TB512 | TERMINAL BOARD: phenolic w/ 3 solder-lug | 306-0001-00 | C643 | CAPACITOR. FDXED, CERAMIC: same as C601 | $916-0362-00$ |
|  | terninals. $11 / 16 \mathrm{ln}$. w by $1-1,8 \mathrm{in} .1 \mathrm{Ig}$ C Cinch Mg. Corp. part no. 1525-A |  | C644 | CAPACITOR, VARIABLE. CERAMIC: 3.0 uuf mín to 12.0 uur max, 350 vdew | $917-1072-00$ |
| TB513 | TERMINAL BOARD: same as TB509 | 306-9032-00 | C645 | CAPACITOR, FIXED, MICA: 33 uuI $\pm 5 \%$. 500 | 912-2780-00 |
| TB514 | TERMINAL BOARD: same as TB507 | 306-9033-00 |  | vdew; Electro Motive part no. DM15F330JO1 |  |
| TB515 | TERMINAL BOARD: same as TB501 | 306-0951-00 | C646 | CAPACITOR, FLXED. MICA: same as C645 | 912-2780-00 |
| TB516 | TERMINAL BOARD: same as TB501 | 306-0951-00 | C647 | CAPACITOR, FIXED. MICA: 560 uuf 5 5\%. 500 | $912-2983-00$ |
| TB517 | TERMINAL BOARD. same as TB501 | 306-0951-00 |  | vdew; Electro Motive part no. DM19F561J |  |
| TB518 | TERMINAL BOARD: phenolic. 12 solder-lug terminals, Vector Mfg. Co. part no. 6H-12 | 306-0909-00 | C648 | CAPACITOR. FIXED. MICA: 1800 uul $15 \%, 500$ vdew; Electro Motive part no. DM20F182J | 912-3333-00 |
| TB519 | TERMINAL BOARD same as TB501 | 306-0951-00 | C649 | CAPACITOR. FIXED. ELECTROLYTIC: 100 uf | 184-7802-00 |
| TB520 | TERMINAL BOARD: phenolic $w / 4$ solder fug terminals; 27/32 in. w by 1-1/2 in. 1g; Cinch Mrg Corp part no 1009 | 306-0838-00 |  | $-15 \%+75 \% .25 \mathrm{vdcw}$; Sprague Electric part no. 109D107C7025T2 |  |
|  | Mg . Corp. part no. 1909 |  | C650 | CAPACITOR. FIXED, PAPER: same as C607 | 931-0170-00 |
| TD521 | TERMINAL BOARD: phenolic, 3 solder-lug | 306-0587-00 | C651 | CAPACITOR. FDCED, CERAMIC: same as C615 | 913-3886-00 |
| Y501 | terminals: $11 / 10 \mathrm{in} . \mathrm{w}$ by $1-1 / 8 \mathrm{in}$.lg CRYSTAL UNT, QUARTZ: 14.0 mc : type | 289-2743-00 | C652 | CAPACITOR, FEXED, CERAMIC: 10.0 uuf $上 1 / 4$ uuf, 500 vdcw | 916-0203-00 |
|  | HC-27/U holder |  | C653 | CAPACITOR. FXXED, MICA: 270 uuf $\pm 5 \%$. 500 vdew; Electro Molive part no. DM15F271J01 | 912-2846-00 |
| MODULATOR |  |  | C654 | CAPACITOR, DIODE: 35 uuf $\pm 20$ C. at $-4 \mathrm{vd}-\mathrm{c}$ voltíge, max $130 \mathrm{vd}-\mathrm{c}$; total capacity range 6 to 88 uul, 130 v d-c to 0.1 v d-c | 922-6002-00. |
| C601 | CAPACITOR. FDKED. CERAMIC: 20.0 uUS $\pm 2 \%$. 500 vdew | 916-0362-00 | CR601 | 88 uul, 130 v d-c to 0.1 vid-c SEMICONDUCTOR DEVICE, DIODE: quick recovery silicon junction diode; Hughes Aircraft | 353-2857-00 |
| C602 | CAPACITOR. FIXED. CERAMIC: same as C601 | 916-0362-00 |  |  |  |
| C603 | CAPACITOR. FDED. CERAMIC: uninsulated. 10.0 uul $\tau 1 / 2$ uuf. 500 vdew | 916-0412-00 | CR602 | SEMICONDUCTOR DEVICE. DIODE: silicon. hermettcally seated; Transitron Elect. Corp. | 353-3304-00 |
| C604 | CAPACITOR. FIXED, MICA: 100 UUF $45^{\circ} \mathrm{D} .500$ vdew: Electro Motive part no. DM15F101J01 | 912-2816-00 | CR603 | part no. SV3173 <br> SEMICONDUCTOR DEVICE DIODE. germanium. |  |
| C605 | CAPACITOR. FIXED, CERAMIC: $0.01 \mathrm{u}-0 \%$ | 913-3680-00 | CR603 | SEMICONDUCTOR DEVICE, DIODE: germanium. <br> Transitron part no. 1N270 | 353-2018-00 |
|  | 100\%. 100 vdew; Eric Resistor Corp. part no. 855-502-X550-103p |  | CR604 | SEMICONDUCTOR DEVICE. DIODE: same as CR603 | 353-2018-00 |
| C606 | CAPACITOR, VARIABLE. CERAMIC: 5.0 uuf min to 37.5 uuf max, 350 vdew; Eric Resistor Corp. | 917-1073-00 | CR605 | SEMICONDUCTOR DEVICE, DIODE: same as CR603 | 353-2018-00 |
| C607 |  | 031-0170-00 | CR60G | SEMICONDUCTOR DEVICE. DIODE: same as | 353-2018-00 |
|  | 200 vdew |  | CR607 | SEMICONDUCTOR DEVICE, DIODE: germanium; | 353-0160-00 |
| C608 | CAPACITOR. FIXED. ELECTROLYTIC: 250 uf $-10 \%-100 \%$. 12 vdew ; Sprague Electric Co. part no. 30D157A1 | 183-1190-00 | CR60日 | Eric Resistor part no. 1 N198 SEMICONDUCTOR DEVICE, DIODE: samc as CR607 | 353-0160-00 |
| C609 | CAPACITOR. FLXED, PAPER: 0.5 uf $-10 \%+20 \%$ 200 vdew | 931-0169-00 | CR609 | SEMICONDUCTOR DEVICE. DIODE: silicon; Texas Instruments part no. IN751A | 353-2710-00 |
| C610 | CAPACITOR. FIXED, PAPER: 20 uf $\pm 20 \% 150$ vdcw; Sprague Electric part no. 143P101M | 951-2004-00 | CR610 | SEMICONDUCTOR DEVCE, DIODE: same as CR807 | 353-0160-00 |
| C611 C612 | CAPACITOR. FIXED. MICA: same as C604 CAPACITOR. FLXED. MIC | $912-2816-00$ $912-2816-00$ | J601 | JACK, TIP: insulated tip u/w standard 0.080 in . | 360-0152-00 |
| C613 | CAPACITOR, FIXED. CERAMIC: same as C605 | 913-3680-00 |  | test probes; brown; E. F. Johnson Co. part no. 105-208-200 |  |
| C614 | CAPACITOR. FIXED, MICA: 330 uUF $55 \% .500$ vdew; Electro Motive part no. DM15F331JO1 | 912-2852-00 | J602 | JACK. TIP: insulated tip u/w standard 0.080 ln . test probes; red; E. F. Johnson Co. part no. | 360-0150-00 |
| C615 | CAPACITOR, FIXED. CERAMIC: 0.1 uf $-20 \%$ $+80^{\circ} \%$. 50 vdew; Sprague Electric part no. 33 C 41 | 913-3886-00 | J603 | $105-202-200$ <br> ЗACK. TIP: insulated $u p \mathrm{u} / \mathrm{w}$ standard 0.080 in . |  |
| C616 | CAPACITOR. FIXED. CERAMIC: same as C615 | 913-3886-00 | J603 | JACK, TIP: insulated tip u/w standard 0.080 in . test probes; orange: E. F. Johnson Co. part no. | 360-0154-00 |
| C617 | CAPACITOR. FIXED. CERAMIC: same as C605 | 913-3680-00 |  | 105-206-200 |  |
| C618 | CAPACITOR. FIXED. CERAMIC: 100 uUf $\pm 20 \%$, 500 vdew | 913-1186-00 | J604 | JACK, TELE PHONE: steel, miniature; panel mig; Switheraft, Inc. part no. 3501 FP | 360-0148-00 |
| C619 | CAPACITOR, FIXED. MICA: 10 uuf $\pm 5 \%$, 500 vdcw; Electro Molive part no. DM15F100J01 | 912-2753-00 | $J 605$ | JACK, TELEPHONE: same as J604 | 360-0149-00 |
| C620 | CAPACITOR, FIXED. MICA: $82 \mathrm{u} u f+5 \% .500 \mathrm{vdew}$; Electro Motive part no. DM15E820J01 | 912-2810-00 |  | ohms max d-c resistance. 570 ma current rating: Jeffers Electronics part no. 10404-112 | 240-0192-00 |
| C 621 | CAPACITOR, FIXED. CERAMIC: same as C615 | 913-3886-00 | L602 | INDUCTOR, RADIO FREQUENCY: toroidal, single | 240-1529-00 |
| C622 | CAPACITOR. FIXED. CERAMIC: same as C615 | 913-3886-00 |  | layer wound, approx 22 turns ${ }^{28}$ double formvar: |  |
| C623 | CAPACITOR. FIXED, CERAMIC: same as C605 | 913-3680-00 |  | 2.4 uh $\pm 2$ \% at 2.6 mc |  |
| C624 C625 | NOT USED <br> CAPACITOR, FIXED. MICA: 22 uuf $\pm 5 \%$. 500 vdew; Electro Motive part no. DM15C220J01 | 912-2768-00 | L603 | COIL. RADIO FREQUENCY: variable; $+15^{\circ} \mathrm{C}$ to $+55^{\circ} \mathrm{C}$ temp range; 100 v d-c dielectric strength | 278-0733-00 |


| ITEM | DESCRIPTION | COLLINS PART NUMBER |
| :---: | :---: | :---: |
| L804 | COIL, RADIO FREQUENCY: universal wound. ( 3 pl: 72 turns ea section. 36 AWG'wire; 220 uh inductance; 100 ma current; Delevan Electric part no. BS-217 | 240-0198-00 |
| L605 | COIL. RADIO FREQUENCY: same as L604 | 240-0198-00 |
| L606 | COIL. RADIO FREQUENCY: same as L603 | 278-0733-00 |
| L607 | COIL. RADIO FREQUENCY: same as L604 | 240-0198-00 |
| L608 | COIL, RADIO FREQUENCY: same as L603 | 278-0733-00 |
| L609 | COIL. RADIO FREQUENCY: same as L604 | 240-0198-00 |
| L610 | COIL. RADIO FREQUENCY: 1,00 uh $\mathrm{t} 10 \%$ 品 0.30 ohm d-c resistance; 850 ma dc; Jeffers Electronics part no. 10100-128 | 240-0062-00 |
| L611 | COIL, RADIO FREQUENCY: same as L603 | 278-0733-00 |
| L612 | COIL, RADIO FREQUENCY: same as L604 | 240-0198-00 |
| L613 | NOT USED |  |
| L614 | COIL, RADIO FREQUENCY: single layer wound. 56 uh inductance. 750 ma current; 1.30 ohms d-c; Jefrers Electronics Div. of Speer Carbon Co. part no. 10404-30 | 240-0191-00 |
| L615 | COIL, RADIO FREQUENCY: same as L601 | 240-0192-00 |
| L616 | COIL. RADIO FREQUENCY: same as L604 | 240-0198-00 |
| P601 | PLUG. TELEPHONE: brass; phenolic insulation, w/ solder-lug terminal; Switcheraft part no. 3501 MC | 361-0062-00 |
| P602. | NOT USED |  |
| P603 |  |  |
| P604 | PHONO, PLUG: w/ solder-lug terminals, phenolic insulation; Switcheralt. Inc. part no. 3501 MC | 361-0062-00 |
| Q601 | TRANSISTOR: germanium; hermetically sealed; Radio Corp. of America part no. 2N139g | 352-0376-00 |
| Q602 | TRANSISTOR: germanium; Radio Corp. of America part no. 2N1225 | 352-0135-00 |
| Q603 | TRANSISTOR: same as Q602 | 352-0135-00 |
| QC04 | TRANSISTOR: hermetically sealed, NPN silicon; Fairchild Semiconductor Corp. part no. 2N708 | 352-0322-00 |
| Q605 | TRANSISTOR: same as Q602 | 352-0135-00 |
| Q606 | TRANSISTOR: same as Q604 | 352-0322-00 |
| Q607 | TRANSISTOR: sllicon planar; hermelically sealed; Fairchild Semiconductor Corp. part no. S4639 | 352-0373-00 |
| Q608 | TRANSISTOR: same as 0601 | 352-0376-00 |
| R601 | $\begin{aligned} & \text { RESISTOR, FLXED, FILM: } 21,500 \text { ohms } \pm 1 \% \\ & 1 / 4 \mathrm{w} \end{aligned}$ | 705-7160-00 |
| R602 | RESISTOR, FIXED, FILM: 12,100 ohms $\pm 10$, 1/4 w | 705-7148-00 |
| R603 | $\begin{aligned} & \text { RESISTOR, FIXED. COMDOSITION: } 1000 \text { ohms } \\ & \pm 10 \%, 1 / 2 \mathrm{w} \end{aligned}$ | 745-1352-00 |
| R604 | $\begin{aligned} & \text { RESISTOR, FLXED, COM POSITION: } 47,000 \text { ohms } \\ & 110 \%, 1 / 2 \mathrm{w} \end{aligned}$ | 745-1422-00 |
| -R605 | RESLSTOR. FIXED, FILM: 1470 ohms $\pm 1 \% \mathrm{p}, 1 / 4 \mathrm{w}$ | 705-7104-00 |
| -R605 | RESISTOR. FIXED. FILM: 1960 ohms $\pm 1 \% .1 / 4 \mathrm{w}$ | 705-7110-00 |
| -R605 | RESISTOR. FIXED FILM: 4220 ohms $\pm 1 \% .1 / 4 \mathrm{w}$ | 705-7126-00 |
| -R605 | RESISTOR, FIXED, FILM: 1000 ohms $\pm 1 \% .1 / 4 \mathrm{w}$ | 705-7096-00 |
| R606 | $\begin{aligned} & \text { RESISTOR. FIXED. FILM: } 19.600 \text { ohms } \pm 1 \% . \\ & 1 / 4 \mathrm{w} \end{aligned}$ | 705-7158-00 |
| R607 | RESISTOR, FIXED, FILM: 2610 ohms $\pm 1 \% .1 / 4 \mathrm{w}$ | 705-7116-00 |
| R608 | RESISTOR, FIXED, COMPOSITION: 10,000 ohms $\pm 10 \% 1 / 2 \mathrm{w}$ | 745-1394-00 |
| RG09 | RESISTOR, FIXED, COMPOSITION: 5600 ohms i $10 \%, 1 / 2 \mathrm{w}$ | 745-1384-00 |
| R610 | RESISTOR, FEXED, COMPOSITION: 27,000 ohms x10\%, $1 / 2 \mathrm{w}$ | 745-1412-00 |
| R611 | RESISTOR. FIXED, COMPOSITION: 1500 ohms $+10 \% .1 / 2 \mathrm{w}$ | 745-1359-00 |
| R612 | RESISTOR, FEXED, COMPOSITION: 1800 ohms i $10 \%, 1 / 2 \mathrm{w}$ | 745-1363-00 |
| R613 |  | 705-7030-00 |
| R614 | RESISTOR. FESED. FILM: 51.1 ohms $\pm 1 \%$ \% $1 / 4 \mathrm{w}$ | 705-7034-00 |
| R615 | RESISTOR. FLXED, COMPOSITION: same as R608 | 745-1394-00 |
| R616 | RESISTOR. FIXED, COMPOSITION: same as R608 | 745-1394-00 |
| R617 | RESISTOR, FLXED, COMPOSITION 4700 ohms $\pm 10^{\circ} \mathrm{ow}, 1 / 2 \mathrm{w}$ | 745-1380-00 |
| R618 | RESISTOR, FIXED, FILM: 261 ohms $\pm 1 \%$ \% $1 / 4 \mathrm{w}$ | 705-7068-00 |
| R619 | RESISTOR, FEXED. COMPOSITION: same as R612 | 745-1363-00 |
| R620 | RESISTOR. FDEED, COMPOSITION: same as R612 | 745-1363-00 |
| R621 | RESISTOR, FDXED, COMPOSITION: 6800 ohms $\pm 10 \%, 1 / 2 \mathrm{w}$ | 745-1387-00 |


| ITEM | DESCRIPTION | COLLINS PART NUMBER |
| :---: | :---: | :---: |
| R622 | RESISTOR. FIXED. COMPOSITION: same as R617 | 745-1380-00 |
| R623 | RESISTOR. FLXED. COMPOSITION: same as R621 | 745-1387-00 |
| R624 | RESISTOR. FEXED, COMPOSITION: 220 ohms 410 \%, $1 / 2$ w | 745-1324-00 |
| R625 | RESISTOR, FIXED. COMPOSITION: same as R612 | 745-1363-00 |
| R626 | RESISTOR. FLXED. COMPOSITION: 2200 ohms $\pm 10 \%, 1 / 2 \mathrm{w}$ | 745-1366-00 |
| R627 | RESISTOR, FDEED. COMPOSITION: 390 ohms $\times 10 \%, 1 / 2 \mathrm{w}$ | 745-1335-00 |
| R628 | RESISTOR, FEXED. FILM: 1960 ohms $\pm 1 \%, 1 / 4 \mathrm{w}$ | 705-7110-00 |
| R629 | RESISTOR. FLXED. FILM: same as R628 | 705-7110-00 |
| R630 | RESISTOR, FIXED, FILM: 8250 ohms $\leq 1 \% .1 / 4 \mathrm{w}$ | 705-7140-00 |
| R631 | RESISTOR, FIXED. FILM: 1100 ohms $\pm 1 \% .1 / 4 \mathrm{w}$ | 705-7098-00 |
| R632 | RESISTOR. FIXED. FILM: same as R631 | 705-7098-00 |
| -R633 | RESISTOR, FIXED, FILM: 3480 ohms $\pm 1 \%, 1 / 4 \mathrm{w}$ | 705-7122-00 |
| ${ }^{*}$ R633 | RESISTOR, FIXED, FILM: 4220 ohms $11^{\circ} \mathrm{CD} .1 / 4 \mathrm{w}$ | 705-71 26-00 |
| ${ }^{-}$R633 | RESISTOR, FIXED, FILM 5110 ohms $\times 1.0 .1 / 4 \mathrm{w}$ | 705-7130-00 |
| -R633 | RESISTOR, FIXED, FILM: 10,000 ohms $41 \%$ \% $1 / 4 \mathrm{w}$ | 705-7144-00 |
| R634 | RESISTOR, FTXED. FILM: 348 ohms $\pm 1 \mathrm{C} .1 / 4 \mathrm{w}$ | 705-7074-00 |
| R635 | RESISTOR, VARIABLE: COMPOSITION; $50.000 \mathrm{ohms} \pm 30$ 免 $1 / 4 \mathrm{w}$ | 376-4737-00 |
| R636 | RESISTOR, FLXED. FILM: $1000 \mathrm{ohms}=1 \% .1 / 4 \mathrm{w}$ | 705-7096-00 |
| R637 | RESISTOR, FIXED, FILM: 7500 ohms $\pm 1 \%$, $1 / 4 \mathrm{w}$ | 705-7138-00 |
| R638 | $\begin{aligned} & \text { RESISTOR, FIXED, FILM: } 100.000 \text { ohms } t 1 \% . \\ & 1 / 4 \mathrm{w} \end{aligned}$ | 705-7192-00 |
| R639 | RESISTOR, FIXED, FILM: 316 ohms $\pm 1$ \% $\%$ \% $1 / 4 \mathrm{w}$ | 705-7072-00 |
| R640 | RESISTOR, FLXED, FILM: 5110 ohms $\pm 1 \%, 1 / 4 \mathrm{w}$ | 705-7130-00 |
| R641 | RESISTOR, VARIABLE: COMPOSITION; 50.000 ohms i $30 \%$. $1 / 4 \mathrm{w}$ | 376-4732-00 |
| R642 | RESISTOR, FIXED, FILM: 13.300 ohns it $\%$ \% $1 / 4 \mathrm{w}$ | 705-7150-00 |
| R643 | RESISTOR, FIXED. FILM: 2870 ohms $\pm 1 \% 1 / 4 \mathrm{w}$ | 705-7118-00 |
| R649 | RESISTOR. FIXED. COMPOSITION: 0.10 megohm $\therefore 10 \%, 1 / 2 \mathrm{w}$ | 745-1436-00 |
| R645 | RESISTOR. FEXED. COMPOSITION: same as R644 | 745-1436-00 |
| R646 | RESIS'OR, FIXED. COMPOSITION: same as R644 | 745-1436-00 |
| R647 | RESISTOR, FIXED, COM POSITION: 150 ohms $\pm 10 \% .1 / 2 \mathrm{w}$ | 745-1317-00 |
| R648 | RESISTOR, FIXED, COMPOSITION: 680 ohms t10'o, $1 / 2$ w | 745-1345-00 |
| R649 | RESISTOR. FIXED, COM POSITION: same as R608 | 745-1394-00 |
| R650 | RESISTOR, FIXED, COMPOSITION: same as R626 | 745-1366-00 |
| R651 | $\begin{aligned} & \text { RESISTOR, FDXED, COMPOSITION: } 22 \text { ohms } \\ & \pm 10^{\circ} \mathrm{o}, 1 / 2 \mathrm{w} \end{aligned}$ | 745-1282-00 |
| T601 | TRANSFORMER, RADIO FREQUENCY: 5 terminals primary, cl; 5/8 in. h by 1-1/8 in. w by 1-1/2 in. lg; Collins Radio Co. | 549-1617-003 |
| TB601 | TERMINAL BOARD: phenolic w/ 3 solder-lug terminals; $11 / 16 \mathrm{in}$. w by $1-1 / 8 \mathrm{in}$. Ig; Cinch | 306-9033-00 |
| TB602 | Mfg. Corp. part no. 1520-A TERMINAL BOARD: phenolic, $1 / 16$ in. by $3 / 8 \mathrm{in}$. by $1-1 / 2$ in.; 4 brass solder lug terminals; Cinch Mig. Corp. part no. I532-A | 306-9032-00 |
| TB603 | TERMINAL BOARD: same as TB602 | 306-9032-00 |
| TB604 | TERMINAL BOARD: same as TB601 | 306-9033-00 |
| TB605 | TERMINAL BOARD: same as TB601 | 306-9033-00 |
| TB606 | TERMINAL BOARD: same as TB601 | 306-9033-00 |
| TB607 | TERMINAL BOARD: phenolle w/ 3 solder-lug terminals; $11 / 16 \mathrm{in}$. w by 1-1/8 in. lg; Cinch MLg. Corp. part no. 1525-A | 306-0001-00 |
| TB608 | TERMINAL BOARD: phenolic, $1-7 / 8 \mathrm{in}$. by $3 / 8 \mathrm{in}$. by $1 / 16$ !n.; 5 brass solder-lug terminals; Cinch Mig. Corp. part no. 1542-A-FV | 306-0951-00 |
| TB609 | TERMINAL BOARD: same as TB608 | 306-0951-00 |
| TB610 | TERMINAL BOARD: laminated phenolic $w / 4$ solder lug terminals; 27/32 in. w by 1-1/12 in. lg overall, Cinch Mrf. Corp. part no. 1909 | 306-0838-00 |
| TB611 | TERMINAL BOARD: same as TB610 | 306-0838-00 |
| TB612 | TERMINAL BOARD: same as TB601 | 306-9033-00 |
| TB613 | TERMINAL BOARD: same as TB601 | 306-9033-00 |
| TB614 | TERNINAL BOARD: phenolic. 12 solder lug terminals; Voctor Mig. Co. part no. 6H-12 | 306-0909-00 |
| TB615 | TERMINAL BOARD: same as TB608 | 306-0951-00 |
| TB616 | TERMINAL BOARD: phenolic, 3 solder-lyg terminals; $11 / 16 \mathrm{in}$. w by $1-1 / 8 \mathrm{in}$. 1 s | 306-0587-00 |



Figure 4-1. Modulator Compartment, Component (Except Resistors) Identification


Figure 4-2. Modulator Compartment, Resistor Identification


Figure 4-3. AFC Compartment, Component (Except Resistors) Identification



Figure 4-5. Power Amplifier Compartment, Component (Except Resistors) Identification


Figure 4-6. Power Amplifier Compartment, Resistor Identification

TD-536
A830-2 10 W Wide-Band FM Broadcast Exciter


Figure 4-7. Power Supply, Component Identification


Figure 4-8. Chassis, Component Identification


Sample Voltages Read on A830 Exciter With Vacuum Tube Voltmeter

Transistor
Q501
Q502
Q503
Q504
Q505
Q506
Q507
Q508
Q509
Q510
Q511
Q512
Q601
Q602
Q603
Q604
Qó05
Q606
Q607
Q608
(
Tube Type
6U8A

## Pin Number to Ground

1
2
3.
6
7
.
9
12AT76AU6
-

## 6AU6

## 5763



Vcltage
$+48.0$
$-1.25$
$+78.0$
$+99.0$
$+0.37$
-0. 8
$+260.0$
$-3.2$
$+1.5$
$+260.0$
-3.1
$+1.5$
-5.2
$+295.0$
+208.0
$+0.67$
$+293.0$
$+260.0$
$+11.0$
$-2.35$

Sample Voltages Read on A830 Exciter - Continued:
be Type
226
1,4,6
3
5
Plate at By-Pass Feedthru TrS01
TP502
TP503

TP504

Output Minimum
$+26.0$
$+188.0$
-15.1
$+360.0$
-8.0 Volts +0.1 or -0.1 volt
+5.7 volts

- . volt

$$
\begin{aligned}
& \text { (with AFC disabled, meter } \\
& \text { varies and goes toward zero) } \\
& \text { (with AFC disabled, pulsing } \\
& \text { begins - release, meter goes } \\
& \text { toward zero, then goes to }+8.0 \\
& \text { volts and gradually returna to } \\
& +5.7 \text { volts) }
\end{aligned}
$$

# 786M-1 Stereo Generator 



## unit instructions

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## SECTION I GENERAL DESCRIPTION

### 1.1 PURPOSE OF INSTRUCTION BOOK.

Unit Instructions TD-537 provides information about $786 \mathrm{M}-1$ Stereo Generator, Collins part number $522-$ 2914-00. Information which is furnished includes a general clescription of the equipment, principles of operation, maintenance procedures, and a parts list.

### 1.2 PURPOSE OF THE EQUIPMENT.

The $786 \mathrm{M}-1$ Stereo Generator is used to convert stereophonic audio input signals into main and stereophonic subchannel signals and to generate a pilot subcarrier. The resultant signal is suitable for modulation of wideband FM broadcast exciters.

### 1.3 DESCRIPTION OF EQUIPMENT.

### 1.3.1 PHYSICAL DESCRIPTION.

The $786 \mathrm{M}-1$ Stereo Generator, shown in figure $1-1$, is constructed on a standard 19 -inch rack-mounted panel. This panel is 19 inches wide, $10-1 / 2$ inches high, 7
inches deep, and weighs approximately 14 pounds. All operating controls are located on the front panel with seldom-used adjustments located inside the back panel. A meter is placed conveniently on the lower left portion of the front panel for monitoring of input and output signals. All transistors and the $38-\mathrm{kc}$ crystal are placed on the front panel for convenient access. Bulky components are grouped in the upper right-hand corner of the front panel leaving the remaining portion of the panel free of obstacles for ease of maintenance and adjustment. All components located in the rear of the unit are protected in a dust-resistant enclosure.

### 1.3.2 ELECTRICAL DESCRIPTION.

The $786 \mathrm{M}-1$ Stereo Generator is an all transistorized unit consisting of the following circuits; a crystal controlled 38 -kilocycle oscillator, a 19 -kilocycle locked oseillator, a meter amplifier, two audio amplifiers, and a balanced modulator. All components for operation of the time division stereo generator are


Figure 1-1. 786M-1 Stereo Generator, Over-all View
contained within the 19 -inch rack-mounted panel with the exception of a left audio channel pre-emphasis network. This function must be supplied externally and is available in the Collins A830-2 10 W Wide-Band FM Broadcast Exciter.

Power input required is $20 \pm 0.1$ volts d-c which is supplied by the A830-2. Remote control can be exercised over the stereophonic and monophonic modes. Power required for operation of remote control is 28 $\pm 2.8$ volts d-c. Instruction books covering the exciter and power amplifiers, used in conjunction with the $786 \mathrm{M}-1$, are listed in table 1-1.

TABLE 1-1
ASSOCLATED EQUIPMENT INSTRUCTION BOOKS

| ASSOCIATED <br> EQUIPMENT | INSTRUCTION BOOK <br> PART NUMBER |
| :--- | :---: |
| A830-2 10 W Wide-Band <br> FM Broadcast Exciter <br> B830-1 250-Watt FM <br> Power Amplifier <br> D830-1 1000-Watt FM <br> Power Amplifier <br> E830-1 5-Kw FM <br> Power Amplifier | TD-536 |

### 1.4 EQUIPMENT SUPPLIED.

Table 1-2 lists equipment that is supplied as part of the $786 \mathrm{M}-1$ Stereo Generator.

TABLE 1-2
EQUIPMENT SUPPLIED

| EQUIPMENT | COLLINS <br> PART NUMBER |
| :---: | :---: |
| $786 \mathrm{M}-1$ Stereo Generator | $522-2914-00$ |

### 1.5 EQUIPMENT REQUIRED BUT NOT SUPPLIED.

Table 1-3 lists equipment that is required for operation of the $786 \mathrm{M}-1$ Stereo Generator but not supplied as part of the $786 \mathrm{M}-1$.

TABLE 1-3
EQUIPMENT REQUIRED BUT NOT SUPPLIED

| EQUIPMENT | COLLINS <br> PART NUMBER |
| :---: | :---: |
| A830-2 10 W Wide-Band <br> FM Broadcast Exciter | $522-2714-00$ |

### 1.6 ACCESSORY EQUIPMENT.

Table 1-4 lists accessory equipment that is available for use with $786 \mathrm{M}-1$ Stereo Generator.

TABLE 1-4 ACCESSORY EQUIPMENT

| EQUIPMENT | COLLINS <br> PART NUMBER |
| :--- | :---: |
| B830-1 250-Watt <br> FM Power Amplifier <br> D830-1 1000-Watt <br> FM Power Amplifier <br> E830-1 5-Kw <br> FM Power Amplifier <br> $250-$ Watt/1-Kw <br> Harmonic Filter | $549-2008-00$ |

### 1.7 EQUIPMENT SPECIFICATIONS.

### 1.7.1 MECHANICAL.

Weight . . . . . . . 14 pounds approximately.
Size . . . . . . . . 19 inches wide, $10-1 / 2$
inches high, 7 inches deep.

Ambient temperature
range . . . . . . $+15^{\circ} \mathrm{C}\left(59^{\circ} \mathrm{F}\right)$ to $45^{\circ} \mathrm{C}$

Ambient humidity
range . . . . . . . . 0 to $95 \%$ relative humidity.
Altitude . . . . . . . 0 to 7500 feet.

### 1.7.2 ELECTRICAL.

$$
\begin{aligned}
& \text { Power source . . . . . } 20 \pm 0.1 \text { volts d-c. } \\
& 28 \pm 2.8 \text { volts d-c (for re- } \\
& \text { mote operation). }
\end{aligned}
$$



Distortion . . . . . Less than $0.5 \%$ over the $\begin{aligned} & \text { frequency range of } 30 \text { to } \\ & 15,000 \mathrm{cps} .\end{aligned}$
38-ke subcarrier suppression . . . . . 40 db below output with 10 dbm input level.

Main channel and stereophonic subcarrier
phase relationship. . . $\pm 3$ degrees for audio frequencies from 50 to 15,000 cps.

Audio-frequency
response . . . . . . . Complies with FCC standard 75-microsecond preemphasis curve (right channel only, left channel pre-emphasis is obtained from exciter).

### 1.8 SEMICONDUCTOR COMPLEMENT.

Table 1-5 lists the semiconductor complement supplied as part of 786 M -1 Stereo Generator.

TABLE 1-5. SEMICONDUCTOR COMPLEMENT

| SYMBOL | QUANTITY | TYPE | FUNCTION |
| :---: | :---: | :---: | :---: |
| Q701 | 1 | 2N1613 | 38-kc oscillator |
| Q702 | 1 | 2N1285 | 38-kc buffer |
| Q703 | 1 | 2N1613 | 38-ke driver |
| Q704 | 1 | 2N1285 | $38-\mathrm{kc}$ isolation amplifier |
| Q705 | 1 | 2N708 | 19-kc locked oscillator |
| Q706 | 1 | 2N1175A | Left audio amplifier |
| Q707 | 1 | 2N1175A | Right audio amplifier |
| Q708 | 1 | 2N1285 | Meter amplifier |
| CR702 | 4 | 1 N 270 | Balanced modulator diode switches |

## SECTION II <br> PRINCIPLES OF OPERATION

### 2.1 GENERAL.

The $786 \mathrm{M}-1$ Stereo Generator provides facilities for the conversion of stereophonic input signals to an output which conforms to the standards approved by the FCC for the transmission of stereophonic signals. The following paragraphs discuss stereophonic principles and the operation of the $786 \mathrm{M}-1$ Stereo Generator.

### 2.2 PRINCIPLES OF FM STEREO.

### 2.2.1 STEREOPHONIC SOUND SYSTEMS.

An elementary stereophonic sound system consists of two directional microphones placed to the right and left of a sound source. See figure 2-1. Each microphone in turn is connected to an amplifier and speaker system. When the listener is situated between the speakers, the left channel will be received by the left ear and the right channel will be received by the right
ear. The effect upon the listener of such a system is to simulate placing the listener at a point midway between the two microphones and receiving a true representation of the originating sound source.

To provide a realistic stereo effect, the difference in time delay and signal amplitude from the sound source to each of the microphones must be maintained through the entire stereo system. If the time delay or amplitude difference is changed in one of the amplifier or speaker systems, the effect to the listener will be a change in direction of the sound source, when in reality no change has occurred. A change intime delay of the left or right channel is referred to as a phase relationship change. This phase relationship change between the channels must be held, in stereo transmitting equipment, to within $\pm 3$ degrees.

If the amplitude difference and time delay in each system is identical (as when the sound source is centered between the microphones), the sound source will


Figure 2-1. Elementary Stereophonic System


Figure 2-2. Spectrum of Signals in Baseband Audio
appear to the listener to be centered between the speakers. This is actually the true relationship of the

To enhance the stereo effect to the listener, it is desirable for each microphone to be directional, as stated previously, so that sounds originating directly in front of the right microphone will be received by the right microphone, and as little as possible by the left microphone, and vice versa. If too much of the right sound source is picked up by the left microphone or vice versa, the effect to the listener will be to move the sound source to the center. This isolation between the two sound systems is known as channel separation and must be held greater than 29.7 db in stereo transmitting equipment.

If proper isolation of the amplifiers is not obtained, and there is an interchange of signals, the sound source will again appear to move toward the center. If the channel separation is reduced to zero, the effect would be to replace the two microphones with a single microphone feeding the same information to both amplifier and speaker systems. It is then understood that monophonic operation can be obtained by paralleling (adding) the left and right microphone outputs. This monophonic component is referred to as $L+R$.

An interchange of information between channels (main and subchannel) is referred to as crosstalk. Crosstalk will deteriorate the stereo signals by adding noise to the signal. In stereo transmitting systems, crosstalk must remain at least 40 dbbelow a single channel level.

### 2.2.2 METHODS OF GENERATING FCC STEREO.

Signals which are prescribed by the FCC for the transmission of stereophonic intelligence is shown in figure $2-2$. This band of frequencies must be generated and transmitted in order that both monaural and stereophonic receivers will be able to detect the FM signal. For monaural receivers, only the L+R (left plus right) channel is received, with the pilot carrier and $L-R$ (left minus right) signals rejected by the pass band of the monaural receiver. Stereophonic FM receivers
detect the complete band of frequencies in a discriminator and will process the signals into left and right stereophonic channels. The $19-\mathrm{kc}$ pilot carrier is used in this process. The method of generating the signals shown in figure 2-2 depends upon the method chosen for modulating an FM signal.

The methods of modulating an FM signal may be broken down into two groups, a direct and an indirect method. These two general categories may be broken down further into various methods of obtaining the end result. Phase modulation is the most generally used method of generating an FM signal by the indirect method. If this system is used to modulate the composite stereo signal, various problems are encountered. The most serious problem is that of frequency response of the phase modulator. As the phase modulator audio response exhibits nonlinear modulation characteristics (rises 6 db per octave from the lowest to the highest frequency), predistortion is employed to compensate for this trait. In a stereo FM phase modulator this predistortion would amount to 65.5 db over the entire modulating frequency range of 50 cps to 75 kc (SCA added to stereo signal). When a $65-\mathrm{db}$ signal to noise ratio and a $60-\mathrm{db}$ dynamic range is added to this, it is apparent that baseband amplifiers cannot be built to give this characteristic.

It is possible to split the phase modulation into two steps and modulate one phase modulator with the L+R signal and the second phase modulator with the L-R and pilot carrier signals. This type of stereophonic phase modulation is not desirable because of the required phase linearity of $\pm 3$ degrees and the gain requirements of stereophonic transmission systems. These requirements are difficult, if not impossible, to maintain.

Another method of FM stereophonic modulation which could be employed is a combination of direct and indirect modulation. With this method the L+R signal directly FM modulates an oscillator, while the L-R signal phase modulates the signal produced in the oscillator, in a later stage. As in phase modulation of the stereophonic signal, it is difficult to maintain phase linearity and gain characteristics.


Figure 2-3. An Elementary Time Division Multiplex System

A third method of generating a stereophonic FM signal is by the use of direct modulation over the entire stereophonic generator frequency range. Phase relationship and gain characteristics are then easy to maintain because of the point input source. Until recently, it has been difficult to directly modulate an FM signal with a wide bandwidth of signals. With the advent of solid state components and specifically the production of the variable capacity diode, this wideband type of modulation is possible. This is the type of modulation used in the Collins A830-2 10 W WideBand FM Broadcast Exciter. The development of the wide-band type of modulator made possible the development of 786M-1 Stereo Generator which is discussed in paragraph 2.3.

### 2.3 PRINCIPIES OF OPERATION OF 786M-1.

The $786 \mathrm{M}-1$ Stereo Generator generates the spectrum of signals shown in figure $2-2$ by the time division multiplex method. By this method, shown simplified in figure 2-3, the left and the right channels are switched alternately at a $38-\mathrm{kc}$ rate. If the receiver switching rate is synchronized with the transmitter switching rate, the original left and right audio signals will be detected. In the receiver, the $19-\mathrm{kc}$ pilot carrier is doubled to synchronize the receiver to the transmitter. It is important that the switching frequency in both the stereo generator and the receiver be of the same phase to retain the identity of the left and right audio signals.

The mathematical analysis of two audio signals being switched alternately by a square wave shows that the resultant signal is made up of two components of interest. One component is directly proportional to the sum of the two audio signals ( $\mathrm{L}+\mathrm{R}$ ) and the other is a double-sideband (DSB) signal centered on a frequency equal to the switching frequency ( 38 kc ). The modulation appearing on this DSB signal is directly proportional to the difference of the two audio channels ( $L-R$ ). If $L$ is defined as the audio signal in the left channel, $\mathbf{R}$ is defined as the audio signal in the right channel, and $f_{C}$ is defined as the switching frequency, the composite signal is equal to:

$$
\begin{aligned}
& \frac{L+R}{2}+\frac{2}{\pi}(L-R) \cos 2 \pi f_{c} t-\frac{2}{3 \pi}(L-R) \cos 6 \pi f_{c} t \\
& \quad+\frac{2}{5 \pi}(L-R) \cos 10 \pi f_{c} t \text { etc. }
\end{aligned}
$$

The first term of this expansion is the main channel component, the second term is the stereophonic subchannel component, and the remaining terms are higher frequency terms which are undesired.
The following is a block diagram explanation of the $786 \mathrm{M}-1$ Stereo Generator which generates the signals just mentioned. Refer to figure 2-4. It is noted on the block diagram that the left audio channel is fed through the pre-emphasis network and high-pass filter of the A830-2 wide-band exciter and then to the $786 \mathrm{M}-1$.

The right audio channel utilizes the pre-emphasis network and the high-pass filter located in the $786 \mathrm{M}-1$. The outputs of the high-pass filters are fed through 15kc low-pass filters where audio components above 15 kc are attenuated sharply. The $15-\mathrm{kc}$ filter outputs are then fed to emitter follower amplifiers where isolation of the two channels from the balanced modulator is obtained. The two-channel audio outputis then fed to a balanced modulator whose action resembles that of a switch. The balanced modulator utilizes the signal generated in the $38-\mathrm{kc}$ oscillator to alternately switch on and off each audio channel. The balanced modulator consists of the composite spectrum which includes both desired and undesired components. The fundamental 38 -kc modulating signal and all evenorder harmonics are balanced out.

The balanced modulator output is mixed with a small amount of direct $\mathrm{L}+\mathrm{R}$ signal which equalizes the peak amplitude of the main and subchannel signals. The modulator output is then fed through a $50-\mathrm{cps}$ to $53-\mathrm{kc}$ low-pass linear phase filter where all odd harmonics above 53 kc are attenuated. The filter output is mixed with a 19 -ke signal from the pilot carrier phase locked oscillator and is fed to the $786 \mathrm{M}-1$ output. All FCC phasing, channel separation, crosstalk, and amplitude specifications are satisfied within the $786 \mathrm{M}-1$ Stereo Generator.

### 2.3.1 DETAILED DESCRIPTION OF 786M-1 STEREO GENERATOR.

Refer to figure 5-1, a schematic diagram of the 786M-1. The right audio channel is identical to the left audio channel except that the pre-emphasis network and the $15-\mathrm{kc}$ filter for the left channel are located within the A830-2 exciter. Only the right channel is discussed in the following paragraphs.


Figure 2-4. 786M-1 Stereo Generator, Block Diagram

The 600 -ohm balanced right audio channel is fedinto a pre-emphasis network, FL701. Due to the inherently low level of high-frequency audio components in program material, pre-emphasis is employed to overcome the effects of noise which is often found in home receivers. The $786 \mathrm{M}-1$ follows the standard 75microsecond pre-emphasis curve established by the FCC. The output of pre-emphasis filter FL701 is fed into a $30-\mathrm{cps}$ high-pass filter, FL702, which sharply attenuates audio components below 30 cps . This is necessary to prevent 5 -cps audio components from interfering with the 5 -cps sampling circuits within the A830-2 exciter. Filter FL702 also transforms the 600ohm balanced input into a 600 -ohm single-ended output.

The output of FL702 is connected to relay K701 which remotely selects either the stereo mode for transmission, or the left or right audio channels for monaural operation. Relay K701 operates by applying 28 volts d-c across the solenoid. This 28 -volt d-c source is supplied by the A830-2 exciter. Selection of either the left or right monaural channel is determined by the position of S701. At this point, if either the left or right channel is selected for monaural operation, the single $50-\mathrm{cps}$ to $15-\mathrm{kc}$ audio signal is fed through an
$8-\mathrm{db}$ loss pad to the output of the $786 \mathrm{M}-1$ Stereo Generator. The $8-\mathrm{db}$ loss pad is made up of R750, R751, and R752. The resulting audio input to the A830-2 is the same as that obtained without the stereo generator. Switch 5701 will also override the remote relay if desired.
If the stereo mode is selected by S701, the audio component is fed to a $15-\mathrm{kc}$ low-pass filter. FL704. FL704 attenuates all frequencies over 15 kc to prevent their interfering with adjacent channels. The output of FL704 is fed to the base of emitter follower Q707, which isolates the audio circuits from the balanced modulator.
The function of the balanced modulator is to generate the $L+R$ and the $L-R$ components shown in figure 2-2. The balanced modulator resembles a switch which samples the left audio channel and the right audio channel in turn. The 38-kc switching frequency and all even order harmonics are balanced out in the modulator output. The $38-\mathrm{kc}$ switching frequency is obtained from the $38-\mathrm{kc}$ driver and is impressed across transformer T701. If the primary switching voltage is negative, the secondary voltage will switch on diodes CR703 and CR704. Thus, right audio will appear at the secondary center tap. If the primary switching


LEFT CHANNEL SAMPLED AT 38KC



COMPOSITE SIGNAL AUDIO OUTPUT

Figure 2-5. Balanced Miodulator Output When $L+R=2, L-R=0$
voltage is positive, the secondary voltage will switch on diodes CR702 and CR705. The left audio channel will then appear on the secondary of T701. A representation of a sine wave input in each channel ( $\mathrm{L}=\mathrm{R}$, $\mathrm{L}+\mathrm{R}=2, \mathrm{~L}-\mathrm{R}=0$ ) switched in this manner is shown in figure 2-5. It is seen in this illustration that the composite signal at the output of the balanced modulator is a sine wave of an amplitude equal to the original signal level in each channel. The spikes shown on the composite sine wave result from imperfect switching and are filtered out in FL705.

Figure 2-6 shows the balanced modulator output when $R=0, L+R=1$, and $L-R=1$. The output of the balanced modulator is an audio component plus DSB components centered on the switching frequency and odd harmonics which form the square wave shape. When the odd harmonics are filtered out by the $53-\mathrm{kc}$ harmonic


Figure 2-6. Balanced Modulator Output When $\mathrm{L}+\mathrm{R}=1$ and $\mathrm{L}-\mathrm{R}=1$
filter, FL705, the third waveform results. Because the fundamental component of a square wave is $\frac{4}{\pi}$ times the square wave amplitude, the DSB component is larger than the audio. The audio component is then increased by $\frac{4}{\pi}$ and the fourth illustration results. The audio component is added by resistors R724 and R'730 which leak a small portion of L+R directly around the balanced modulator. Potentiometer R755 adjusts the audio component so the $\frac{4}{\pi}$ loss in filtering is exactly compensated. Capacitors C736 through C739 are selected capacitor values which balance out the stray balanced modulator capacitance. This balanced condition reduces the suppressed $38-\mathrm{kc}$ switching frequency level to well below the 40 db required by the FCC. On some units these capacitors are unnecessary.

Figure 2-7 shows the time division signal when $L=-R$, or $L+R=0, L-R=2$. The composite waveform from the balanced modulator is shown in the third illustration. This waveform is composed of audio components and odd harmonics centered on the switching frequency. When the odd harmonics are removed by filtering in FL705, the waveshape in the fourth illustration results. This waveshape is a DSB signal which equals $L-R$ as required by the matrix process.


Figure 2-7. Balanced Modulator Output When $\mathrm{L}+\mathrm{R}=0$ and $\mathrm{L}-\mathrm{R}=2$

The output of the balanced modulator and $\mathrm{L}+\mathrm{R}$ mixing is fed to a low-pass 53 -kc filter, FL705. Filter FL705 removes all harmonics and noise above 53 kc to form the DSB waveshape as shown in figures 2-6 and 2-7. The output from FL705 is mixed with a $19-\mathrm{kc}$ pilot carrier and fed to the stereo override switch, S701, and the remote relay, K701. Operation at this point is similar to audio switching which was discussed earlier. If relay K701 is energized and S701 is in the STEREO ON position, the composite stereo is fed to J701 for connection to the A830-2 10 W Wide-Band FM Exciter.

The balanced modulator switching frequency is obtained from crystal-controlled oscillator Q701. Oscillations are sustained by taking the output of L701


Figure 2-8. 786M-1 Control and Adjustment Locations
and feeding it into the base of Q 701 . The $38-\mathrm{kc}$ output of L701 is also capacitively coupled into the $38-\mathrm{kc}$ buffer amplifier, Q702. The output of Q702 is tuned to 38 kc by C714 and L702. The output of buffer amplifier Q702 is further amplified to approximately 4 volts peak to peak by driver amplifier Q703. The gains of Q701, Q702, and Q703 are stabilized by emitter degeneration to reduce gain variations between transistors. The output of Q703 is capacitively coupled to the primary of T701 (balanced modulator switching transformer) and to the 19-kc pilot carrier locked oscillator through an isolation stage, Q704.

The pilot carrier oscillator, Q705, is basically a grounded base oscillator which is synchronized by injecting a $38-\mathrm{kc}$ signal into the base. The oscillator output is a $19-\mathrm{kc}$ resonant tank placed across the base to emitter junction by means of a capacity voltage divider. The $19-\mathrm{kc}$ output is taken from the emitter circuit and is injected into the output of FL705. The pilot carrier phase, which must be maintained in phase with the output of FL705, is adjusted by varying the inductance of L704. Pilot carrier level is adjusted with R748.

Metering circuits are provided within the $786 \mathrm{M}-1$ to assist in trouble shooting. Meter amplifier Q705
provides isolation of the matrixing and oscillator circuits from the metering circuits. The right audio and left audio channels are fed directly from the 600ohm balanced input through meter multiplying resistors R711 and R710 to meter M701.

### 2.3.2 CONTROL FUNCTIONS.

The following paragraphs describe the functions of all controls in the $786 \mathrm{M}-1$ Stereo Generator. Refer to figure 2-8 for control locations.
Meter selector S703 connects meter M701 into various circuits for monitoring purposes. The metering positions are as follows; L AUDIO (left audio), R AUDIO (right audio), MXOUTPUT (multiplex output), R AUDIO OSC ( $38-\mathrm{kc}$ oscillator), 38 KC BUF ( $38-\mathrm{kc}$ buffer amplifier), 38 KC DRIVER, and 19 KC LKD OSC (19kc locked oscillator output).
Audio input switch 5701 selects one of three possible audio inputs; left audio, right audio. and stereo. If switch 5701 is placed in the left audio or right audio positions, remote relay K701 is able to provide remote control over the monaural or stereo modes. When S701 is in the stereo mode, relay K 701 is disabled and has no effect on stereo generator inputs.

CARRIER BALANCE controls R703 and R726 balance out the $38-\mathrm{kc}$ carrier and $76-\mathrm{kc}$ second harmonic in the secondary of T701. These controls are adjusted for zero indication at TP701 with no audio in either channel.
Channel separation $L+R$ amplitude control (CHAN SEP $\mathrm{L}+\mathrm{R}$ AMPL) R755 adjusts the amount of $\mathrm{L}+\mathrm{R}$ fed around the balanced modulator to raise the L+R level by $\frac{4}{\pi}$.
Inductor L701 adjusts the frequency of the $38-\mathrm{kc}$ oscillator. Resistor R713 adjusts the level of the $38-\mathrm{kc}$
driver output into the balanced modulator. This level is set for 6 volts peak to peak at TF701 at the factory and should never need readjustment.

PILOT CARRIER PHASE control L704 adjusts the phase of the $19-\mathrm{kc}$ pilot carrier. The control is set for an in-phase condition with relation to the output of FL705. PILOT CARRIER LEVEL control R748 adjusts the level of the 19-kc pilot carrier. This control is set for 0.009 volt rms at TP701. PILOT CARRIER switch S702 turns the 19-kc pilot carrier off and on for adjustment and testing purposes.

### 3.1 GENERAL.

This section contains information concerning the maintenance of the $786 \mathrm{M}-1$ Stereo Generator.

## NOTE

As some transistor cases are electrically above ground, do not short transistor cases to ground or damage to the transistor may result. Always replace transistors with the transistor locating mark placed adjacent to the transistor socket.

### 3.2 SERVICING TRANSISTOR CIRCUITS.

Servicing procedures and test equipments that have been used in the past with other types of electronic equipment, for the most part, may be used with transistor circuits. Some special precautions which must be used are listed below.

### 3.2.1 TEST EQUIPMENT.

Damage to transistors by test equipment is usually the result of accidentally applying too much voltage to the transistor elements. Common causes of damage from test equipment are as follows:
a. Test equipment with a transformerless power supply is one source of such voltage. This type of test equipment can be used by employing an isolation transformer in the power line.
b. It is still possible to damage transistors from line voltage even though the test equipment has a power transformer in the power supply, if the test equipment is equipped with a line filter. This filter may act like a voltage divider and apply 55 volts $a-c$ to the transistor. To eliminate trouble from this situation, connect a ground wire from the chassis of the test equipment to the chassis of the equipment under test before making any other connections.
c. Another cause of transistor damage is a multimeter that requires excessive current for adequate
indications. Multimeters that have sensitivities of less than 5000 ohms per volt should not be used. A multimeter with lower sensitivity will draw too much current through many types of transistors and damage them. Use of 20,000 -ohm-per-volt meters or vacuumtube volt meters is recommended. Check the ohmmeter circuits (even those in vtvm's) on all scales with an external, low-resistance milliammeter in series with the ohmmeter leads. If the ohmmeter draws more than one milliampere on any range, this range cannot be used safely on small transistors.

### 3.2.2 ELECTRIC SOLDERING IRONS.

The following are possible causes of transistor damage from soldering irons:
a. Electric soldering irons may damage transistors through leakage current. To check a soldering iron for leakage current, connect an a-c volt meter between the tip of the iron and a ground connection, allow the iron to heat, then check for a-c voltage with the meter. Reverse the plug in the a-c receptacle and again check for voltage. If there is any indication on the meter, isolate the iron from the a-c line with a transformer. The iron may be used without the isolation transformer if the iron is plugged in and brought to temperature then unplugged for the soldering operation. It is also possible to use a ground wire between the tip of the iron and the chassis of the equipment being repaired to prevent damage from leakage current.
b. Light-duty soldering irons of 20 to 25 watts capacity are adequate for transistor work and should be used. If it is necessary to use a heavier duty iron, wrap a piece of number 10 copper wire around the tip of the iron and make it extend beyond the tip of the iron. Tin the end of the piece of copper wire and use it as the soldering tip.

### 3.2.3 SERVICING PRACTICES.

a. If a transistor is to be evaluated in an external test circuit, be sure that no more voltage is applied
to the transistor than normally is used in the circuit from which it came.
b. Test prods should be clean and sharp. Because many of the resistors used in transistorized equipments have low values, any additional resistance produced by a dirty test prod will make a good resistor appear to be out of tolerance.

### 3.2.4 TROUBLE SHOOTING.

The usual trouble-shooting practices apply to transistors. Be sure the test equipment and tools meet the requirements outlined in the above paragraphs. It is recommended that transistor testers be used to evaluate the transistor.
If a transistor tester is not available, a good ohmmeter may be used for testing. Be sure the ohmmeter meets the requirements as set forth in the paragraph on test equipment, above. To check a PNP transistor, connect the positive lead of the ohmmeter to base and the negative lead to the emitter. (The
red lead is not necessarily the positive lead on all ohmmeters.) Generally, a resistance reading of 50,000 ohms or more should be obtained. Connect the negative lead to the collector; again a reading of 50,000 ohms or more should be obtained. Reconnect the circuit with the negative lead of the ohmmeter to the base. With the positive lead connected to the emitter, a value of resistance in the order of 500 ohms or less should be obtained. Likewise, with the positive lead connected to the collector, a value of 500 ohms or less should be obtained.

Similar tests made on an NPN transistor produce results as follows: With the negative ohmmeter lead connected to the base, the value of resistance between the base and the emitter and between the base and the collector should be high. With the positive lead of the ohmmeter connected to the base, the value of resistance between the base and the emitter and between the base and collector should be low. If the readings do not check out as indicated, the transistor probably is defective and should be replaced.

## CAUTION

If a defective transistor is found, make sure that the circuit is in good operating order before inserting the replacement transistor.

Make sure that the value of the bias resistors in series with the various transistor elements are as shown on the schematic diagram. The transistor is very sensitive to improper bias voltages; therefore, a short or open circuit in the bias resistors may damage the transistor. For this reason, do not trouble-shoot by shorting various points in the circuit to ground and listening for clicks.

### 3.3 TROUBLE SHOOTING.

Trouble shooting can best be accomplished by using standard trouble-shooting techniques. Suspected troubles should be isolated to individual stages before
components are replaced. The pilot carrier can be turned off with switch 5702 as an aid in trouble shooting and testing.

### 3.4 ADJUSIMENTS AND TESTS.

The $786 \mathrm{M}-1$ is fitted with adjustments which adjust 38 -kc oscillator tuning, carrier balance, pilot carrier level, and pilot carrier phase.

## NOTE

Do not attempt the following adjustments without using the proper test equipment as serious deterioration of the $786 \mathrm{M}-1$ output quality may result from the use of inferior test equipment.

The test equipments or their equivalents required to perform the specified tests are listed in table 3-1.

TABLE 3-1
TEST EQUIPMENT REQUIRED

| EQUIPMENT | MANUFACTURER <br> AND TYPE |
| :--- | :--- |
| Oscilloscope | Tektronix Model 545A with <br> Type 53/54C plug-in unit and <br> a Type D plug-in unit |
| Oscillator <br> Distortion and <br> noise meter <br> Hewlett-Packard Model 200AB | Hewlett-Packard Model 330D <br> Vewlett-Packard Model 400H <br> (or equivalent) |

Figure 3-1 is a standard transistor base, viewed from the bottom, which provides a transistor element reference.

### 3.4.1 38-KC OSCILLATOR TUNING.

Turn on the A830-2 10 W Wide-Band FM Exciter. Connect an a-c vtum to the collector of Q702. (See figure 3-1.) Turn R713 fully clockwise. Adjust L701 for a maximum indication on the vtvm. The oscillator output at the collector of Q702 should be approximately 1.5 volts.


Figure 3-1. Transistor Base Configuration

### 3.4.2 38-KC AMPLITUDE CHECK.

Connect a calibrated Tektronix oscilloscope, provided with a Type D plug-in unit, across terminals 1 and 2 of T701. The voltage at this point should be 6 volts peak to peak as read on the oscilloscope. Adjust R713 if necessary to obtain 6 volts.

### 3.4.3 CARRIER BALANCE.

Turn the PILOT CARRIER switch to OFF. Remove any audio from the left and right audio channels. Connect the Tektronix oscilloscope with the Type D
plug-in unit to TP702 and ground. Adjust in turn R703 and R726 in small steps for a minimum indication on the oscilloscope. The final indication on the oscilloscope must be more than 40 db below 100 millivolts ( 10 millivolts).

### 3.4.4 PILOT CARRIER PHASE.

Connect the Tektronix oscilloscope to the $786 \mathrm{M}-1$ Stereo Generator as shown in figure 3-2. Connect the audio oscillator into the $786 \mathrm{M}-1$ through $10-\mathrm{db}$ pads to give an $\mathrm{L}=-\mathrm{R}$ signal (right audio channel 180 degrees out of phase with the left audio channel) into the audio input terminals at a frequency of 1000 cps and a level of 7.8 volts rms. Set the PILOT CARRIER switch to OFF. Switch the CONTROL switch to STEREO ON. Adjust the RILOT CARRIER PHASE control until both traces on the oscilloscope are stationary and an exact coincidence of the zerocrossings of the $19-\mathrm{kc}$ pilot carrier and the $\mathrm{L}-\mathrm{R}$ signal is obtained as shown in figure 3-3. Expand the sweep to 5 X , and adjust the horizontal position knob to check the two points of coincident zero crossing.

### 3.4.5 PILOT CARRIER LEVEL.

Remove any audio from the $786 \mathrm{M}-1$ audio input channels and connect a vtvm to TP702. Set the PILOT CARRIER switch to ON, and adjust the PILOT CARRIER LEVEL control for a reading of 0.009 volt rms as read on the vtvm.

| CONTROL SETTINGS |  |
| :--- | :---: | :---: |
| 1. CHANNEL A | $0.05 \mathrm{~V} / \mathrm{CM}, \mathrm{TP} 702$ |
| 2. CHANNEL G | $0.05 \mathrm{~V} / \mathrm{CM}, \mathrm{TP70/}$ |
| 3. MODE - ALTERNATE |  |
| 4. TRIGGERING MODE -AUTOMATIC TRIGGER SLOPE - <br> + EXTERNAL |  |
| 5. SWEEP TIME/CM 5 USEC |  |
| 6. MAGNIFIER |  |

Figure 3-2. Pilot Carrier Phase Test Setup

maladjustment of pilot CARRIER PHASE CONTROL.


PROPER ADJUSTMENT OF PILOT
CARRIER PHASE CONTROL.


PROPER ADJUSTMENT OF PILOT CARRIER PHASE CONTROL, EXPANDED HORIZONTAL DEFLEC TION.

Figure 3-3. Pilot Carrier Phase Adjustment, Oscilloscope Pattern

### 3.4.6 CHANNEL SEPARATION ADJUSTMENT.

Set the audio oscillator to 5000 cps , and connect it to the left audio input of the $786 \mathrm{M}-1$. Connect the Tektronix oscilloscope with the type $D$ plug-in unit to TP702 and ground, and adjust the audio oscillator for a 300 -millivolt peak-to-peak indication on the oscilloscope. Adjust the CHAN SEP LiR AMPL control to produce a straight zero axis (within 4 millivolts) as shown in figure 3-4. Repeat with the audio input into the right audio channel. The final adjustment must bring the zero axis to within 4 millivolts of a straight zero axis.

### 3.5 MINIMUM PERFORMANCE STANDARDS.

### 3.5.1 OVER-ALL GAIN.

a. Connect the Tektronix oscilloscope to TP702 and ground.
b. Switch the PILOT CARRIER switch to OFF.
c. Connect the audio oscillator through 10 -db pads to the $786 \mathrm{M}-1$ in such a way to obtain an $\mathrm{L}=\mathrm{R}$ signal (left channel equal in amplitude and phase with right channel).
d, Adjust the audio oscillator frequency to 1000 cps , and adjust the audio oscillator output to obtain 0 VU on the $786 \mathrm{M}-1$ VU meter when the METER switch is set to L AUDIO or R AUDIO. The peak-to-peak indication on the oscilloscope shall be from 200 to 300 millivolts.
e. Connect the audio input so $\mathrm{L}=-\mathrm{R}$ (right channel equal in amplitude but 180 degrees out of phase with
the left channel). The peak-to-peak indication shall be from 200 to 300 millivolts.

### 3.5.2 FREQUENCY RESPONSE.

a. Connect the distortion analyzer between TP702 and ground.
b. Switch the PILOT CARRIER switch to OFF.
c. Connect the audio oscillator through $10-\mathrm{db}$ pads to each channel in such a way to obtain an $L=R$ signal (left channel equal in amplitude and in phase with right channel).
d. Adjust the audio oscillator frequency to 1000 cps , and adjust the audio oscillator output to obtain 0 VU on the $786 \mathrm{M}-1$ VU meter when the METER switch is set to $L$ AUDIO. Set the distortion analyzer to 0 db .
e. Set the audio oscillator to 50 cps , and adjust the audio level from the audio oscillator for 0 VU on the $786 \mathrm{M}-1$ VU meter. The indication on the distortion analyzer shall be within $\pm 0.5 \mathrm{db}$ of the level at 1000 cps.
f. Repeat step e at $15,000 \mathrm{cps}$. The indication on the distortion analyzer shall be within $\pm 1.5 \mathrm{db}$ of the level at 1000 cps.
g. Repeat steps d, e, and f with the METER switch set at R AUDIO.

### 3.5.3 HARMONIC DISTORTION.

a. Connect the test setup as described in paragraph 3.5.2, steps $a, b$, and $c$.
b. The distortion at 50,1000 , and $15,000 \mathrm{cps}$ should be not more than one percent.


MALADJUSTMENT OF CHANNEL SEPARATION L AND R AMPL CONTROL


PROPER ADJUSTMENT OF CHANNEL
SEPARATION LANDR AMPL CON-
TROL.

Figure 3-4. Channel Separation Adjustment, Oscilloscope Pattern

| ITEM | DESCRIPTION | COLLINS PART NUMBER |
| :---: | :---: | :---: |
|  | 786M-1 STEREO GENERATOR | 522-2914-00 |
| C701 | CAPACITOR, FIXED, ELECTROLYTIC: 30 uf $-10 \%+100 \%, 10 \mathrm{v}$ d-c | 183-1377-00 |
| C702 | CAPACITOR, FIXED, ELECTROLYTIC: 50 uf $-10 \%+100 \%, 25 \mathrm{v} \mathrm{d}-\mathrm{c}$ | 183-1379-00 |
| C703 | CAPACITOR, FIXED, ELECTROLYTIC: same as C701 | 183-1377-00 |
| C704 | CAPACITOR, FIXED, ELECTROLYTIC: same as C702 | 183-1379-00 |
| C705 | CAPACITOR, FIXED, ELECTROLYTIC: 250 uf $-10 \%+100 \%, 12 \mathrm{v}$ d-c; Sprague Electric part no. 30D157A1 | 183-1190-00 |
| C708 | CAPACITOR, FIXED, ELECTROLYTIC: same as C705 | 183-1190-00 |
| C707 | CAPACITOR, FIXED, ELECTROLYTIC: 15 uf $-10 \%+100 \%$, 25 v d-c; Sprague Electric part no. 40D180N1 | 183-1362-00 |
| C708 | CAPACITOR, FIXED, MICA: 6800 uf $\pm 10 \%, 300 v$ d-c | 935-2110-00 |
| C709 | CAPACITOR, FIXED, PAPER: $0.047 \mathrm{uT} \pm 10 \%$. 400 v d-c; Sprague Electric part no. 160p47304 | 931-0295-00 |
| C710 | CAPACITOR, FIXED, PAPER; $0.1 u f \pm 10 \%, 400 \mathrm{v}$ d-c; Sprague Electric part no. I60P10494 | 031-0299-00 |
| C711 | CAPACITOR, FIXED, ELECTROLYTIC: 20 Uf $-10 \%+100 \%, 25 \mathrm{v}$ d-c; Sprague Electric part no. 40D181A2 | 183-1365-00 |
| C712 | CAPACITOR, FIXED, ELECTROLYTIC: same as C711 | 183-1365-00 |
| C713 | CAPACITOR, FIXED, ELECTROLYTIC: same as C711 | 183-1365-00 |
| C714 | CAPACITOR, FIXED, MICA: 1800 uUf $\pm 5 \%, 500 v$ d-c; Electro Motive part no. DM20F182J500WV | 912-3333-00 |
| C715 | CAPACITOR, FIXED, PAPER: same as C710 | 931-0299-00 |
| C716 thru C719 | CAPACITOR, FIXED. ELECTROLYTIC: same as C711 | 183-1365-00 |
| C720 | CAPACITOR, FIXED, MICA: 510 uuf $\pm 5 \%, 500$ v d-c; Electro Motive part no. DM19E511J | 012-2980-00 |
| C721 | CAPACITOR, FIXED, ELECTROLYTIC: same as C711 | 183-1385-00 |
| C722 | CAPACITOR, FIXED, ELECTROLYTIC: same as C711 | 183-1385-00 |
| C723 | CAPACITOR, FIXED, FILM: same as C710 | 931-0299-00 |
| C724 | CAPACITOR, FIXED, ELECTROLYTIC: same as C711 | 183-1365-00 |
| C725 | CAPACITOR, FIXED, MICA: 10,000 uuf $上 2 \%, 500$ $v$ d-c; Electro Motive part no. DM30F103G | 912-2734-00 |
| C726 | CAPACITOR, FIXED, PAPER: 0.0015 uf $\pm 10 \%$, 1000 v d-c: Surague Electric Co. part no. 106 | 931-0279-00 |
| C 727 | CAPACITOR, FIXED, ELECTROLYTIC: same as C711 | 183-1365-00 |
| C728 | CAPACITOR, FIXED, ELECTROLYTIC: same as C711 | 183-1365-00 |
| C729 | CAPACITOR, FIXED, ELECTROLYTIC: same as C711 | 183-1365-00 |
| C730 | CAPACITOR, FIXED, ELECTROLYTIC: 20 uf $-10 \%-100 \%, 50 \mathrm{vd}-\mathrm{c}$ | 183-1369-00 |
| C731 | CAPACITOR, FIXED, ELECTROLYTIC: same as C711 | 183-1385-00 |
| C732 | CAPACITOR, FIXED, CERTAMIC: 4700 uUI $\pm 20 \%$, 500 v d-c: MIL type CK62AW472M | 913-1187-00 |
| C733 thru C735 | CAPACITOR, FIXED, CERAMIC: sume as C732 | 913-1187-00 |
| -C736 | CAPACITOR, FIXED, CERAMIC: 5 uuf $\pm 1 / 2$ uuf $_{\text {, }}$ 500 v d-c: MIL type CC20CH050D | 916-0118-00 |
| -C736 | CAPACITOR, FIXED, CERAMIC: 10 uuf $\pm 1 / 2$ uUf, 500 vd-c; MLL type CC20CH100D | 916-0138-00 |
| - 6738 | CAPACITOR, FIXED, CERAMIC: 15 uU $\pm 5 \%, 500$ $v$ d-c: MIL type CC20CHI50J | 916-0671-00 |
| ${ }^{\text {c }}$ C738 | CAPACITOR, FIXED, CERAMIC: 20.0 Uuf $\mathbf{5} \%$, 500 v d-c; MIL type CC20CH200J | 916-0677-00 |
| *Selec | and added by test in the vicinity of TB714. TB718, | T 701 |


| ITEM | DESCR!PTION | COLLINS PART NUMBER |
| :---: | :---: | :---: |
| -C737 | CAPACITOR, FIXED. CERAMIC: 5 uuf $\pm 1 / 2$ uuf, 500 v d-c: MIL type CC20CH050D | 916-0118-00 |
| *C737 | CAPACITOR, FIXED, CERAMIC: 10 uuf $\pm 1 / 2$ uff, 500 v d-c. MIL type CC20CH100D | 916-0138-00 |
| - C 738 | CAPACITOR, FIXED. CERAMIC: 1.0 uUf $\pm 1 / 2$ uUf, 500 v d-e: MIL type CC20CKO10D | 916-0071-00 |
| -C738 | CAPACITOR, FIXED, CERAMIC. 2.0 uu $\pm 1 / 2$ uul. 500 v d-c: MIL type CC20CK020D | 916-0076-00 |
| -C738 | CAPACITOR, FIXED, CERAMIC: 3.0 uuf $\pm 1 / 2$ uur, 500 v d-c: MIL type CC20CJ030D | 916-0145-00 |
| - ${ }^{\text {C738 }}$ | CAPACITOR, FIXED, CERAMIC: 4.0 uuf $\pm 1 / 2$ uuf, 500 v d-c; MIL type CC20CH040D | 916-0114-00 |
| -C739 | CAPACITOR, FIXED, CERAMIC: 1.0 uf $\pm 1 / 2$ uuf, 500 ₹ d-c; MIL type CC20CK010D | 916-0071-00 |
| -C739 | CAPACITOR, FIXED, CERAMIC: 2.0 uuf $\pm 1 / 2$ uur, $500 \mathrm{vd-c}$; MLL type CC20CK020D | 916-0076-00 |
| ${ }^{\text {c }}$ C739 | CAPACITOR, FIXED, CERAMIC: 3.0 uf $\pm 1 / 2$ uur, 500 v d-c; MIL type CC20CJ030D | 916-0145-00 |
| -C739 | CAPACITOR, FIXED, CERAMIC: 4.0 uuf $\pm 1 / 2$ uuf, 500 v d-c; MIL lype CC20CH040D | 916-0114-00 |
| CR701 | NOT USED |  |
| CR702 | SEMICONDUCTOR DEVICE, SET: four hermeti- | 353-2041-00 |
| $\begin{aligned} & \text { A,B,C, } \\ & \& D \end{aligned}$ | eally sealed matched germanium diodes; Hughes Products part no. MQ4032 |  |
| FL701 | ATTENUATOR, FIXED: pre-emplasis network for $u / i n$ FM commercial broadcast equipment; 75 microseconds. 600 ohms input and output | 379-0426-00 |
| FL702 | FILTER, HIGH PASS: metal encased, hermetically sealed, input 600 ohms, output 600 ohms. 4 soldertype terminals, continuous duty cycle: A.D.C. part no. D10390 | 673-0869-00 |
| FL703 | FILTER, LOW PASS: continuous dut; cycle, input 600 ohms $\pm 20 \%$. output 600 ohms $\pm 20 \%$, metal encased, hermetically sealed; C.A.C. part no. 90-1015-00 | 673-0871-00 |
| FL704 | FILTER, LOW PASS: same as FL703 | 673-0871-00 |
| FL705 | FILTER, LOW PASS: Hnear, continuous duty cycle, input 600 ohms $\pm 20 \%$, output 600 ohms $\pm 20 \%$, metal encased, hermetically sealed, solder-type terminals; C.A.C. part no. 90-1012-00 | 673-0870-00 |
| J701 | JACK, TIP: insulated for $\mathbf{u} / \mathrm{w} 0.080 \mathrm{in}$. test probes; brown; E.F. Johnson part no. 105-208-200 | 360-0152-00 |
| J702 | JACK, TIP: insulated for u/w 0.080 in . test probes; red: E.F. Johnson part no. 105-202-200 | 360-0150-00 |
| K701 | RELAY, ARMATURE. 4 C contact arrangement: $0.25 \mathrm{amp}, 300 \mathrm{v}$ d-c, 1 Inductive winding. 250 ohms resistance, $27.5 \vee \mathrm{~d}-\mathrm{c} ; 0.11 \mathrm{amp}$ approx operating current; Aemco, Inc. part no. 94-3473 | 974-0127-00 |
| L701 | COIL, RADIO FREQUENCY: multilayer solenold type winding; 2.3 ohms; $-15^{\circ} \mathrm{C}$ to $+55^{\circ} \mathrm{C} ; 0.5$ to 3.5 mh : Chicago Standard Transformer Corp. part no. WC. 7 | 278-0734-00 |
| L702 | COIL, RADIO FREQUENCY: single layer wound, 10,000 uh, 66.5 ohms d-c, 75 ma current rating; Delevan part no. 2500-76 | 240-2564-00 |
| L703 | COIL, RADIO FREQUENCY: same as L702 | 240-2564-00 |
| L704 | COIL, RADIO FREQUENCY: 8 mh inductance; w/core; $1-5 / 32 \mathrm{in}$. by $1-5 / 32 \mathrm{in}$. by $2-1 / 2 \mathrm{in}$. excl terminals; Communications Coil part no. X-544-2 | 278-0780-00 |
| M701 | METER, AUDIO LEVEL. VU Meter for use in equipnents exposed to enviromments; background color, white | 456-0056-00 |
| 0701 | KNOB: setscrew type; black phenolic body; 1.125 in. dia by 0.843 in. thk $w /$ skirt | 546-1294-003 |
| 0702 | KNOB: same as 0701 | 546-1294-003 |
| P701 | PLUG, TELEPHONE: brass; phenolic Insulation, w/solder-lug terminal; Switcheraft part no. 3501 MC | 361-0062-00 |
| Q701 | TRANSISTOR: hermetically sealed NPN diffused silicon planar transistor; Fairchild Semiconductor Corp. part no. 2N1613 | 352-0349-00 |
| Q702 | TRANSISTOR: germanium; RCA part no. 2N1285 | 352-0243-00 |
| Q703 | TRANSISTOR: same as Q701 | 352-0349-00 |
| Q704 | TRANSISTOR: same as Q702 | 352-0243-00 |
| Q705 | TRANSISTOR: hermetically sealed; NPN sillcon; Fairchild Semiconductor Corp. part no. 2N708 | 352-0322-00 |

TD-537
786M-1 Stereo Generator

| ITEM | DESCRIPTION | COLLINS part number | ITEM | DESCRIPTION | COLLINS <br> part number |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Q706 | ANSISTOR: hermetically sealed. PNP germa- | 352-0315-00 | R760 | RESISTOR, FIXED, COMPOSITION: same as R719 | 745-1398-00 |
|  | nium; General Electric part no. 2N1175A |  | R761 | RESISTOR. FLXED. FILM: same as R724 | 705-7110-00 |
| Q707 | TRANSISTOR: samie as Q706 | 352-0315-00 | R762 | RESISTOR. FIXED, COMPOSITION: same as R717 | 745-1352-00 |
| Q708 | TRANSISTOR: same as Q702 | 352-0243-00 | R763 | RESISTOR, FIXED, COMPOSITION: 390 ohms | 745-1335-00 |
| R701 | RESISTOR: FIXED, FILM: 750 ohms $+1 \% .1 / 4 \mathrm{w}$ | 705-7090-00 |  | +10\%, $1 / 2 \mathrm{w}$ |  |
| R702 | RESISTOR. FIXED, FILM: same as r701 | 705-7090-00 | R764 | RESISTOR, FLXED. COMPOSITION: 0.12 megohm | 745-1440-00 |
| R703 | RESISTOR, VARIABLE, WIREWOUND: 10 ohms $\pm 10^{0}, 2 w$ | 377-0113-00 | R765 | $\begin{aligned} & \pm 10 \%, 1 / 2 w \\ & \text { RESISTOR, FIXED. FILM: } 5110 \text { ohms } \pm 1 \%, 1 / 4 \mathrm{w} \end{aligned}$ | 705-7130-00 |
| R704 | RESISTOR, FIXED. COMPOSITION: 3900 ohms | 745-1377-00 | R766 | RESISTOR, FIXED, FILM: same as R765 | 705-7130-00 |
|  | $\pm 10_{6,1 / 2}^{c}$ w |  | R767 | RESISTOR, FIXED. FILM: same as R765 | 705-7130-00 |
| R705 | RESISTOR. FIXED, COMPOSITION: $15,000 \mathrm{ohms}$ | 745-1401-00 | R768 | RESISTOR. FIXED, FILM: same as R765 | 705-7130-00 |
|  | $10^{\text {c }}$ \%, $1 / 2 \mathrm{w}$ |  | R769 | RESISTOR. FIXED. COMPOSITION: same as R717 | 745-1352-00 |
| 8706 | RESISTOR. FIXED. COMPOSITION: 3300 ohms | 745-1373-00 | R770 | RESISTOR, FIXED, COMPOSITION: same as R719 | 745-1422-00 |
|  | ${ }^{+10 \%} 1 / 2 \mathrm{w}$ |  | R771 | RESISTOR, FIXED, COMPOSITION- same as R705 | 745-1401-00 |
| R707 | RESISTOR, FIXED, COMPOSITION: same as R704 | 745-1377-00 | R772 | RESISTOR, FLXED, COMPOSITION: same as R705 | 745-1401-00 |
| R708 | RESISTOR, FIXED. COMPOSITION: same as R705 | 745-1401-00 | S701 | SWITCH SECTION, ROTARY: 6 circuit, 3 position; | 259-1597-00 |
| R709 | RESISTOR. FIXED. COMPOSITION: same as R706 | 745-1373-00 |  | 3 section; 4 noving. 16 fixed contacts; Oak Mig. |  |
| R710 | RESISTOR. FIXED. FILM: 1330 ohms $\pm 150.1 / 4 \mathrm{w}$ | 705-7102-00 |  | Co. part no. 222274-AH3 |  |
| R711 | RESISTOR. FIXED, FILM: same as R710 | 705-7102-00 | S702 | SWITCH, TOGGLE: spdt; 40 amp continuous; 28 v | 266-3099-00 |
| R712 | RESISTOR, FIXED, COMPOSITION: 33.000 ohnis $+10 \mathrm{G} .1 / 2 \mathrm{w}$ | 745-1415-00 |  | d-c, 20 amp resistive. 15 amp inductance; 115 v . 400 chs .10 amp resistance, 10 amp inductance; |  |
| R713 | RESISTOR, VARLABLE, COMPOSITION: 5000 ohms $\mathrm{t} 2 \mathrm{O}_{\mathrm{O}}^{\prime}, 0.2 \mathrm{w}$ | 376-0205-00 | S703 | Hetherington, Ine. part no. T1003-AN SWITCH SECTION, ROTARY: 4 circuit, 7 position, | 259-1596-00 |
| R714 | RESISTOR. FIXED. COMPOSITION: 10.000 ohms ${ }^{*} 10^{c}$ c. $1 / 2 \mathrm{w}$ | 745-1304-00 |  | 4 section; 3 moving, 24 fixed contacts; Grigsby Athison Co., Inc. part no. A25242-4MLR-4 |  |
| R715 | RESISTOR, FIXED, COMPOSITION: 120 ohms $+10 \%, 1 / 2 \mathrm{w}$ | 745-1314-00 | T701 | TRANSFORMER, RADIO FREQUENCY, BALANCED: c/o plastic fabric base phenolic board | 549-1639-00 |
| R716 | RESISTOR. FIXED, COMPOSITION: 4700 ohms $\pm 10 \% .1 / 2 \mathrm{w}$ | 745-1380-00 |  | 1/16 in. by $1-3 / 16 \mathrm{in}$. by $1-3 / 16 \mathrm{in}$.: plus 3 coils. 75 turns ea; coil 11 , wound cew , coils in 2 and in3, |  |
| R717 | RESISTOR, FIXED, COMPOSITION: 1000 ohms $\pm 10 \%$ \% $1 / 2 \mathrm{w}$ | 745-1352-00 | TB701 | cw; plus plastic rod 0.159 in . w by 0.413 in , dia TERMINAL BOARD: phenolic. barrier type w/ lug | 367-0020-00 |
| R718 | RESISTOR, FIXED. COMPOSITION: same as R712 | 745-1415-00 |  | for back connection. 12 terminals |  |
| R719 | RESISTOR, FIXED. COMPOSITION: 12,000 ohms $+10 \mathrm{C}_{1} 1 / 2 \mathrm{w}$ | 745-1308-00 | TB702 | TERMINAL BOARD: bakelite, 4 terminals, $3 / 8 \mathrm{in}$. by 1/2 in. by 1-1/2 in.; Cinch Mig. Corp, part no. | 306-2240-00 |
| R720 | RESISTOR. FIKED. COMPOSITION: 6800 ohms $\pm 10 \%, 1 / 2 \mathrm{w}$ | 745-1387-00 | TB703 | 1534-A <br> TERMINAL BOARD: same as TB702 | 306-2240-00 |
| R721 | RESISTOR, FIXED, COMPOSITION: 180 ohms $\div 10 \%, 1 / 2 \mathrm{w}$ | 745-1321-00 | TB704 | TERMINAL BOARD: 4 solder-lug terminals. brass: $3 / 8$ in, by 1-1/2 in. overall | 306-0698-00 |
| R722 | RESISTOR. FEXED, COMPOSITION: 2700 ohms $\dot{10} 0 \% 1 / 2 \mathrm{w}$ | 745-1370-00 | TB705 | TERMINAL BOARD: phenolic, 4 brass solder lug terminats; $1 / 16 \mathrm{in}$. by $3 / 8 \mathrm{in}$. by $1-1 / 2 \mathrm{in}$.; Cinch | 306-0032-00 |
| R723 | RESISTOR, FDKED, COMPOSITION: 39 ohms $\pm 10 \%, 1 / 2 \mathrm{w}$ | 745-1293-00 | TB706 | Mfg. Corp. part no. 1532A TERMINAL BOARD: phenolic. 3 solder-lug | 306-0587-00 |
| R724 | RESISTOR. FIKED, FILM: 1960 ohms $\pm 1 \%$, $1 / 4 \mathrm{w}$ | 705-7110-00 |  | terminals; $11 / 16 \mathrm{in}$. by $1-1 / 8 \mathrm{in}$. lf |  |
| R725 | RESISTOR, FLXED, FILM: 464 ohms $\pm 1$ \% ${ }^{\text {a }}$, $1 / 4 \mathrm{w}$ | 705-7080-00 | T8707 | TERMINAL BOARD: same as TB706 | 306-0587-00 |
| R726 | RESISTOR, VARIABLE, WIREWOUND: same as R703 | 377-0113-00 | TB708 TB709 | TERMINAL BOARD: same as TB702 <br> TERMINAL BOARD: phenolic. 5 brass solder lug | $\begin{aligned} & 306-2240-00 \\ & 306-0951-00 \end{aligned}$ |
| R727 | RESISTOR, FIXED, FILM: samie as R725 | 705-7080-00 |  | terminals; $1 / 16$ in. by $3 / 8$ in. by $1-7 / 8$ in., Cinch |  |
| R728 | RESISTOR, FIXED. FILM: same as R725 | 705-7080-00 |  | Mig. Corp. part no. 1542-A-FV |  |
| R729 | RESSSTOR, FIXED, FILM: same as R725 | 705-7080-00 | TB710 | TERMNAL BOARD same as TB702 | 306-2240-00 |
| R730 | RESISTOR, FIXED. FILM: same as R724 | 705-7110-00 | TB711 | TERMINAL BOARD: laminated phenolic w/ 4 | 306-0838-00 |
| R731 | RESISTOR, FIXED, COMPOSITION: same as R716 | 745-1380-00 |  | solder lug terminals, $27 / 32 \mathrm{in}$. w by $1-1 / 2 \mathrm{in}$. Ig; |  |
| R732 | RESISTOR, FIXED, COMPOSITION: same as R706 | 745-1373-00 |  | Cinch Mfg. Corp. part no. 1909 |  |
| R733 | RESISTOR, FIXED, COMPOSITION: 150 ohms | 745-1317-00 | TB712 | TERMINAL BOARD: same as TB702 | 306-2240-00 |
|  | $\pm 10 \%, 1 / 2 \mathrm{w}$ |  | TB713 | TERMINAL BOARD: same as TB706 | 306-0587-00 |
| R734 | RESISTOR, FIXED, COMPOSITION: same as R705 | 745-1401-00 | TB714 | TERMINAL BOARD: same as TB711 | 306-0838-00 |
| R735 | RESISTOR, FIXED, FILM: $1 / 4 \mathrm{w}$ $\mathbf{1 3 , 3 0 0 \mathrm { ohms } \pm 1 \%}$ | 705-7150-00 | $\begin{aligned} & \text { TB715 } \\ & \text { TB716 } \end{aligned}$ | TERMINAL BOARD: same as TB711 <br> TERMINAL BOARD: phenolic w/3 solder-lug | $306-0838-00$ $306-9033-00$ |
| R736 | RESISTOR, FIXED, COMPOSITION: 560 ohms $\pm 10 \%, 1 / 2 \mathrm{w}$ | 745-1342-00 |  | terminals; $11 / 16 \mathrm{in}$. w by $1-1 / 8 \mathrm{in}$. ig; Cinch Mig. Corp. part no. 1520-A |  |
| R737 | RESISTOR, FIXED, COMPOSITION: same as R719 | 745-1422-00 | TB717 |  | 306-0001-00 |
| R738 | RESISTOR, FIXED. COMPOSITION: same as R733 | 745-1317-00 |  | terminals; $11 / 16 \mathrm{in}$. w by 1-1/8 in. ig; Cinels Mig. |  |
| R739 | RESISTOR, FIXED, COMPOSITION: same as R719 | 745-1422-00 |  | Corp. part no. 1525A |  |
| R740 | RESISTOR, FIXED, COMPOSITION: same as R714 | 745-1394-00 | TB718 | TERMINAL BOARD: same as TD704 | 306-0698-00 |
| R741 | RESISTOR. FIXED, FILM: 287 olms $\pm 1 \%$, $1 / 4 \mathrm{w}$ | 705-7070-00 | TB719 | TERMINAL BOARD: same as TB702 | 306-2240-00 |
| R742 | RESISTOR. FIXED. COMPOSITION: same as R720 | 745-1391-00 | TB720 | TERMINAL BOARD: same as TB717 | 306-0001-00 |
| R743 | RESISTOR. FIXED, COMPOSITION: same as R716 | 795-1380-00 | TB721 | TERMINAL BOARD: same as TB704 | 306-0698-00 |
| R744 | RESISTOR. FIXED. COMPOSITION: same as R717 | 745-1352-00 | xFL701 | SOCKET, ELECTRON TUBE: ${ }^{\text {a prong octal tube }}$ | 220-1005-00 |
| $R 745$ | RESISTOR. FIXED, COMPOSITION: same as R716 | 745-1380-00 |  | socket $w_{\text {/ }} /$ steel mitg plate; Amiphenol-Borg |  |
| R746 | RESISTOR. FIXED. COMPOSITION: same as R716 | 745-1380-00 |  | Electronics part no. 88-8TM |  |
| R747 | RESISTOR. FIXED. COMPOSITION: same as R7I6 | 745-1380-00 | XQ701 | SOCKET, TRANSISTOR: 3 contacts spaced on | 352-9903-00 |
| R748 R749 | RESISTOR. VARIABLE: composition; 10.000 ohms $\pm 30 \%$. $1 / 4$ w | $376-4730-00$ $745-1415-00$ | XQ702 | 0.200 in. dia ctrele: Elco Corp. part no. 3307X SOCKET. TRANSISTOR: 4 contacts spaced on | 352-9902-00 |
| R749 R750 | RESISTOR. FIXED, COMPOSITION: same as R712 RESISTOR, FIXED, FILM: 562 ohms $+1 \chi_{\mathrm{k}} 1 / 4 \mathrm{w}$ | $745-1415-00$ $705-7084-00$ | XQ703 | 0:200 in. dia circle; Elco Corp. part no. 3307 SOCKET, TRANSISTOR: same as XQ701 | 352-9903-00 |
| R751 | RESISTOR, FIXED, FILM: 261 ohms $\pm 1 \%$, $1 / 4 \mathrm{w}$ | 705-7068-00 | XQ704 | SOCKET. TRANSISTOR: same as XQ702 | $352-9902-00$ |
| R752 | RESISTOR. FIXED. FILM: same as R751 | 705-7068-00 | XQ705 | SOCKET, TRANSISTOR: same as XQ701 | 352-9903-00 |
| R753 | RESISTOR. FIXED, COMPOSITION: same as R714 | $745-1384-00$ | XQ706 | SOCKET, TRANSISTOR: same as XQ702 | 352-9902-00 |
| R754 | RESISTOR. FIXED. COMPOSITION: same as R714 | 745-1394-00 | XQ707 | SOCKET, TRANSISTOR: same as XQ702 | 352-9002-00 |
| R755 | RESISTOR. VARIABLE: composition; 250 ohms $+20^{\circ}, 1 / 4 \mathrm{w}$ | 376-4725-00 | $\begin{aligned} & \mathrm{XQ708} \\ & \mathrm{XY} 701 \end{aligned}$ | SOCKET, TRANSISTOR: same as XQ702 SOCKET, CRYSTAL: 2 regularly spaced contact | $\begin{aligned} & 352-9902-00 \\ & 202-0082-00 \end{aligned}$ |
| R756 | RESISTOR, FIXED. FILM: 619 ohms $\pm 1 \%$ \%, $1 / 4 \mathrm{w}$ | 705-7086-00 |  | positions, $0.486 \mathrm{in}, \mathrm{c}$ to c ea contact, 0.243 in . | 202-0082-00 |
| R757 | RESISTOR, FIXED, COMPOSITION: same as R736 | 745-1342-00 |  | from center; cadmium plated phosplior bronze or |  |
| R758 | RESISTOR, FIXED, COMPOSITION: same as R736 | 745-1342-00 |  | beryllium copper; Hugh H, Eby part no. 8879 |  |
| R759 | RESISTOR, FIXED, COMPOSITION: same as R723 | 745-1293-00 | Y701 |  | 289-5392-00 |



Figure 4-1. $786 \mathrm{M}-1$ Stereo Generator, Rear View, Resistor Location


Figure 4-3. 786M-1 Stereo Generator, Rear View, Miscellaneous Farts Location


Figure 4-4. 786M-1 Stereo Generator, Front View, Parts Location

SECTION V


# PRODUCTION TEST PROCeDURE: 

FOR
786M-1
STEREO GENERATOR

Prepared by: G. Thomas<br>Reviewed by: C. Dixon

Dated: October 4, 1962
Revised: August 22, 1964

1
COLLINS RADIO COMPANY
CEDAR RAPIDS, IONA

## REVISION RECOHD

TM 1859-5 Reviscd nugust 22. 1964
7.5 Added for clarification.
8. 2 Correction to data sheet.
1.0 SCOPEThis Production Test Procedure applier to the Collins Typo 706M-1 StereoGenerator Part No. 522201400.

- 2.0 REFERTNCE: INFORMATION
2.1 Spocirications
Production Toat Requiromenta, Part Number 569217000Equipmert Sprocifications, Pirt Numbor 568170200
2.2 publicntionsInstruction Jook, Part Number 2D-537
2.3 Drawing ..... "
Schematic Díagram, Stereo Generator $786 \mathrm{M}-1$, Part Number 5538707005Test Jig for Stereo Generator 786M-1, Figure 1
Scope Setup for Phase Adjustment and Left Channel Check, Flgure 2Phase Relationship between Pilot Carrier and L $=-$ R Signal, Figure 3Scope Pattern for Amplitude Adjustrent, Figure 4
Scope Pattern for Left Channel Check, Figure 5
3.0 TEST EqUIPMENT REVUIREDThe following oquipments or their equivalents are required to perform thespecified tests.

1. Oscilloscope, Tektronic Model 545A with Type CA plug-in unit.
2. Oscilloscope, Tektronic Model Ri4503.
3. Oscillator, Hewlett-Yackard Model 200AB
4. Test Jig, Collins Radio, Special, No. 029804700 .
5. Power Suppiy, 28 volts.
6. Distortion and Noise Meter, Hewlett-Packard Model 330D.
7. Frequency Counter, Hewlott-Packard Model 524B/D.
8. VTVM, Hewlett-Packard Model 410B
9. AC VTVM, Hevlett-Packard Model 400D.
10. Resistive Pad to use in place of Pre-Erphasis Network, Figure 1.
11. Output Filter, Figure l.
4.0 TEST CONDITIONS
Unless otherwise specified, all tests shall be performed under thefollowing conditions:
4.1 Power Supply Voltage, Frequency and Dhase
$115 v, 60 \mathrm{cpa}, 1 \varnothing$
28 V DC
4.2 Ambient Tomperature, Humidity and Atmospheric Fressure
Nornal factory ambient.
4.3 Shieldinir and Isolation Rexrirements: None
4.4 OperationaJ. Duty Cycle: Continuous
4.5 Whrm-Uy Period: ..... None

### 5.1. Equigment Correction3

Connect the 7GM-1 Stereo Generutor to the tost jis ohown in figure 1. Use resistive pad in place of Pre-Emphoois Network in tost jig.
5.2 Transfor Reluy Oneration

Switch ON 28 volt power supply. S2 (3tereo) switch on test pinel should operute K 701 rolay in $786 \mathrm{M}-1$.
5.3 Trannistor Installation

Check to sea that tabs on transibtors are adjacent to locating black dots on the chassic.
6.0 INITTAL ADJUSTMENTS
6.138 KC Subcarrier Oscillator Tuning

Connect the RF probe of the VTVM to the collector of Q 702. Adjust R 713 fully clockwise and adjust $L 701$ for maximum output. Record level.
6.238 KC Amplitude Adjustment

Connect Tektronix RM 503 scope across terminals 1 and 2 of $T 701$. Adjust R 713 for 6 volts peak to peak amplitude.
6.3 Carrier Balance

Set pilot carrier to OFF. Set control switch to Stereo position, Set test jiE switch 54 to the $0-15 \mathrm{KC}$ position and Sl (Mode) to $\mathrm{L}=\mathrm{R}$. Adjust the audio oscillator frequency to 400 cps and the level to read $0 \mathrm{v}_{\mu}$ on the $786 \mathrm{M}-1 \mathrm{v}_{\mathrm{j}}$ meter when the meter switch is placed in the NX output position. This sets up a.reference level ( $90 \%$ modulation) at the output of the stereo generator to which the residual carriers will later be referred. Measure this voltage by connecting the AC VTVM to the output termindls of the test jlg. The carrier level readings will later be recorded in db and trill be referenced to this Odb level. Remove modulation by turning the test jig switch, $S 1$, to the OFF position. Set test jig switch, $S^{4}$, to the 23-53 KC position and observe the residual 38 KC carrier level as read on the audio VTVA. This readine: must be at least 60 db below the preceedine reference level. It will not be this far down however until the balancing procedure which follows is completed. Turn the test jie swivch, 54 , to the 76 KC position and observe the residual 76 KC carrier level as read on the audio VTVM. It must be 70 db below the reference level. It, too, will be high until the following balancing procedure is completed.
Set test jig switch $\mathrm{S}_{4}$ to MX OUT. Adjust R 703 and $R 726$ for minimum indication on the AC VTVM. Recheck the 38 KC and 76 KC carrier levels as explained in the preceeding paragraphs. If these carrier levels are still too great, capacitors must be selected to balance then to a lower level. The 38 KC ccmponent can be reduced by step (a) and the 76 KC component can be recuced by step (b).

## 6. 3 C.arrior EDlance (Continued) <br> Sten (a)

Add a $0-50 \mathrm{pf}$ trimmer from pin 5 of f 701 to eround. Adjust thic trimmer, $R 705$, and $R$ \%2b for minimum output from tho tost jis whon gly is in the MX OU' porsition. If the trimmer goes to minimum capacitance, disconnect it from pin 5 and connect it to pin 3 of $T$ 70l. Repeat adjustment.
Step (b)
Switch 54 to 76 KC . Place your fingor in turn on the junction between CR 702 B and R 729, CR 702 D and $\mathrm{R} 728, \mathrm{CR} 702 \mathrm{~A}$ and R 727 , and CR 702 C and R 725 . Obscrvef which function causes a reduction in the 76 KC output. Connect a $0-50$ pf capacitor from thits point to ground. Suitch 54 to MX OU'r and tune the two trimers, R703, and R726 for minimum carrier output as before.
The last step is to check the level of the 38 KC and 76 KC carriers individually as explained before and record these decibel readings. The 38 KC carrier must be at least 60 db below the reference level ( $90 \%$ modulation) and the $76: \mathrm{KC}$ carrier must be at least 70 db below the reference level.

NOTE 1 Use a heat sink to install the trinaers because heat wild effect the diodes and the balance of the circuit.

NOTE 2 Interforence from near by transmiters may make it difficult to make this adjustrient but will not effect the circuit when balanced. .

### 6.4 Pilot Carrier Level and Frecuency

Connect AC VTV南 to TP 702. Set Pilot Carrier to $O N$ and Control to STEREO ON. Adjust $P$ P 78 for a readine of .009 V RMS. Connect frequency counter to TP 701 and record pilot carrier frequency. If frequency is off and crystal is changed repeat paragraph 6.3.

### 6.5 Locked Oscillator Phase

Connect the Tektronix 545A scope to the 786m-1 stereo generator and adjust as shown in figure 2. Set test jig to $L=-R$ and the audio: oscillator to 1000 cps and 7.8 VRMS. Hith PILOT CARPIER at OFF adjust L 704 until both traces are statimary and exact coincidence of the zero crossings of the 19 KC pilot carrigr and the $\mathrm{L}-\mathrm{R}$ signal is obtained as shown in Figure 3a. Expand the sweep to 5 X and adjust the horizontill position knob to check two points of coincident zero crcssing.

### 6.6 Channel Identification

Connect the Tektronix 545 A scope to the $786 \mathrm{M}-1$ stereo generator and adjust as shown in Figure 2. Set test jig to $\mathcal{L}$. If wiring is correct the 38 KC subcarrier and stereo sigual envelope will have the phase relation ship shown in figure 5b. If reversed the phase relationship will be as shoum in 5a.

### 6.7 Auplitude Correction

Set the audio oscillaton to 5000 cps . Switch Pilot Carrier OrF. Connoct the lextronix Fin 503 oscjliloscone to TP 702. Set the teat jig to $L$ and adjust the aulio oscillator for a $250 \mathrm{mv} P / \mathrm{P}$ sirnal.

### 6.7 Ampliture Correction (Conthued)

 in ligure $4^{2}$ Ghock over the frequency rongo 50 to 15000 cpa with scopo horlzontel acrottivity set to $10 \mathrm{~W} / \mathrm{Chi}$. Reyeat check ovor froquency reme with tho tost jig seti to $R$. Uso appropriate sottines of weop tirgecm an froquency io varfed. road and rocond tho musimu


### 6.8 Hotor Ragintor Soleciton

Suitch to tho 30 KC Osc poaitiono If tho meter readjen ia lese than 75 cr eroater than full neale, molect $P 720$ to brjue the readine within those Idinitt. Susich tho 786 moter sutith to 78 KC ETF porition and balect 1276 for a woter roadide betwon. 75 and fuJd ocalc.
6.9 moter jovelas
 osc. Rond and recond level for cach position.

## 7.0 mas bequaparine

### 7.1 Previminocy Rects

preliminary terts as outlined in coction 5. .
7.2 Indelat Adystrents

Initial adinomonto as outhmed in ooction 6 。

### 7.3 Ging

Replaco resistivo pads with premempagio noworis. Set test jig sutch, S., to $L=$ R. Adjust eudio ocicillator frequency to 400 cps and amplitudo to read 0 on the Vij moter when tho roter suitch is set to $I$ or $R$ endio. Connect the Tektronix RA503 oscilloscone to TP7O2. Hith PILOT CADRIER oit reod and record tho peak-to-psals amplitude as read on the oscijno-

 amplitude at T TP 702。
 poak-to-pook emplikude at ropoz. Set compot eritch to $R$ ama towt jict swith, S.l, to R. Read and secord, poak-io-peak amplitude at mp 702. Set the test jig sultch, S? (otoreo), to the storeo position. The oscilloccope pattern should change foom a einucoid to the conp?a, right only, btoreo presentation that :ios observed in tion Amolitude Correction ádjustrani (par. 6.7). Fonove pre-tuphasis netuost sud replaco with resiotive pads.
7.4 Frequency Rosponse

Set the cost jics to $L=\pi$. fidjust the oscijuator outpat to road o iJ on the VU noter whom the HETh sixtcin is set to L or $I$ audio. poad and recond the output for oscijlator frequencios of 50,1000 , and 15000 cps .

## 7.5 crome Tblk



## Limanio Distortion

Connect the outpat of the stereo cenerator to tho test fig J.e FIUTM
 torminalo. Set toat 括g to L a R and odjuct oudio oscillator amplitnde to read $O$ on the VU ineter when tho meter suiteh is set to $L$ or $\mathbb{R}$ audio. Hoasure harmonic distortion at $50,2000,5000 \mathrm{c} / \mathrm{se}$.




嘘


FRACTIONS DECIMALS ANGLES
$=1 / 64=003=1$



The 7203/4CX25()B and 7204/4CX250F are compact, forced-air cooled, externalanode radial-beam tetrodes with a maximum plate dissipation rating of 2.50 watts and a maximum input-power rating of 500 watts. The $7203 / 4 \mathrm{CX} 250 \mathrm{~B}$ is designed to operate with a heater voltage of 6.0 volts, while the $7204 / 4 \mathrm{CX} 250 \mathrm{~F}$ is designed for operation at a heater voltage of 26.5 volts. Otherwise, the two tube types have identical characteristics.

Both tube types are of all-ceramic-and-metal construction and are recommended for use in equipments of new design.

## general CHARACTERISTICS

## ELECTRICAL



## RADIO-FREQUENCY POWER AMPLIFIER OR OSCILLATOR

Class.C Telegraphy or FM Telephony<br>[Key-down conditions!<br>MAXIMUM RATINGS<br>D-C PLATE VOLTAGE<br>- 2000 MAX. VOLTS<br>-C SCREEN YOLTAGE - - - . 300 MaX. Volts<br>D.C GRID Voltage<br>- 250 MAX. VOLTS<br>D.C Plate current<br>250 MAX. MA<br>plate dissipation<br>250 MAX. WATTS<br>SCREEN DISSIPATION - . . . 12 MAX. WATTS<br>GRID DISSIPATION . . - . 2 MAX. WATTS


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## PLATE-MODULATED RADIO-FREQUENCY

## AMPLIFIER

| Class-C Telophony (Carrier conditions) |  |  |
| :---: | :---: | :---: |
| MaXimum ratings |  |  |
| d.C plate yoltage | - . . | 1500 MAX. VOLTS |
| D.C Screen voltage |  | 300 MAX V VOLTS |
| d.C GRID Voltage | - - - | -250 MAX. VOLTS |
| D.C Plate current |  | 200 Max. Ma |
| Plate dissipation |  | 165 MAX. WATTS |
| SCREEN DISSIPATION |  | 12 MAX. WATTS |
| GRID DISSIPATION |  |  |


| D.C Plate Yoltaqe | - - | . - | 500 | 1000 |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| D.C Screen Vollage | . | - - | 250 | 250 | 250 volts |
| D-C Grid Vollaqe | - | - - | -100 | -100 | - 100 volts |
| D-C Plate Current | - - | - - | 200 | 200 | 200 ma |
| D.C. Screen Current* | - • | - - | 31 | 22 | 20 md |
| D.C Grid Current* | . . - | - - | 15 | 14 | 14 ma |
| Peak R-F Grid Input | Vollage* | - - | 118 | 117 | 117 volts |
| Driving Power* | - . . | - - | 1.8 | 1.7 | 1.7 watts |
| Plate Input Power |  | - - | 100 | 200 | 300 waths |
| Plata Oulput Powor | - - | - - | 60 | 145 | 235 waits |

## AUDIO-FREQUENCY AAPLIFIER OR MODULATOR

Class-A日
MAXIMUM RATINGS (Par tuba)
D-C PLATE YOLTAGE . . . . 2000 MAX. VOlts
D-C SCREEN Voltage . - - - 400 max. volts
D.C PLATE CURRENT . . . - 250 MAX. MA

PLATE DISSIPATION . . . . 250 maX. WATTS
SCREEN DISSIPATION - . . - 12 MAX. WATTS
GRID DISSIPATION - - . . 2 MAX. WATTS

TYPICAL OPERATION (Sinusoidal wave, two tubes unless noted]
D-C Plate Yoltagn . . . . . 100015002000 volls
D-C Screen Voltage - - - - . 350 350 350 volts D.C Grid Yoltdge' - . . . . 55 - 55 - 55 volts Zero-Signal D-C Plato Current . . . $200200 \quad 200 \mathrm{ma}$ May-Signal D.C Plate Current . . . $500 \quad 500 \quad 500 \mathrm{ma}$ Max-Signal D.C Screen Current . . . 201610 md Eliective Load, Plate to Plale . . . $15006200 \quad 9500$ ahms Peak A.F Grid Input Yollage (per tube)* . $50 \quad 50 \quad 50$ volis Driving Power . - . . . . . 0 0 0 watls Max-Signal Plate Oulput Power . . . 240 430 600 watts -Approximale values.
'Adjust grid bias to oblain listed zero-signal plate current.

## RADIO-FREQUEMCY LINEAR APAPLIFIER

Class-AB, (Carriar canditions)
MAXIMUM RATINGS
D-C PLATE VOLTAGE - . . . 2000 MAX. VOLTS
D.C SCREEN VOLTAGE . . . - 400 MAX. VOLTS
D.C PLATE CURRENT . . . . 250 MaX. MA PLATE DISSIPATION - . - - 250 MAX. WATTS
SCREEN DISSIPATION . - - . 12 MAX. WATTS
GRID DISSIPATION - . . - 2 MAX. WATTS

TYPICAL OPERATION (Frequencies up to 175 Mc )
D.C Plate Voltage . . . . . . 1000
C. . . . . $\quad .1000$ late Voltage 2000 volls D.C Screen Voltage - . . . . . $350 \quad 350 \quad 350$ volis D.C Grid Yoltage' . . . . . . -55 -55 -55 volts

Zero-Signal D.C Plate Current
D-C Plata Current
D.C Sereen Current* ${ }^{*}$ - - 150

Plate Output Power . . . . . $\quad-\quad-\quad 25 \quad 25 \quad 25$ valts
Approximate values
'Adjust grid bias to obtain listed zero-signal plate current.

## RADIO-FREQUENCY LINEAR AAPLIFIER

Class-AB, (Single-Sideband Suppressed-Carrier Oparation) maximum ratings
D-C PLATE VOLTAGE - - - - 2000 MAX. VOLTS
D-C SCREEN VOLTAGE . . . . 400 MAX. VOLTS
D.C plate current - . . . 250 max. Ma

PLATE DISSIPATION . . . . 250 MAX. WATTS
SCREEN DISSIPATION - . . . 12 MAX. WATTS
GRID DISSIPATION - - . . 2 MAX. WATTS


[^0]
## APPLICATION

## MECHANICAL

Mounting-The 4 CX 250 B and 4 CX 250 F may be operated in any position. An Eimac Air-System Socket, SK-600 series, or a socket having equivalent characteristics, is required. Sockets are available with or without built-in sereen capacitors and may be obtained with either grounded or ungrounded cathode terminals.
Cooling-Sufficient forced-air cooling must be provided for the anode, base seals, and body seals to maintain operating temperatures below the rated maximum values. Air requirements to maintain anode core temperatures at $200^{\circ} \mathrm{C}$ with an inlet air temperature of $50^{\circ} \mathrm{C}$ are tabulated below. These requirements apply when a socket of the Eimac SK-600 series and an Eimac SK-606 chimney are used with air flow in the base to anode direction.

|  | SEA LEVEL |  | 10.000 FEET |  |
| :---: | :---: | :---: | :---: | :---: |
| Plate <br> Oisspation <br> (Wans) | Air Flaw <br> (CFM) | Pressure <br> Orop (Inches <br> of Water) | Air Flow <br> (CFM) | Pressura <br> Drop (Inchos <br> of Wator) |
| 200 | 5.0 | 0.52 | 7.3 | 0.76 |
| 250 | 6.4 | 0.82 | 9.3 | 1.20 |

The blower selected in a given application must be capable of supplying the desired airflow at a back pressure equal to the pressure drop shown above plus any drop encountered in ducts and filters. The blower must be designed to deliver the air at the desired altitude.

At 500 Mc or below, base-cooling air requirements are satisfied automatically when the tube is operated in an Eimac Air-System Socket and the recommended air-flow rates are used. Experience has shown that if reliable long-life operation is to be obtained, the cooling air-flow must be maintained during standby periods when only the heater voltage is applied to the tube. The anode cooler should be inspected periodically and cleaned when necessary to remove any dirt which might interfere with effective cooling.
Vibrafion - These tubes are capable of satisfactorily withstanding ordinary shock and vibration, such as encountered in shipment and normal handling. The tubes will function well in automobile and truck mobile installations and similar environments. However, when shock and vibration are expected to exceed approximately 5 g units, it is suggested that the Eimac 4CX300A or 4CX250R be employed.

## ELECTRICAL

Heater - The rated heater voltage for the 4 CX 250 B and 4 CX 250 F is 6.0 volts and 26.5 volts, respectively, and the voltage should be maintained as closely as practicable. Short-time changes of $\pm 10 \%$ will not damage the tube, but variations in performance must be expected. The heater voltage must be maintained within $\pm 5 \%$ to minimize these variations and to obtain maximum tube life.

At frequencies above approximately 300 megacycles, transit-time effects begin to influence the cathode temperature. The amount of driving power diverted to heating the cathode by back-bombardment will depend upon frequency, plate current, and driving power. When the tube is driven to maximum input as a "straight-through" class-C amplifier, the heater voltage should be reduced according to the table below:

| Frequency, Me | $4 C \times 250 \mathrm{~g}$ | $4 \mathrm{CX250F}$ |
| :---: | :---: | :---: |
| 300 and lower | 6.00 volis | 26.5 volts |
| 301 to 400 | 5.75 volis | 25.5 volts |
| 401 to 500 | 5.50 volis | 24.3 volts |

Cathode Operation - The oxide-coated unipotential cathode must be protected against excessively high emission currents. The maximum rated d-c input current is 200 milliamperes for plate-modulated operation and 250 milliamperes for all other types of operation except pulse.

The cathode is internally connected to the four even-numbercd base pins, and all four of the corresponding socket terminals should be used to make connection to the external circuits. At radio frequencies it is important to keep the cathode leads short and direct and to use conductors with large areas to mini-
mize the inductive reactances in series with the cathode leads.

It is recommended that rated heater voltage be applied for a minimum of 30 seconds before other operating voltages are applied. Where the circuit design requires the cathode and heater to be operated at different potentials, the rated maximum heater-tocathode voltage is 150 volts regardless of polarity.
Control-Grid Operation - The maximum rated d-c grid bias voltage is -250 volts and the maximum grid dissipation rating is 2.0 watts. In ordinary audio and radio-frequency amplificrs the grid dissipation usually will not approach the maximum rating. At operating frequencies above the 100 -megacycle region, driving-power requirements for amplifiers increase noticeably. At 500 megacycles as much as 20 watts of driving power may have to be supplied. However, most of the driving power is absorbed in circuit losses other than grid dissipation, so that grid dissipation is increased only slightly. Satisfactory 500 -megacycle operation of the tubes in a stable "straight-through" amplifier is indicated by grid-current values below approximately 15 milliamperes.

The grid voltage required by different tubes may vary between limits approximately $20 \%$ above and below the center value, and means should be provided in the equipment to accommodate such variation. It is especially important that variations between individual tubes be compensated when tubes are operated in parallel or push-pull circuits, to assure equal load sharing.

The maximum permissible grid-circuit resistance per tube is 100,000 ohms.

Screen-Grid Operation - The maximum rated power dissipation for the screen grid is 12 watts, and the screen input power should be kept below that level. The product of the peak screen voltage and the indicated d-c screen current approximates the screen input power except when the screen current indication is near zero or negative.

In the usual tetrode amplifier, where no signal voltage appears between cathode and screen, the peak screen voltage is equal to the $\mathrm{d}-\mathrm{c}$ screen voltage.

When signal voltages appear between screen and cathode, as in the case of screen-modulated amplifiers or cathode-driven tetrode amplifiers, the peak screen-to-cathode voltage is the sum of the d-c screen voltage and the peak a-c or r-f signal voltage applied to screen or cathode.

Protection for the screen should be provided by an over-current relay and by interlocking the screen supply so that plate voltage must be applied before screen voltage can be applied.

The screen current may reverse under certain conditions and produce negative current indications on the screen milliammeter. This is a normal characteristic of most tetrodes. The screen power supply should be designed with this characteristic in mind so that the correct operating voltage will be maintained on the screen under all conditions. A current path from screen to cathode must be provided by a bleeder resistor, gaseous voltage regulator tubes, or an electron tube shunt regulator connected between screen and cathode and arranged to pass approximately 15 milliamperes per connected screen. An electron tube series regulator can be used only when an adequate blecder resistor is provided.

Self-modulation of the screen in plate-modulated tetrode amplifiers using these tubes may not be satisfactory because of the screen-voltage screen-current
characteristics. Screen modulation from a tertiary winding on the modulation transformer or by means of a small separate modulator tube will usually be more satisfactory. Screen-voltage modulation factors between 0.75 and 1.0 will result in $100 \%$ modulation or platemodulated r-f amplifiers using the 4 CX 250 B or 4CX250F.

Plate Operation - The maximum rated plate-dissipation power is 250 watts. In plate-modulated applications the carrier plate-dissipation power must be limited to 165 watts to avoid exceeding the plate dissipation rating with $100 \%$ sine wave modulation. The maximum dissipation rating may be exceeded for brief periods during circuit adjustment without dam. age to the tube.

Multiple Operation - Tubes operating in parallel or push-pull must share the load equally. It is good engineering practice to provide individual metering and individual adjustment of bias or screen voltage to equalize the inputs.

Where overload protection is provided, it should be capable of protecting the surviving tube(s) in the event that one tube fails.

UHF Operasion-The 4CX250B and 4CX250F are suitable for use in the UHF region. Such operation should be conducted with heavy plate loading, minimum bias, and the lowest driving power consistent with satisfactory performance. It is often preferable to operate at a sacrifice in efficiency to obtain increased tube life.
Special Applications - If it is desired to operate these tubes under conditions widely different from those given here, write to Power Grid Tube Marketing, Eitel-McCullough, Inc., San Carlos, California, for information and recommendations.


CONTACT SURFACE

| DIMENSIONAL DATA |  |  |  |
| :---: | :---: | :---: | :---: |
| REF | MIN. | max. | nom. |
| A | 2.324 | 2.464 |  |
| B | 1.610 | 1.540 |  |
| C | 1.810 | 1.910 |  |
| 0 | . 750 | . 910 |  |
| E | . 710 | . 790 |  |
| $F$ |  | 1.406 |  |
| G | . 187 |  |  |
|  | 8ASE: 88-236 |  |  |
| H | (JEDEC OESIGNATION) |  |  |
| $J$ | . 559 | . 573 |  |
| K | 240 | 280 |  |

DIMENSIONS IN INCHES

$$
\begin{aligned}
& \text { PIN NO I. SCREEN GRID } \\
& \hline \text { PIN NO 2. CATHODE } \\
& \hline \text { PIN NO. 3. HEATER } \\
& \hline \text { PIN NO.4. CATHODE } \\
& \hline \text { PIN NO. 5. I.C. DO NOT USE FOR } \\
& \hline \\
& \text { PIN NO.6. CATHOOE } \\
& \hline \text { PIN NO. } 7 . \text { HEATER } \\
& \hline \text { PIN NO. B. CATHODE }
\end{aligned}
$$




## $4 C \times 250 B$

# unit instructions 

## B830-1 <br> 250-Watt FM <br> Power Amplifier

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## SECTION I GERERAL DESCRIPTION

### 1.1 PURPOSE OF INSTRUCTION BOOK.

Unit Instructions TD-538 provides information about B830-1 250-Watt FM Power Amplifier. Information which is furnished covers a general description of the equipment, principles of operation, maintenance procedures, and a parts list.

### 1.2 PURPOSE OF EQUIPMENT.

The B830-1 250-Watt FM Power Amplifier can be used for continuous monaural or multiplex and SCA FM broadcast service on a single frequency in the range from 88 to 108 megacycles withanexciter input of 10 watts and an output power of 250 watts.

### 1.3 DESCRIPTION OF EQUIPMENT.

### 1.3.1 PHYSICAL DESC RIPTION.

The B830-1 250-Watt FM Power Amplifier, shown in figure $1-1$, is contained in a single cabinet that is 38 inches wide, 76 inches high, 27 inches deep, and weighs approximately 596 pounds. All B830-1 operating controls are located behind the doors on the front of the cabinet. The filament and plate on-off controls and four monitoring meters are located at the top front of the cabinet. The meters may be observedeasily while operating the tuning controls. The B830-1 uses three tubes, all of which are accessible from the front of the transmitter. The bottom front of the B830-1 cabinet is removable to allow access to components on the bottom of the inside panel.
Large doors at the upper rear of the cabinet (see figure $1-2$ ) allow access to the upper part of the B830-1 for servicing and maintenance. Operating personnel are protected by both electrical and mechanical interlocks on the rear doors and panel. These interlocks remove the B830-1 plate voltage and ground the high-voltage circuits when the doors are opened or the panel is removed. The power amplifier plate-tuning cavity is located in an interlocked compartment at the front of the B830-1.
Inside the $\mathrm{B} 830-1$, heavy iron-core components are at the bottom of the cabinet. A standard 19 -inch rack is provided for mounting the 10 -watt exciter within the confines of the $\mathrm{B} 830-1$. Room also is provided on the standard 19 -inch rack for the mounting of SCA generators. An optional harmonic filter, which can be connected to the B830-1 output, is suspended from the top of the cabinet.
Cooling air for the B830-1 is drawn through a permanent air filter at the rear of the cabinet by a lowspeed, high-volume fan and exhausted through a shielded opening in the top of the cabinet. A single pressure blower supplies cooling air directly to the power amplifier tube.


Figure 1-1. B830-1 250-Watt FM Power Amplifier, Over-all View

### 1.3.2 ELECTRICAL DESCRIPTION.

The B830-1 250-Watt FM Power Amplifier consists of a single, air-cooled, power amplifier tube capable of being driven to full power by a 10 -watt exciter. All


Figure 1-2. B830-1 250-Watt FM Power Amplifier, Rear View with Bottom Panel Removed
associated power supply and control circuitry is included for operation of the B830-1. The B830-1 r-f input impedance is 50 ohms nominal, unbalanced. The B830-1 output power is at least 250 watts over the frequency range of 88 to 108 megacycles into a $50-$ ohm load, with an swr not exceeding 2:1.

Line power input required is 50 or 60 cycles, single phase, with primary taps on all power transformers to compensate for line voltage variations from 200 to 250 volts. (For $50-\mathrm{cps}$ operation, a special blower is available on request.) Circuit breakers in the input side of the line are provided for primary circuit protection. The control circuits and the final amplifier filament are fused. A time delay relay provides protection of the power amplifier tube during warmup. Remote control can be exercised over
filament-on, filament-off, plate-on, and plate-off functions of the power amplifier. Instruction books covering the exciters and power amplifiers used in conjunction with B830-1 250-Watt FM Power Amplifier are listed in table 1-1.

TABLE 1-1
ASSOCIATED EQUIPMENT INSTRUCTION BOOKS

| ASSOCIATED <br> EQUIPMENT | INSTRUCTION BOOK <br> NUMBER |
| :--- | :---: |
| A830-2 10 W Wide-Bandi <br> FM Broadcast Exciter <br> $786 \mathrm{M}-1$ Stereo <br> Generator <br> E830-1 5-Kw FM Power <br> Amplifier | TD-536 |

### 1.4 EQUIPMENT SUPPLIED.

Table 1-2 lists equipment that is supplied as part of B830-1 250-Watt FM Power Amplifier.

TABLE 1-2 EQUIPMENT SUPPLIED

| EQUIPMENT | COLLINS <br> PART NUMBER |
| :---: | :---: |
| B830-1 250-Watt FM Power <br> Amplifier | $549-2008-00$ |

### 1.5 EQUIPMENT REQUIRED BUT NOT SUPPLIED.

Table 1-3 lists equipment that is required for operation of B830-1 250-Watt FM Power Amplifier but not supplied as part of the power amplifier.

TABLE 1-3
EQUIPMENT REQUIRED BUT NOT SUPPLIED

| EQUIPMENT | COLLINS <br> PART NUMBER |
| :---: | :---: |
| A830-2 10 W Wide-Band <br> FM Broadcast Exciter <br> 786M-1 Stereo Generator <br> (for use with A830-2 only) | $522-2714-00$ |

### 1.6 ACCESSORY EQUIPMENT.

Table 1-4 lists accessory equipment that is available for use with B830-1 250-Watt FM Power Amplifier.

TABLE 1-4
ACCESSORY EQUIPMENT

| EQUIPMENT | COLLINS <br> PART NUMBER |
| :---: | :---: |
| 250-Watt/1-Kw Harmonic Filter <br> (used only if the B830-1 is fed <br> directly to an antenna and not to <br> a higher power amplifier) | $549-2010-00$ |

### 1.7 EQUIPMENT SPECIFICATIONS.

### 1.7.1 MECHANICAL.



### 1.7.2 ELECTRICAL.



### 1.8 TUBE AND SEMICONDUCTOR COMPLEMENT.

Table 1-5 lists the tube and semiconductor complemeat supplied as part of B830-1 250-Watt FM Power Amplifier.

TABLE 1-5
TUBE AND SEMICONDUCTOR COMPLEMENT

| QUANTITY | TYPE | FUNC TION |
| :---: | :--- | :--- |
| 1 | 4CX250B/7203 | R-f power amplifier |
| 2 | OD3 | Voltage regulation |
| 2 | 4JA421EM20AB1 | Plate and screen <br> voltage rectifiers |

## SECTION II PRINCIPLES OF OPERATION

### 2.1 GENERAL.

The B830-1 250-Watt FM Power Amplifier consists of a power amplifier and associated control circuitry for the amplification of 10 watts of exciter drive to 250 watts of r-f power. The B830-1 output can drive a higher power amplifier or can be fed through a harmonic filter to an antenna.

Refer to figure 2-1. The 230 volts a-c is fed to a control circuit power transformer which reduces the input voltage to 115 volts a-c for use by the blower and control circuits. A second transformer located within the control circuitry reduces the 115 volts a-c to 6.3 volts for use by the power amplifier filament. The 230 volts a-c also is fed to a transformer which supplies the power amplifier plate and screen voltages. The primary power input to the plate power supply is controlled by the plate contactor.

The control circuits provide cabinet interlocks for protection of personnel from all high voltage, local or remote filament-on and filament-off controls, local or remote plate-on and plate-off controls, and time delay to prevent the application of high voltage before the power amplifier filament has heated sufficiently. Provisions also are available within the control circuitry for connection to the control circuitry of a higher power amplifier. The higher power amplifier will then control the B830-1 plate and filament power.

Exciter input power is applied directly to the power amplifier where it is raised to 250 watts of $r$-f power. The power amplifier consists of a forced-air-cooled ceramic tetrode, V201. The plate of V201 is connected to a tuned cavity ( $\frac{\lambda}{4}$ coaxial line resonator).
The output from the tuned cavity then can be fed to a higher power amplifier, or to a harmonic filter and antenna if higher power amplification is not necessary. A sample of the power amplifier output is taken from the plate-tuned cavity for monitoring purposes.

Metering circuits are provided for the power amplifier plate current and voltage, screen current and voltage, grid current, and output power.

### 2.1.1 CONTROL CIRCUTS.

Refer to figure 2-2. The primary purpose of the control circuitry is to provide filament and plate on and off control. The power amplifier filament is turned on in the following manner: When momentary FILAMENT ON switch S 112 is pressed, a ground is placed on filament control relay $K 301$. As 115 volts a-c is present on terminal 2 of relay K301 from 115 -volt a-c supply transformer T301 the filament control relay will be energized. This closes contacts 3 and 4 , holding the relay in the energized position after momentary FILAMENT ON switchS112 is released. The greenfilament indicator lamp, DS301, will light. When filament control relay K301 is energized, relay contacts 6 and 7 will


Figure 2-1. B830-1 250-Watt FM Power Amplifier, Block Diagram



Figure 2-3. B830-1 250-Watt FM Power Amplifier
Power Amplifier Circuits, Simplified
close, starting the cabinet fan and power amplifier blower B301 and supplying 115 volts a-c to the 10 watt exciter. When blower B301 comes up to speed, air interlock switch S 34 will close, applying 115 volts a-c to filament transformer T302 and time delay relay K303. After approximately 30 seconds (time for power amplifier filament V201 to warm up), time delay relay K303 will be energized, closing contacts 3 and 4, supplying 115 volts a-c to plate contactor relay K304.

The plate supply now can be turned on by pressing the momentary PLATE ON switch (providing all interlocks are closed). Pressing the momentary PLATE ON switch places a ground on plate control relay K302. Relay K302 then will energize, closing contacts 3 and 4, holding K302 in the energized position. Contacts 9 and 10 will also close energizing plate contactor K304 and lighting the red plate indicator lamp. When the plate contactor closes, the primary a-c supply will be applied to plate and screen supply transformer T303. The plate power supply then will furnish approximately 2000 volts d -c to power amplifier tube v201.

Momentary Filament OFF switch S111 removes power from the control and power amplifier circuits. It accomplishes this by opening the 115-volt a-c lead to filament control relay K301, and plate control relay K302. This de-energizes plate contactor K304 and blower interlock S314, removing filament, plate, and screen voltages. The time delay relay will immediately reset for the next turn-on procedure.
Momentary PLATE OFF switch S114 removes only plate and screen voltage from power amplifier V201. This is accomplished by momentarily opening the ground lead of plate control relay K302, which deenergizes plate contactor K304, removing 230 -volt a-c power from the plate supply.
Plate overload protection is provided by overload relay K305. As the plate current increases, the voltage across R303 will increase until overload relay K305 energizes, opening the ground lead of plate control relay K302, and removing plate voltage. Potentiometer R304 adjusts the point at which excess plate current will activate overload relay K 305.

### 2.1.2 POWER AMPLIFIER CIRC UITS.

Refer to figure 2-3. The power amplifier consists of a forced-air-cooled tetrode amplifier, working into a tuned cavity over the standard frequency-modulated broadcast band of 88 to 108 megacycles.

The power amplifier plate and screen voltages are obtained from a full-wave, semiconductor-type, rectifier circuit. Transformer T303 increases the 200- to 250 -volt, single-phase, 50 - or 60 -cps primary input to approximately 4000 volts rms ac ross terminals 8 and 10 (T303 secondary winding, 2000 volts each side of center tap). The primary winding of transformer T303 has six adjustable taps to compensate for line voltages from 200 to 250 volts. The output voltage from the secondary winding of T303 is fed to a

Conventional full-wave rectifier consisting of diodes CR301 and CR302, inductors L301 and L302, and capacitors C301 and C302. Resistor R320 and capacitors C 303 and C304 form a transient suppressor which eliminates the transient voltages formed when power is first applied to transformer T303, and when power is switched off. The 2000 -volt d-c output from the power supply is fed through P.A. PLATE CURRENT meter M303 through an r-f filter consisting of C210, C217, and L202 to the plate of V201. Plate voltage and plate current are read directly from P.A. PLATE VOLTAGE meter M302 and P.A. PLATE CURRENT meter M303 respectively.

The screen voltage is obtained from the 2000 -volt plate supply. The plate supply is fed to a combination bleeder resistor and voltage divider consisting of resistors R305, R308, and R309. Voltage regulator tubes V301 and V302 are placed in parallel with resistors R308 and R309 to form a regulated 300 -volt supply which is supplied to the screen grid of V201. Capacitor C212 shunts any r-f energy present on the screen grid to ground

Power amplifier V201 is a cathode and grid leak biased, class $C$ operated tetrode. The control grid circuit of V201 consists of a parallel-tuned resonant tank circuit (L201 and C213) with swamping resistor R201 placed in parallel to provide a low-impedance, broadband load to the control grid. Control grid tuning is accomplished by capacitor C213.

The cathode circuit of V201 contains an output power adjusting potentiometer which raises or lowers the cathode resistance. As V201 is cathode biased, adjusting POWER OUTPUT ADJUST potentiometer R317 changes the cathode bias, as well as the screen potential, which controls the power output of V201. Capacitors C201 through C204 form the cathode bypass.

The plate cavity is formed by a short section of coaxial transmission line resonating with the plate capacity of V201 and plate tuning capacitor C209. The coaxial transmission line is roughly tuned initially by adjusting a shorting plate which lengthens or shortens the coaxial transmission line. Final plate tuning is accomplished by C209. The output coupling network formed by C208 and L203 is connected to the plate transmission line adjacent to the plate of V201 for correct impedance matching. Inductance L203 and capacitor C209 act as an L-section, lowpass filter for frequencies above 130 megacycles to provide additional harmonic suppression. A monitor output is connected directly to the plate cavity for use by the station program monitor.

MULTIMETER M301 is located on the front panel of the power amplifier to enable the station operator to monitor screen voltage, screen current, and grid current. The multimeter functions are selected by the multimeter switch located within the power amplifier cabinet. Screen voltages are determined by connecting MULTIMETER M301 across the regulated screen


Figure 2-4. B830-1 250-Watt FM Power Amplifier, Control Locations
supply with switch S307. Meter multiplier resistor R322 limits the current through the multimeter for calibration purposes. Screen current is determined by the voltage differential across shunt R307. This voltage is fed to the multimeter when switch S307 is in the proper position. Grid current is determined by the voltage differential across shunt R316. This voltage is also available to the multimeter when S 307 is in the proper position.

### 2.2 CONTROL FUNCTIONS.

The following paragraphs describe all the functions of controls in B830-1 250-Watt FM Power Amplifier. Refer to figure 2-4 for control locations.
The controls located directly on the front panel under the meters include the FILAMENT ON, FILAMENT OFF, PLATE OFF, and PLATE ON switches. The
filament indicator and plate indicator lamps are placed in line with the above mentioned controls. The FILAMENT ONswitch, S112, energizes the power amplifier filament, the poweramplifier, and cabinet blowers, and supplies power to the 10 -wattexciter. The FILAMENT OFF switch, S111, de-energizes all transmitter circuits. The PLATE ON switch, S113, energizes the plate power supply, starting the 250 -watt power amplifier. The PLATE OFF switch, S114, removes plate and screen voltage. The green filament indicator lamp, DS301, comes on when the FILAME NT ON switch is pressed and indicates that voltage is available to the filament control relay and the PA blower. The PA hlower will activate the PA blower air interlock which energizes the power amplifier filament. The red plate indicator lamp, DS302, indicates the plate voltage has been applied to the power amplifier.

The following controls are located directly under the left front door on the power amplifier panel. The POWER OUTPUT ADJUST potentiometer, R317, adjusts the power amplifier bias and screen potential, thus changing the output power. The WATTMETER switch, S308, connects the R.F. WATTMETER to either the reflected power or forward power section of the directional coupler. The WATTMETER switch normally is left in the FORWARD 400W position. The MULTIMETER switch, S307, selects either screen voltage, screen current, or grid current for MULTIMETER M301. Table 3-1 lists the MULTIMETER switch positions and typical indications for each of the three meter circuits.

The following controls are located directly behind the left front door on the power amplifier compartment. The OUTPUT COUPLING control, C208, adjusts the coupling of the load to the plate cavity. The PLATE TUNING control, C209, tunes the plate cavity to resonance and is set very near the minimum indication on P.A. PLATE CURRENT meter M303. At this point the power output should be at the peak as indicated on R. F. WATTMETER M304. The GRID TUNING control, C213, tunes the grid tank and is set for maximum indication on MULTIMETER M301 with the MULTIME TER switch, S307, set to GRID FS 40 MA.

The following controls are located on the power panel directly behind the bottom front panel of the power
amplifier cabinet. The LOCAL-REMOTE switch, S302, allows the power amplifier to be operated from a remote position or from the power amplifier. With switch S302 in the RE MOTE position, filamenton, filament-off, plate-on, and plate-off functions may be selected from either a remote position or at the power amplifier. With switch S302 in the LOCAL position, filament-on, filament-off, plate-on, and plate-off functions may be selected only at the power amplifiercabinet. The PLATE circuitbreaker, CB301, is a protective device which monitors the plate supply transformer primary current. The circuit breaker will activate if the transformer primary current exceeds 5 amperes. The control circuits fuses protect the control circuits from overloads. The two 5 -ampere fuses protect the control transformer primary, while the 4 -ampere fuse protects the control transformer secondary.

The wattmeter adjusting potentiometer, R321, is located directly below the R.F. WATTMETER when the upper switch and meter panel is raised. The wattmeter adjusting potentiometer is set at the factory and normally does not require adjustment.

The overload adjusting potentiometer, R304, is located inside the right rear door on the relay panel. The overload adjusting potentiometer is set at the factory and normally does not require adjustment.

# SECTION III <br> MAINTENANCE 

### 3.1 GENERAL.

This section contains information concerning the maintenance of B830-1 250-Watt FM Power Amplifier.

## WARNING

Voltages present in this equipment are dangerous to life. Observe safety precautions when performing any maintenance. Do not reach inside the B830-1 cabinet whenever high voltage is applied. Do not depend entirely on door interlocks. Always shut down the B830-1 before doing any work inside the B830-1 cabinet. Immediately upon opening the rear doors, short out all high-voltage points using the shorting stick located inside the left rear door.

### 3.2 PREVENTIVE MAINTENANCE.

Most service interruptions in equipment of this type are caused by dirt and corrosion. Corrosion is
accelerated by the presence of moisture and dust. Dust should be removed periodically with a soft brush or a dry, oil-free air jet. Remove dust as often as a perceptible quantity accumulates at any point in the power amplifier.

When the B830-1 is operated near salt water or in other corrosive atmospheres, inspect and clean interlock switches, cable connectors, tube prongs, and other metal parts more frequently to keep the equipment in top operating condition.

### 3.2.1 AIR FILTER CLEANING.

At least once each month, or more often if needed, clean the air filter according to the following procedure:
a. Remove the air filter from the B830-1 cabinet by loosening the two thumb screws located above the air filter. Slide the air filter to the extreme right, and pull the left side of the air filter out as soon as the filter clears the panel. Slide the air filter to the left and remove.
b. Mark with an arrow the direction of airflow.
c. Wash by passing a fine spray of hot water through the filter in the direction opposite that of the airflow. Gently shake the water out of the filter.
d. Dip the filter in a water-soluble oil, such as Filter-kote " $M$ " available from Collins Radio Company, Service Parts Department, Cedar Rapids, Iowa (Collins part number 005-0609-00).
e. Remove the filter from the oil; lay the filter face down until oil ceases to drip from the filter.
f. Replace the filter into the lower rear panel with the airflow arrow (marked when the filter was removed) pointing in the direction of the airflow. Tighten the two thumb screws.
g. Replacement filters are Collins part number 009-1069-00.

### 3.2.2 PA TUBE CLEANING.

The power amplifier tube depends upon a stream of air passing through the fins to cool the anode. When these fins become dirty, the airflow is reduced and the tube life is shortened. The radiator fins should be cleaned as follows:
a. Remove the r-f amplifier tube as described in paragraph 3.2.2.1.

## CAUTION

Special care must be used in removing or installing the power amplifier tube.
b. Direct a low-pressure ( 50 psi ) air stream through the fins in the direction opposite to the normal airflow until all dust is removed.
c. Replace the $r$-f amplifier tube as described in paragraph 3.2.2.1.

### 3.2.2.1 PA TUBE REMOVAL.



Voltages present within the plate cavity are dangerous to life. Shut down the B830-1 before doing any work inside the cavity. Short the plate to ground immediately on opening the plate cavity door. Do not depend entirely on the door interlock.

The power tube may be removed as follows:
a. Open the power amplifier cavity, and loosen the anode clamp.
b. Grasp the anode with a tube puller for air-cooled tubes (or with the fingers) and lift. If the anode clamp
has not been loosened enough, it will cause binding when the power amplifier tube is removed. Care should be taken not to distort the anode clamp.
c. Replacement is the reversal of the removal procedure.

### 3.2.3 INSPEC TION.

Once each week check and clean the three interlock switches and the two shorting switches at the rear of the B830-1 cabinet to be sure they are in gooc working order.

Once each month check all connections in the B830-1. Tighten any nuts, bolts, or screws that may be loose. Check cable connections to see that they are clean and mechanically secure. Check moving parts such as tuning controls for excessive wear. Check the plate cavity slider for oxidation around ground springs.

### 3.2.4 LUBRICATION.

The PA blower is to be lubricated once every six months with two drops of SAE no. 20 oil in each bearing. The cabinet fan has bearings that are lubricated for the life of the equipment. No other lubrication of the $8830-1$ is required.

### 3.2.5 TUBE MAINTENANCE.

The power amplifier, V201, should be inspected (tube in place) once each week to ensure that an accumulation of dust does not build up on the radiator fins. If dust is present, clean as described in paragraph 3.2.2. When tuning the B830-1, care should be taken not to exceed the maximum plate current shown in table 3-1.

### 3.3 TROUBLE SHOOTING.

The most common cause of trouble will probably be traced to tube failure. If a tube is suspected of failure, replace it with a tube of known quality, and note any change in performance. A small loss in emission of V201 can be compensated for by a change in the setting of the POWER OUTPUT ADJUST potentiometer. Voltage regulator tubes V301 and V302 can be assumed to be operating properly if the screen voltage is held between 280 and 320 volts as read on MULTIME TER M301.

Four meters are located on the B830-1 front panel to assist in locating any trouble which may occur. Table 3-1 contains typical meter indications. These average indications are obtained from several production power amplifiers, the indications of some B830-1 may vary slightly outside the given limits without affecting the power amplifier performance. A list of panel meter indications for each individual power amplifier should be taken when the $B 830-1$ is operating properly in its particular installation. Any abnormal deviation from these values will then be apparent during a check of meter indications.

TABLE 3-1. TYPICAL ME TER INDICATIONS

| METER | METER SWITCH POSITION | INDICATION |
| :--- | :--- | :--- |
| MULTIMETER | SCREEN FS 400 VDC | 280 to 320 volts |
| MULTIMETER | SCREEN FS 40 MA | 5 to 20 ma |
| MULTIMETER | GRID FS 40 MA | 5 to 20 ma |
| P.A. PLATE VOLTAGE |  | 2000 to 2200 volts |
| P.A. PLATE CURRENT | Forward | Not more than 250 ma |
| R. F. WATTMETER | Reflected | 250 watts |
| R. F. WATTMETER | Less than 40 watts |  |

### 3.4 CABLE CHART.

Table 3-2 contains from-to information for cables installed in B830-1 250-Watt FM Power Amplifier. The table is useful in locating point-to-point wiring within the B830-1' cabinet. The from column is listed in alphabetical and numerical order. To find a particular wire, establish the point on the B830-1
from which wire tracing is to be initiated. Find this point in the from column of table 3-2, and the to column will give the location of the other end of that particular wire. The wire code column gives the type and color of wire used in each case. Refer to the back inside cover of this manual for the wire code explanation. When the wire code CBSJ is encountered, the letters SJ mean shield with jacket.

TABLE 3-2. FROM-TO INFORMATION

| WIRE CODE | FROM | TO | WIRE CODE | FROM | TO |
| :---: | :---: | :---: | :---: | :---: | :---: |
| RE91 | C206-1 | T302-5 | VG90 | CB301-1 | TB301-1 |
| LE9 | C210 | M303-2 | VG90 | C B301-1 | XF301-1 |
| RC4 | C211 | E 303 | RE 90 | CB301-2 | K304-5 |
| RC93 | C214 | S307-11 | VG9 | CB301-3 | XF302-1 |
| RC905 | C215 | R317-3 | VG9 | С B301-3 | TB301-2 |
| RC95 | C218 | S302-11 | RE 95 | CB301-4 | K304-3 |
| RC95 | C219 | TB304-3 | KE0 | CR301-1 | T303-10 |
| KE0 | C301-1 | L301-2 | KE0 | CR302-1 | T303-8 |
| RC90 | C301-2 | R304-3 | KE0 | CR302-2 | L301-1 |
| RC90 | C301-2 | T303-9 | C BSJ905 | DC301-3 | S308-2 |
| RC90 | C301-2 | C 302-2 | Shield | DC 301-3 | E313 |
| RC90 | C302-2 | C 301-2 | CBSJ903 | DC301-4 | S308-3 |
|  |  |  | Shield | DC301-4 | E313 |

TABLE 3-2. FROM-TO INFORMATION (Cont)

| WIRE CODE | FROM | TO | WIRE CODE | FROM | TO |
| :---: | :---: | :---: | :---: | :---: | :---: |
| KEO | E301 | L302-2 | RCg | E 310 | T301-9 |
| KEO | E 301 | S309 | RC9 | E311 | E310 |
| KEO | E 301 | TB308-1 | RC 9 | E312 | E 310 |
| KEO | E 301 | R305-1 | RCO | E 312 | K305-2 |
| KE0 | E301 | M303-1 | RC9 | E312 | K303-1 |
| RC4 | E 303 | C211 | Shield | E313 | DC301-3 |
| RC1 | E303 | S307-10 | Shield | E 313 | DC301-4 |
| RC96 | E 304 | R308 | Shield | E 313 | E306 |
| RC96 | E 304 | S307-4 | DA91 | E 314 | T302-7 |
| Shield | E 306 | M304-1 | RC9 | E317 | TB309-2 |
| Shield | E306 | E313 | RC905 | E 317 | TB303-10 |
| RC92 | E307 | TB309-1 | RC9 | E317 | S307-9 |
| RC91 | E307 | K301-6 | RC9 | E 322 | K302-6 |
| RC91 | E 307 | S314-1 | RC. 9 | E 322 | K301-3 |
| RC91 | E307 | TB307-1 | RCO | J305-1 | K301-12 |
| RC913 | E 308 | S113-1 | RC90 | J305-2 | K301-9 |
| RC913 | E308 | K305-5 | RC96 | K301-1 | S112-2 |
| RC913 | E308 | K302-3 | RC96 | K301-1 | S302-3 |
| RC913 | E308 | TB304-7 | RC902 | K301-2 | S111-2 |
| VG9 | E 309 | E 310 | RC 902 | K301-2 | K302-2 |
| RC9 | E 309 | Power supply panel | RC9 | K301-3 | E 322 |
| VG9 | E 310 | E309 | RC916 | K301-14 | K302-7 |
| VG9 | E 310 | TB301-3 | RC91 | K301-6 | E301 |
| RC9 | E310 | E311 | RC 902 | K301-7 | XF303-2 |
| RC9 | E 310 | E312 | RC 90 | K301-9 | J305-2 |
| RC9 | E 310 | M302-2 | RC93 | K301-10 | T301-5 |
| RC9 | E310 | TB303-2 | RC0 | K301-12 | J305-1 |
| RC9 | E 310 | TB304-10 | RC92 | K301-13 | T301-1 |

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B830-1 250-Watt FM Power Amplifier
TABLE 3-2. FROM-TO INFORMATION (Cont)


TABLE 3-2. FROM-TO INFORMATION (Cont)

| WIRE CODE | FROM | TO | WIRE CODE | FROM | TO |
| :---: | :---: | :---: | :---: | :---: | :---: |
| RC91 | S302-1 | TB303-1 | RC93 | S307-11 | C214 |
| RC96 | S302-3 | TB304-6 | CBSJ902 | S308-1 | R321-3 |
| RC96 | S302-3 | K301-1 | CBSJ905 | S308-2 | DC301-3 |
| RC912 | S302-4 | TB303-5 | DBSJ903 | S308-3 | DC301-4 |
| RC912 | S302-4 | TB304-11 | VD902 | S309 | S310 |
| RC 90 | S302-5 | TB303-4 | KE0 | S309 | E310 |
| RC90 | S302-5 | S311-1 | VD902 | S310 | S309 |
| RC92 | S302-7 | TB303-3 | RC90 | S311-1 | S302-5 |
| RC915 | S302-9 | TB304-1 | RC912 | S311-2 | 5312-1 |
| RC93 | S302-10 | TB302-8 | RC912 | S312-1 | S311-2 |
| RC95 | S302-11 | C218 | RC913 | S312-2 | S313-1 |
| RC95 | S302-11 | TB302-9 | RC913 | S313-1 | S312-2 |
| RC906 | S111-1 | TB302-8 | RC915 | S313-2 | S114-1 |
| RC902 | S111-2 | K301-2 | RC91 | S314-1 | E 307 |
| RC902 | S111-2 | XDS301-1 | RC92 | S314-3 | K303-4 |
| RC9 | S112-1 | TB308-6 | RC92 | S314-3 | T302-1 |
| RC96 | S112-2 | TB308-8 | RC92 | T301-1 | TB302-3 |
| RC96 | S112-2 | K301-1 | RE92 | T301-1 | XF302-2 |
| RC913 | S113-1 | E 308 | RC92 | T301-1 | K301-13 |
| RC 916 | S113-2 | K302-1 | REg1 | T301-4 | XF301-2 |
| RC 915 | S114-1 | S313-2 | RC93 | T301-4 | TB306-3 |
| RC916 | S114-2 | K305-3 | RC93 | T301-5 | K301-10 |
| RC903 | S307-2 | M301-1 | RC96 | T301-8 | XF303-1 |
| RC96 | S307-4 | E 304 | RCg | T301-9 | E310 |
| RC96 | S307-4 | R308-2 | RC92 | T302-1 | S314-3 |
| RC902 | S307-8 | M301-2 | DA91 | T302-5 | C 206-1 |
| RC9 | S307-9 | E317 | DA91 | T302-7 | E314 |
| RC1 | S307-10 | E303 | RE 93 | T303-1 | R323-2 |

TABLE 3-2. FROM-TO INFORMATION (Cont)

| WIRE CODE | FROM | TO | WIRE CODE | FROM | TO |
| :---: | :---: | :---: | :---: | :---: | :---: |
| RE91 | T303-5 | K304-6 | RC925 | TB304-5 | K304-2 |
| KE0 | T303-8 | C R302-1 | RC96 | TB304-6 | S302-3 |
| RC90 | T303-9 | C301-2 | RC913 | TB304-7 | E308 |
| KE0 | T303-10 | CR301-1 | RC913 | TB304-7 | TB303-7 |
| VG90 | TB301-1 | CB301-1 | RC902 | TB304-8 | K302-2 |
| VG9 | TB301-2 | CB301-3 | RC9 | TB304-10 | E310 |
| RC9 | TB301-3 | TB304-10 | RCg | TB304-10 | TB301-3 |
| VG9 | TB301-3 | E 310 | RC91 | TB302-2 | TB306-2 |
| RC90 | TB302-1 | TB306-1 | RC92 | TB302-3 | TB301-1 |
| RC91 | TB302-2 | TB306-2 | RC96 | TB302-7 | TB307-8 |
| RC92 | TB302-3 | T301-1 | RC93 | TB302-8 | S302-10 |
| RC96 | TB302-7 | TB307-9 | RC906 | TB302-8 | S111-1 |
| RC93 | TB302-8 | S302-10 | RC95 | TB302-9 | S302-11 |
| RC906 | TB302-8 | S111-1 | RC91 | TB303-1 | S302-1 |
| RC95 | TB302-9 | S302-11 | RC9 | TB303-2 | E310 |
| RC91 | TB303-1 | S302-1 | RC92 | TB303-3 | S302-7 |
| RC9 | TB303-2 | E310 | RC90 | TB303-4 | S302-5 |
| RC92 | TB303-3 | S302-7 | RC 912 | TB303-5 | S302-4 |
| RC90 | TB303-4 | S302-5 | RC913 | TB303-7 | TB304-7 |
| RC912 | TB303-5 | S302-4 | RC902 | TB303-8 | TB308-4 |
| RC913 | TB303-7 | TB304-7 | RC903 | TB303-9 | R318- Bottom |
| RC902 | TB303-8 | TB308-4 | RC 905 | TB303-10 | E 317 |
| RC903 | TB303-9 | R318- Bottom | RC915 | TB304-1 | S302-9 |
| RC905 | TB303-10 | E 317 | RC 923 | TB304-2 | K302-9 |
| RC915 | TB304-1 | S302-9 | RC95 | TB304-3 | C219 |
| RC923 | TB304-2 | K302-9 | RC93 | TB304-4 | XF303-2 |
| RC95 | TB304-3 | C219 | RC925 | TB304-5 | K304-2 |
| RC93 | TB304-4 | XF303-2 | RC96 | TB304-6 | S302-3 |

TABLE 3-2. FROM-TO INFORMATION (Cont)

| WIRE CODE | FROM | TO | $\begin{aligned} & \text { WIRE } \\ & \text { CODE } \end{aligned}$ | FROM | TO |
| :---: | :---: | :---: | :---: | :---: | :---: |
| RC913 | TB304-7 | E 308 | RC 9 | TB308-6 | S112-1 |
| RC913 | TB304-7 | TB303-7 | RC9 | TB308-6 | M302-2 |
| RC902 | TB304-8 | K302-2 | RC9 | TB308-6 | TB307-2 |
| RC9 | TB304-10 | E 310 | RC903 | TB308-7 | XDS301-2 |
| RC9 | TB304-10 | TB301-3 | RC96 | TB308-8 | S112-2 |
| RC912 | TB304-11 | S302-4 | RC925 | TB308-9 | XDS302-2 |
| TAS9 | TB305-1 | TB307-3 | RC916 | TB308-10 | K302-10 |
| TASO | TB305-2 | TB307-4 | RC92 | TB309-1 | E 307 |
| Shield | TB305-3 | TB30'ı-5 | RC9 | TB309-2 | E 317 |
| TAS9 | TB305-4 | TB307-6 | RC902 | XDS301-1 | S111-2 |
| TAS2 | TB305-5 | TB307-7 | RC903 | XDS301-2 | TB308-7 |
| RC90 | TB306-1 | TB302-1 | RC923 | XDS302-1 | K304-2 |
| RC91 | TB306-2 | TB302-2 | RC925 | XDS302-2 | TB308-9 |
| RC93 | TB306-3 | TB301-5 | VG90 | XF301-1 | CB301-1 |
| RC91 | TB307-1 | E 307 |  |  |  |
| RC9 | TB307-2 | TB308-6 | RE91 | XF301-2 | T301-4 |
| TAS9 | TB307-3 | TB305-1 | VG9 | XF302-1 | CB301-3 |
| TASO | TB307-4 | TB305-2 | RE 92 | XF302-2 | T301-1 |
| Shield | TB307-5 | TB305-3 | RC96 | XF303-1 | T301-8 |
| RC96 | TB307-8 | TB302-7 | RC93 | XF303-2 | TB304-4 |
| TAS9 | TB307-6 | TB305-4 | RC902 | XF303-2 | K301-7 |
| TAS2 | TB307-7 | TB305-5 | RC92 | XV301-2 | XV302-5 |
| KE0 | TB308-1 | E 301 | RC5 | XV301-5 | R305-2 |
| RC902 | TB308-4 | TB303-8 | RC96 | XV301-5 | R308 |
| RC902 | TB308-5 | M302-1 | RC92 | XV302-5 | XV301-2 |

# SECTION IV PARTS LIST 

| 1TEM | DESCRIPTION | COLLINS PART NUMBER |
| :---: | :---: | :---: |
|  | B830-1 250.WATT FM POWER AMPLIFIER | 549-2006-00 |
| C201 | CABACITOR, FDXED, CERAMIC: $0.02 \mathrm{ul}+100 \%$ -20\%, 500 vdew, Sprague Electric of Wisconsin part nc. 33C2 | 013-2142-00 |
| $\begin{aligned} & \text { C202 } \\ & \text { thru } \end{aligned}$ | CAPACITOR, FIXED, CERAMIC: same as C201 | 913-2142-00 |
| C205 |  |  |
| C206 | CAP ACITOR, FIXED, CERAMIC: 1000 Uuf $+80 \%$ -20\%. 500 vdcw; Erie Resistor part no. DA722-002 | 913-1292-00 |
| C207 | CAPACITOR, FDXED. CERAMIC- 1000 UUT $\pm 20 \%$, 5000 vdew; Centralab part no. DA-858-003 | 913-0101-00 |
| C20日 | PLATE, CAPACITOR, SOLDERED: c/obrass plate 2.500 in . dia by 0.125 in . thk; brass hub \& copper electrical lead. Irregular shape; Collins Radio Co. | 549-2059-002 |
| C209 | CAPACITOR, VARIABLE, AIR: 4.5 uuf min to 8.5 uuf max, 7000 volts; E. F. Johnson type 9G70 | 922-0570-00 |
| C210 | CAPACITOR, FIXED, CERAMIC: same as C207 | 913-0101-00 |
| C211 | CAPACITOR, FDED. CERAMIC: same as C206 | 913-1292-00 |
| C212 | Part of XV201 |  |
| C213 | CAPACITOR, VARIABLE, AIR: 3.9 uuf min to 50 uuf max; single section; Hanmerlund type APC-50B | 822-0016-00 |
| C214 | CAPACITOR. FIXED. CERAMIC; same as C206 | 913-1292-00 |
| C 215 | CAPACITOR, FIXED, CERAMIC: same as C206 | 913-1292-00 |
| C216 | CAPACITOR, FIXED. CERAMIC: same as C207 | 913-0101-00 |
| C217 | CAPACITOR. FIXED. CERAMIC. same as C 207 | 913-0101-00 |
| C218 | CAPACITCR, FIXED, CERAMIC: same as C206 | 913-1292-00 |
| C219 | CAPACITCR, FIXED: CERAMIC: same as C206 | 913-1292-00 |
| C220 | CAPACITOR, FIXED, PLASTIC: c/o 2 external brass capacitor plates \& 1 center plate, 8 Teflon washers, 4 eyelets; $1-9 / 32 \mathrm{in}$. dia by 0.225 in . Ig overall; Collins Radlo Company | 549-2126-002 |
| J201 | CONNECTOR, RECEPTACLE. ELECTRICAL. type BNC panel mtg, 1 contact; R. F. Products Co. part no. 87075 | 357-8183-00 |
| J202 | CONNECTOR, RECFPTACLE, ELECTRICAL: 1 round female contact, straght shape; 1 in. by 1 in, by 1.106 in . overall | 357-8003-00 |
| J203 | CONNEC'TOR. RECEPTACLE, ELECTRICAL: same as J201 | 357-9183-00 |
| L201 | LEAD, ELECTRICAL: pitch copper, silver plated; 0.032 in. by $1-3 / 16 \mathrm{in}$. by $4-3 / 4 \mathrm{in}$.; Collins Radio Co. | 549-2132-002 |
| L202 | COIL, RADIO FREQUENCY: 4.7 uh $\pm 10 \%, 0.60$ ohms de resistance. 950 ma d-c. single layer wound; phenolic coil form; Jeffers Electronics part no. 10402-32 | 240-0178-00 |
| L203 | COIL, RADIO FREQUENCY: copper; $3 / 4 \mathrm{in}$. Id by 2 in. $\mathrm{Ig}_{\mathrm{g}}$ Collins Radio Co. | 549-2133-00 |
| R201 | RESISTOR, FIXED, COMPOSITION: 2700 ohms $\pm 10 \%$. 2 w | 745-5670-00 |
| R202 | RESISTOR, FIXED, COMPOSITION: 3300 ohnis $\pm 10 \% 2$ w | 745-5673-00 |
| R203 | RESISTOR, FIXED, COMPOSITION: 47 ohms $\pm 10 \% .1 \mathrm{w}$ | 745-3296-00 |
| R204 | RESISTOR. FIXED, COMPOSITION: same as R202 | 745-5673-00 |
| S201 | LEAD, ELECTRICAL: berylitum copper, silver plated; $3 / 8 \mathrm{in}$. w by 2-15/16 in. Ig approx; Collins Radio Co. | 549-2060-00 |
| S202 | SWITCH. INTERLOCK: spdt: 10 amp at 250 volts a-c; screw-type terminals; Micro Switch part no. 3 AC5 | 266-8013-00 |
| V201 | ELECTRON TUEE: tetrode; RCA type 7203/ 4C×250B | 256-0138-00 |
| XV201 | SOCKET, ELECTRON TUBE: alr system tube sceket for $u / w$ ungrounded cathodes \& built-in capacitor. 1100 uuf $\pm 20$ uuf. 1000 volts $\mathrm{d}-\mathrm{c}$; E. F. Johnson part no. 124-115-2; Includes C212 | 220-1294-00 |
| Z201 | PA CAVITY. RIVETED: c/o aluminum cavity. 6 in . w by 24-13/16 in, lg ; aluminum cover 6-1/8 in. $w$ by $23-13 / 16 \mathrm{fn}$. Ig and cavity hinge; Collins Radio Co. | 549-2130-002 |
|  | electaical cabinet | 548-385日-005 |
| B301 | FAN. CENTRIFUGAL: 115 volts a-c, 60 cps . single phase, shaded pole. direct connected centrifugal (an; continuous duty cycle; Dayton Electrical Mig. Co. part no, 2C610 | 009-1576-00 |

\begin{tabular}{|c|c|c|}
\hline ITEM \& DESCRIPTION \& COLLINS PART NUMBER <br>
\hline B302 \& MOTOR, ALTERNATING CURFENT: 230 volts. 50/ $60 \mathrm{cps}, 1550 \mathrm{rpm}, \mathrm{ccw}$ rotation, totally enclosed; General Electric part no. 5KSP5ICL17 \& 230-0164-00 <br>
\hline C301 \& CAPACITOR, FIXED, PAPER: $4 \mathbf{u t} \pm 10 \%, 4000$ vdew; Sprague Electric part no. S4878 \& 930-0705-00 <br>
\hline C302 \& CAPACITOR, FIXED. PAPER: same as C301 \& 930-0705-00 <br>
\hline C303 \& CAPACITOR, FIXED, CERAMIC: $0.002 \mathrm{w} \pm \mathbf{2 0 \%}$. 6000 vdcw; Centralab type DD60 \& 113-3540-00 <br>
\hline C304 \& CAPACITOR, FIXED. CERAMIC: same as C303 \& 913-3540-00 <br>
\hline C305 \& CAPACITOR, FDXED. CERAMIC: 25 uut $\pm 10 \%$. 2500 volts rms at $2.0 \mathrm{mc}, 1500$ volts rms at 16.0 mc; Centralab part no. 850-001 \& 913-4253-00 <br>
\hline C306 \& CAPACITOR, FIXED, CERAMIC: same as C305 \& 913-4253.00 <br>
\hline C307 \& CAPACITOR, FIXED. CERAMIC: 33 uUf $\pm 10 \%$. 15,000 vdew, Centralab type DA-857A \& 913-1425-00 <br>
\hline C308 \& CAPACITOR, FIXED, CERAMIC: same as C305 \& 913-4253-00 <br>
\hline C309 \& CAPACITOR, FIXED, CERAMIC: same as C305 \& 913-4253-00 <br>
\hline CB301 \& CIRCUIT BREAKER: double pole, magnetic; back connected, $5.0 \mathrm{amp} \mathrm{a}-\mathrm{c}$ contact ratling, 230 volts a-c, 125 volts d-ci Heinemann Electric Co. part no. 2263 S \& 260-0239-00 <br>
\hline CR301 \& SEMICONDUCTOR DEVICE, DIODE: sllicon; General Electric Co. part no. 4JA421EH20AB1 \& 353-1794-00 <br>
\hline CR302 \& SEMICONDUCTOR DEVICE, DIODE: same as CR301 \& 353-1794-00 <br>
\hline DC301 \& COUPLER UNTT: double coupler; 400 w incident. 40 w rellected; 50 ohms impedance; 1-1/4 in. w by $4-1 / 4 \mathrm{in} .1 \mathrm{~g} ; \mathrm{M}, \mathrm{C}$. Jones Electronics Co. part no. $576 \mathrm{~N}(\mathrm{Mod})$ \& 277-0156-00 <br>
\hline DS301 \& LAMP. INCANDESCENT: pilot light bulb w/ candelabra base and tapered bult \& 262-3310-00 <br>
\hline DS302 \& LAMP, INCANDESCENT; same as DS301 \& 262-3310-00 <br>
\hline E301 \& TERMINAL STUD single ended insulated standoff terminai $w /$ metal case; $1 / 4 \mathrm{in}$. hex by $3 / 4 \mathrm{in}$. 1 g ; Armel part no. RTMT16-6M \& 306-0979-00 <br>
\hline E302 \& INSULATOR, STANDOFF: ceramic, white glaze on surfaces $w$ / heavy lines, other surfaces unglazed; $1.250 \mathrm{in} . \mathrm{ig}$ by $3 / 4 \mathrm{in} . \mathrm{w}$ \& 190-0025-00 <br>
\hline E303
thru \& NOT USED \& <br>
\hline E309 \& \& <br>
\hline E310 \& INSULATOR, STANDOFF: same as E302 \& 190-0025-00 <br>
\hline F301 \& FUSE, CARTRIDGE: 250 voltg, 5.0 amp , direct current; normal instantaneous \& 264-0361-00 <br>
\hline F302 \& FUSE, CARTRIDGE: same as F301 \& 264-0361-00 <br>
\hline F303 \& FUSE, CARTRIDGE: 4 amp. 125 volts, glass enclosed. 4 spares furnished; Bussman Mfg. Co. part no. MDX-4 \& 264-0217-00 <br>
\hline J301 \& CONNECTOR, RECEPTACLE. ELECTRICAL: 1 round female contact, straight shape; 1 in . by 1 in . by 1.106 in . overall \& 357-9003-00 <br>
\hline J302 \& CONNECTOR, RECEPTACLE, ELECTRICAL: same as J301 \& 357-9003-00 <br>
\hline J303 \& CONNECTOR, RECEPTACLE, ELECTRICAL: female jack connector for $u / w$ coaxial cable; 500 volts operating voltage, 50 ohms impedance; Teflon insulation, brass body, beryllium copper contact \& 357-9476-00 <br>
\hline J304 \& CONNECTOR, RECEPTACLE. ELECTRICAL: for u/w RG-55/U, RG-58/U coaxial cables; Tellon insulation, copper contact. brass shell; $11 / 16$ in. across hex, 1-5/32 in. 1g; American Phenolic part no. 31-206 \& 357-9248-00 <br>
\hline J305 \& FAN, CONNECTOR: twist-lock, 3-wire midget, 10 amp, 250 volts; $15 \mathrm{amp}, 125$ volls \& 368-0014-00 <br>
\hline K301 \& RELAY, ARMATURE: 4C contact arrangement; 115 volts a-c, $10 \mathrm{amp} ; 330$ ohms reslstance; Aemco Inc. part no. 83-3544 \& 970-1933-00 <br>
\hline K302 \& RELAY. ARMATURE: same as K301 \& 970-19.2-00 <br>
\hline K303 \& RELAY, SWITCH: adjustable time delay, snapaction switch, 20 amp at 125 volts a-c or 250 volts a-c reslstive load; 60 cps, 120 volts \& 402-0126-00 <br>
\hline K304 \& RELAY: power contactor; 2 contacts corrosion and moisture resistant; 60 cps coil rating; 25 amp . non-inductive load 600 volts; 4 in . w by $5-7 / 8 \mathrm{in}$, ig \& 405-0124-00 <br>
\hline K305 \& RELAY, ARMATURE: IC contact arrangement; 28 volts d-c or 115 volts a-c; 300 ma ; continuous duty cycle; Sigma Instruments Co. part no. 85062 \& 400-1114-00 <br>
\hline L301

L302 \& CHOKE: 10 h at 10 volt rms, $60 \mathrm{cps} \mathrm{w} /$ rated d-c current; $0.350 \mathrm{mmp} ; 90$ ohms max; 1000 volts rms ripple volt, $100 / 120 \mathrm{cps}$ ripple frequency; Electro Engr. Works part no. El2321 \& 668-0014-00 <br>
\hline L302 \& CHOKE: same as L301 \& 668-0014-00 <br>
\hline
\end{tabular}

| ITEM | DESCRIPTION | COLLINS PART NUMBER |
| :---: | :---: | :---: |
| M301 | METER, ARBITRARY SCALE: permanent magnet moving cotl type d-c milliammeter; $0-1$ ma d-c meter range; 100 ohms meter resistance; double scale, 0-40, 0-80 in 40 scale divisions | 458-0649-00 |
| M302 | VOLTMETER: permanent magnet moving coil type d-c milliammeter; 0-1 ma d-c meter range; 100 millivolts approx meter movement; $0-3000$ vde ( 60 scale divisions) | 459-n640-00 |
| M303 | AMMETER: permanent magnet moving coll tyne d-c milllammeter; $0-1$ ma d-c meter range; $0-300$ ma scale markings | 458-0638-00 |
| M304 | METER, ARBITRARY SCALE: permarent magnet moving coil type d-c microammeter; 0-200 ua meter range; 900 ohms resistance, $0-40$ scale | 458-0638-00 |
| 0301 | KNOB: black phenolic $w / 6$ flutes, aluminum insert $w /$ molded diamond knurl; 27/32 in. h by 1.500 in . dia; includes phenollc skirt; Collins Radio Co. | 546-1293-003 |
| 0302 <br> thru | KNOB: same as 0301 | 546-1293-003 |
| 0305 |  |  |
| P301 | CONNECTOR. PLUG. ELECTRICAL: type BNC cable mtg, 1 male contact, R. F, Products Div. part no. UG-88C/U | 357-8292-00 |
| P302 | CONNECTOR, PLUG, ELECTRICAL: same as P301 | 357-8292-00 |
| P303 | CONNECTOR, PLUG, ELECTRICAL: brass body and contacts; Tellon insulation; $3 / 4 \mathrm{in}$. dia approx by 1-1/2 in. 1g approx; Amphenol part no. UG-1185/U | 357-9326-00 |
| PJ04 thru | CONNECTOR, PLUG, ELECTRICAL: same as P303 | 357-9326-00 |
| P307 |  |  |
| P308 | CONNECTOR. PLUG. ELECTRICAL: same as P301 | 357-8292-00 |
| P309 | CONNECTOR, PLUG, ELECTRICAL: twist lock, 3 -wire midget; 10 amp at 250 volts, 15 amp at 125 volts | 368-0013-00 |
| R301 | RESISTOR, FIXED, COMPOSITION: 1000 ohms $\pm 10 \%$. 2 w | 745-5652-00 |
| R302 | RESISTOR, FIXED, COMPOSITION: same as R301 | 745-5652-00 |
| R303 | RESISTOR. FDSED WIREWOUND: 10 ohms $\pm 5 \%$, 5 w | 747-5420-00 |
| R304 | RESISTOR, VARIABLE: wirewound; 50 chms $\pm 10 y_{1}$ 2 w | 377-0610-00 |
| R305 | RESISTOR, FIXED. WIREWOUND: 31,000 ohms $45 \% .210 \mathrm{w}$ | 746-6727-00 |
| R308 | RESISTOR, FIXED, COMPOSITION: 10 ohms $\pm 10 \% .2 \mathrm{w}$ | 745-5568-00 |
| R307 | RESISTOR, FIXED. WIREWOUND: 2.56 ohms $\pm 1 \%$, 2.5 w; Dale Products part no. RSM2C/2R560F | 746-9448-00 |
| R308 | RESISTOR, FIXED. WIREWOUND: 5000 ohms $\pm 5 \%$ 10w | 710-2913-00 |
| R309 | RESISTOR, FIXED, WIREWOUND: same as R308 | 710-2913-00 |
| R310 | RESISTOR. FIXED. FILM: 1 megohm $\pm 1 \%$, 2 w | 705-4254-00 |
| R311 | RESISTOR, FIXED, FILM: same as R310 | 705-4254-00 |
| R312 | RESISTOR, FIXED, FILM: same is R310 | 705-4254-00 |
| R313 | RESISTOR. FIXED. FILM: 5110 ohms $\pm 1 \%, 1 / 2 \mathrm{w}$ | 705-7630-00 |
| R314 | RESISTOR, FIXED, COMPOSITION: 10,000 ohms $\pm 10 \%, 1 \mathrm{w}$ | 745-3394-00 |
| R315 | NOT USED |  |
| R316 | RESISTOR, FIXED, WIREWOUND: same as R307 | 746-9448-00 |
| R317 | RESISTOR, Variable: power type; 500 ohms $\pm 10 \mathrm{~h} .50 \mathrm{w}$ | 736-0456-00 |
| R318 | RESISTOR, FLXED, WIREWOUND: 160 ohms $\pm 5 \%$, 10 w | 710-2821-00 |
| R319 | RESISTOR, FIXED. WIREWOUND: 20 ohms $\pm 5 \%$. 5 w | 710-3035-00 |
| R320 | RESISTOR, FIXED, COMPOSITIUN: 380 ohms $\pm 10 \%$. 2 w | 745-5635-00 |
| R321 | RESISTOR, VARIABLE: composition; 5000 ahms $\pm 30 \%, 1 / 4$ w | 378-4729-00 |
| R322 | RESISTOR, FIXED, FILM: 402,000 ohms $\pm 1 \%, 1 \mathrm{w}$ | 705-3287-00 |
| R323 | RESISTOR, FDXED, WIREWOUND: $1.0 \mathrm{ohm} \pm 5 \%$, 26 w | 747-1626-00 |
| S301 | NOT USED |  |
| S302 | SWITCH, ROTARY: 4-clrcuit, 4-pole, 2-position, 1 -section; 4 moving, 12 fixed contacts | 258-1564-00 |
| S303 | SWITCH, PUSH: normally open and closed; Bakelite; Arrow-Hart \& Hegeman part no. B-2 | 260-2020-00 |


| ITEM | DESCRIDTION | $\begin{gathered} \text { COLLING } \\ \text { PART NUMBER } \end{gathered}$ |
| :---: | :---: | :---: |
| S304 | SWITCH, PUSH: same as S303 |  |
| thris |  |  |
| S306 |  |  |
| S307 | SWITCH, ROTARY: 2-circult, 2-pole, 3-position, 1 -section; 2 moving. 8 fixed contacts | 259-1565-00 |
| S308 | SWITCH, TOGGLE: spdt; 125 volts or 250 volts d-c, 5 amp; Micro Swltch, Div. MinneapollsHoneywell part no. 6AT11-T | 260-1509-00 |
| ③09 | SPRING. INTERLOCK: beryllium copper, silver plated; 0.040 in . by 2-1/8 in. by 4-1/32 in. approx. | 548-2315-003 |
| S310 | SPRING, INTERLOCK: same as S309 | 549-2315-003 |
| S311 | CONTACT ASSEMBLY, ELECTRICAL: 5/8 in. by 11/16 in. by 1-7/8 in. overall; Neptune Electronjes part no. M-7460330G4 | 260-4040-00 |
| S312 | CONTACT ASSEMBLY, ELECTRICAL: same as 311 | 260-4040-00 |
| S313 | CONTACT ASSEMBLY. ELECTRICAL; same as S3I1 | 260-4040-00 |
| S314 | SWITCH, AIRFLOW INTERLOCK: spdt; 5 amp at 250 volts a-c, metal case | 266-8307-00 |
| T301 | TRANS FORMER, POWER, STEP-DOWN: 200, 210, 220. $230,240 \& 250$ volt input, $50 / 60 \mathrm{cps}, 120$ volts at 4.25 amp output; $5-1 / 8 \mathrm{in}$. by $5-1 / 8 \mathrm{in}$. by $5-3 / 8$ In.; Electro Engineering Works part no. E 12322 | 662-0043-00 |
| T302 | TRANSFORMER, POWER, STEP-DOWN: 115 volts, 230 volts, $50 / 60 \mathrm{cps}$ input, 6.3 -volt center tapped output; $2-17 / 32 \mathrm{in}$, by 3-1/32 in. by 3-3/32 in.; Cbicago Std. Transformer part no. 12256 | 662-0162-00 |
| T303 | TRANSFORMER, POWER, STEP-UP: 200, 210, 220. $230,240 . \& 250$ volt input, $50 / 60 \mathrm{cps}, 2100$ volts center tapped at 250 ma output; 6-3/8 in. by 6-1/2 in. by 7-5/16 in.; Raytheon Mfg. Co. part no. 292-5783G1 | 662-0041-00 |
| TB301 | TERMINAL BOARD: barrier type; black phenolle connector strip; 4 terminals; 1-5/16 in. w by 3-7/32 in. 1g; Howard B. Jones part no. 4-142 | 367-5040-00 |
| TB302 | TERMNNAL BOARD: Bakelite; black finish; 10 terminal connector strips; 5-40 terminal screws; Howard B. Jones, Dtv, Cinch Mig. Co. part no. 140-10 | 367-3100-00 |
| TB303 | TERMINAL BOARD: same as TB302 | 367-3100-00 |
| TB304 | TERMINAL BOARD: phenolic; $13 / 32 \mathrm{in}$. by $7 / 8 \mathrm{in}$. by 5-11/64 in.; Includes 12 screw-type terminals; barrier type; Kulka Electric Corp, part no. 600-11-M | 367-0518-00 |
| TB305 | TERMINAL BOARD: same as TB304 | 367-0518-00 |
| TB306 | TERMINAL BOARD: black Bakelite; $1 / 2 \mathrm{in}$. by 1-1/8 in. by 2-1/2 in.; 4 terminals included; Howard B. Jones, Div. Cinch Mfg. Co. part no. 4-141 | 367-4040-00 |
| TB307 | TERMINAL ROARD: barrier type, 14 terminals; 5/8 in. w by 8-3/8 ln. Ig; Howard B. Jones, Div. Cinch Mig. Corp. part no. 14-162A-R | 367-0300-00 |
| TB308 | RESISTOR ASSEMBLY: includes plastic terminal board, $1 / 16$ in. by $3-1 / 2 \mathrm{in}$. by $4-9 / 16 \mathrm{in}$. \& 7 fixed resistors | 549-2102-002 |
| TE309 | TERMINAL BOARD: phenolic; $1 / 16 \mathrm{in}$. thk, $3 / 8 \mathrm{in}$. w by 1-1/8 In. Ig overall excluding terminals; Cinch Mfg. Corp. part no. 18A18697 | 306-0168-00 |
| v301 | ELECTRON TUBE: RCA type OD3/VR150 | 257-0001-00 |
| v302 | ELECTRON TUBE: same as V301 | 257-0001-00 |
| xDs301 | LAMPHOLDER: panel mounting for $u / w$ candelabra screw base lamp | 262-0255-00 |
| XDS302 | LAMPHOLDER: game as XDS301 | 262-0255-00 |
| XFJ01 | FUSEHOLDER: extractor post type; transparent with 3 AG luses; Bussman MIg. Co. part no. HKL-JRZ | 265-1040-00 |
| XF302 | FUSEHOLDER: same as XF301 | 265-1040-00 |
| XF303 | FUSEHOLDER: same as XF301 | 265-1040-00 |
| xv301 | SOCKET, ELECTRON TUBE: 8 prong octal tube socket; molded construction plastic; Amphenol part no. 88-8 TM | 220-1005-00 |
| XV302 | SOCKET, ELECTRON TUBE: same as XV301 | 220-1005-00 |

TD-538


Figure 4-1. B830-1 250-Watt FM Power Amplifier, Rear View, Parts Location


Figure 4-2. Meter Panel Subassembly, Parts Location


Figure 4-3. Plate Cavity Subassembly, Parts Location


Figure 4-4. Relay Panel Subassembly, Parts Location


Figure 4-5. Cabinet Blower Location


# 830D-1A <br> 1000 Watt FM Broadcast Transmitter 

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## general description

### 1.1 Purpose of Instruction Book.

This instruction book is a guide for installing, adjusting, operating, and maintaining 830D-1A 1000 Watt FM Broadcast Transmitter.

### 1.2 Purposes of Equipment.

The 830D-1A 1000 Watt FM Broadcast Transmitter is used for continuous monophonic or optional stereophonic FM broadcast service on a single frequency in the range from 88 to 108 megacycles with an output power of 1000 watts.

### 1.3 Desuriphion of Equipmant.

### 1.3.1 PHYSICAL DESCRIPTION.

The 830D-1A 1000 Watt FM Broadcast Transmitter, shown in figure $1-1$, is contained in a single cabinet that is 38 inches wide, 76 inches high, 27 inches deep, and weighs approximately 776 pounds. All transmitter operating controls are located behind the doors on the front of the cabinet. The filament and plate onoff controls and four monitoring meters are located at the top front of the cabinet. The meters may be observed easily while operating the tuning controls. The transmitter uses 6 tubes and 20 transistors, all of which are accessible from the front of the transmitter. The bottom front of the transmitter cabinet is removable to allow access to components on the bottom of the inside panel.

Large doors at the upper rear of the cabinet (see figure 1-2) allow access to the upper part of the transmitter for servicing and maintenance. The lower rear half of the transmitter cabinet is covered by a removable panel that contains a ventilating fan and permanent type air filter. Operating persomel are protected by both electrical and mechanical interlocks on the rear doors and panel. These interlocks ground the transmitter high-voltage circuits when the doors are opened or the panel is removed. The power amplifier plate-tuning cavity is located in an interlocked compartment at the front of the transmitter.

Inside the transmitter, heavy iron-core components are at the bottom of the cabinet. The exciter portion of the transmitter is mounted on a 19-inch rack on one side of the cabinet. A harmonic filter that is connected to the transmitter output is suspended from the top of the cabinet.

Cooling air for the transmitter is drawn through a permanent type air filter at the rear of the cabinet by a low-speed, high-volume fan, and exhausted through a shielded opening in the top of the cabinet. An individual pressure blower supplies cooling air directly to the power amplifier tube.

### 1.3.2 ELECTRICAL DESCRIPTION.

The 830D-1A 1000 Watt FM Broadcast Transmitter is composed of two electrically connected subunits: (1) a wide band exciter that furnishes a 10 -watt $F M$ output to drive (2) a 1000 -watt power amplifier. Instruction books covering the exciter and power amplifier used in the transmitter are listed in table 1-1. These two books are supplied following section 5 of this system instruction book. The subunit instruction books contain detailed descriptions of the two transmitter subunits.

TABLE 1-1
SUBASSEMBLY INSTRUCTION BOOKS

| PUBLICATION | INSTRUCTION BOOK <br> PART NUMBER |
| :---: | :---: |
| A830-2 10 W Wide-Band FM <br> Broadcast Exciter <br> D830-1 1000 Watt FM <br> Power Amplifier | TD-536 |

1.1 Equipment Supplied.

Table 1-2 lists equipment that is supplied as part of 830D-1A 1000 Watt FM Broadcast Transmitter.

TABLE 1-2
EQUIPMENT SUPPLIED

| EQUIPMENT | COLLINS PART NO. |
| :---: | :---: |
| A830-2 10 W Wide-Band FM <br> Broadcast Exciter | $522-2714-00$ |
| D830-1 1000 Watt FM <br> Power Amplifier | $522-2948-00$ |
| 250 Watt/1 KW Harmonic <br> Filter | $549-2010-00$ |



Figure 1-1. 830D-1A 1000 Watt FM Broadcast Transmitter, Over-all View

### 1.5 Accessory Equipment.

Table 1-3 lists accessory equipment that is available for use with 830D-1A 1000 Watt FM Broadcast Transmitter. Information on the $786 \mathrm{M}-1$ Stereo Generator will be found in the applicable unit instructions.

TABLE 1-3
ACCESSORY EQUIPMENT

| EQUIPMENT | COLLINS PART NO. |
| :---: | :---: |
| $786 \mathrm{M}-1$ Stereo Generator | $522-2914-00$ |

### 1.6 Equipment Sperilicalions.

### 1.6.1 MECHANICAL.

Weight
Size
Ventilation
Ambient temperature range . . . . . . . . . . . . . $+10^{\circ} \mathrm{C}\left(50^{\circ} \mathrm{F}\right)$ to $+55^{\circ} \mathrm{C}\left(131^{\circ} \mathrm{F}\right)$.
Ambient humidity range . . . . . . . . . . . . . . 0 to 95 percent relative humidity.
Altitude
0 to 6000 feet.
Shock and vibration
Normal handling and transportation.

### 1.6.2 ELECTRICAL.





Figure 1-2. 830D-1A 1000 Watt FM Broadcast Transmitter, Rear View with Lower Panel Removed

## installation

## 2. L Unpacking and luspecting.

Be careful when uncrating the transmitter and components to avoid damaging the equipment. Inspect the transmitter carefully for scratches, dents, or other physical damage. Check for loose screws and bolts. Inspect all controls, such as switches, for proper operation as far as can bedetermined without applying jower to the transmitter. Examine cables and wiring, making sure that all connections are tight and clear of each other and the chassis. File any damage claims promptly with the transportation company. If such claims are to be filed, retain all packing material.

## NOTE

Before installation, check all transistors for proper placement. The transistor location tab must be pointing to the transistor socket locating mark placed adjacent to the transistor socket.

### 2.2 Transmitter Location.

Plan transmitter and wiring placement carefully before starting installation work. Refer to figure 2-1, the transmitter installation diagram. This diagram shows the location of all wiring openings in the transmitter cabinet. As will be noted, several alternate wiring arrangements can be used. Select the combination that most nearly meets the station requirements.

Allow adequate clearance both in front and back of the transmitter. There should be a minimum of clearance of $3-1 / 2$ feet behind the transmitter to provide sufficient room for service work.

### 2.3 External Connedions.

Refer to figure 2-1 for assistance in making the following external connections.

## WARNING

Disconnect the transmitter 230 -volt ac power from the fused cutout box before making any connections to, or within, the transmitter.
a. Connect the audio input to the transmitter. Bring the audio signal through the bottom of the cabinet (or other optional input) on a shielded twisted pair. Connect the two audio leads to terminals 1 and 2 of TB305. If the optional stereophonic operation is employed, the left audio leads are connected to TB305-1 and 2 and the right audio leads to TB305-4 and 5. TB305 is located about half way up the cabinet on the left side as viewed from the rear of the cabinet. Connect the shield(s) to terminal 3 of TB305.
b. Connect the FM monitor to the monitor output on the top of the cabinet. Refer to figure 2-1. Use type RG-58U coaxial cable to make this connection.
c. Comnect the antenna transmission line to the r-f output connector on top of the $830 \mathrm{D}-1 \mathrm{~A}$ cabinet.

## CAUTION

Before making this antenna comection, be sure that the transmission line and antemna present a nominal impedance of 50 ohms and an swr of not more than $2: 1$ at the transmitter operating frequency. If the transmitter output is improperly matched, the transmitter will not operate properly and may be damaged. THIS IS IMPORTANT.
d. Comnect the power input cable to the transmitter. This power cable should be brought from an external fused cutout box rated for 12 amperes. Use number 12 wire or larger to make this comnection. Connect the power leads to terminals 1 and 2 of TB301, located at the lower left corner of the transmitter cabinet as viewed from the rear. Comnect the neutral wire to terminal 3 of TB301. The power cable may be brought into the transmitter through holes in either the bottom or rear of the cabinet. Make sure that the PLATE circuit breaker on the front panel is set to OFF before making these power connections.

### 2.4 Internal Comections.

The 830D-1A 1000 Watt FM Broadcast Transmitter 830D plate and control circuit power transformers are fitted with adjustable taps to compensate for line voltage variations. These taps compensate for line variations from 200 to 250 volts in 10 -volt steps. To adjust transformer T301 and T303 for line voltage variations, perform the following steps.


Figure 2-1. 830D-1A 1000 Watt FM Broadcast
Installation Dutine and Installation Drawing
6
a. Measure the line voltage at the transmitter fused cutout box.
b. Remove the solder lug from T301, terminal 4, and move to the transformer terminal whose input voltage is nearest to the voltage measured in step a. Do not move the solder lug from transformer terminal 5 , as this terminal supplies 230 volts to the cabinet fan for all line input connections. See figure 2-2 for transformer terminal numbers versus input voltage.
c. Remove the solder lug from T 303 , terminal 5 , and move to the transformer terminal whose input voltage is nearest to the voltage measured in step a. See figure $2-2$ for transformer terminal numbers versus input voltage.
d. Tighten all transformer terminal connections.

The following connections on TB304 should be checked to ensure that the plate-on and filament-off functions will operate: TB304-1 to TB304-2, TB304-3 to TB304-4, TB304-10 to TB304-11.
If the optional stereo generator is installed any time after the initial 830D-1A installation, the $18-\mathrm{db}$ audio pad will have to be removed from the audio circuitry of the exciter.

### 2.5 Remote Control.

Remote control of 830D-1A 1000 Watt FM Broadcast Transmitter can easily be accomplished by connection to terminal boards TB302 and TB303. Table 2-1 lists the terminal board connections and the remote functions of each pair of terminals. Remote "on'" switches should be the normally open momentary type. Remote "off" switches should be the normally closed momentary type. For remote operation, the LOCAL-REMOTE switch within the transmitter cabinet should be in the REMOTE position. When in the REMOTE position, it is still possible to control the transmitter from the transmitter panel switches.

For simplified operation, the FILAMENT ON and PLATE OFF switches could be eliminated. The PLATE ON switch starts a sequence of operations which turns the filaments on and the plate voltage on after the filament time delay is completed. The FILAMENT OFF switch shuts down all transmitter functions.

Equipment is available that will completely control and monitor transmitter operation from a remote location through standard telephone pairs. When such remote control equipment is used, necessary installation and connection information will be supplied with the remote equipment.

If an optional stereo generator is employed in the $830 \mathrm{D}-1 \mathrm{~A}$, remote control of the stereo mode may be accomplished by a ground on TB302-7. If the ground is present, the transmitter will be in the stereo mode. If the ground is removed, the transmitter will be in the monaural mode. Local control of the stereo mode is also available at the transmitter.

TABLE 2-1
REMOTE CONTROL CONNECTIONS

| FUNCTION | TERMINALS |  |
| :--- | :---: | :---: |
|  | TB302 | TB303 |
| FILAMENT ON |  | 1 and 2 |
| FILAMENT OFF | 8 and 9 |  |
| PLATE OFF |  | 4 and 2 |
| PLATE ON |  | 3 and 2 |

### 2.6 Frequency Change.

If the transmitter operating frequency is changed, the following components will have to be changed or adjusted. These components are (1) exciter heterodyning oscillator crystal (2) plate cavity slider (3) grid tank inductance (4) the screen neutralization.

Table 2-2 lists the channel frequency versus crystal frequency and the Collins part number for each crystal. Figure 2-3 shows the distance the plate cavity slider


Figure 2-2. Transformer Details

TABLE 2-2. CRYSTAL PART NUMBERS

| CHANNEL <br> FREQ <br> (mc) | CRYSTAL <br> FREQ <br> (mc) | COLLINS <br> PART NUMBER | CHANNEL <br> FREQ <br> $(\mathrm{mc})$ | CRYSTAL <br> FREQ <br> $(\mathrm{mc})$ | COLLINS <br> PART NUMBER |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 88.1 | 74.10000 | $289-2744-00$ | 98.1 | 84.10000 | $289-2794-00$ |
| 88.3 | 74.30000 | $289-2745-00$ | 98.3 | 84.30000 | $289-2795-00$ |
| 88.5 | 74.50000 | $289-2746-00$ | 98.5 | 84.50000 | $289-2796-00$ |
| 88.7 | 74.70000 | $289-2747-00$ | 98.7 | 84.70000 | $289-2797-00$ |
| 88.9 | 74.90000 | $289-2748-00$ | 98.9 | 84.90000 | $289-2798-00$ |
|  |  |  |  |  |  |
| 89.1 | 75.10000 | $289-2749-00$ | 99.1 | 85.10000 | $289-2799-00$ |
| 89.3 | 75.30000 | $289-2750-00$ | 99.3 | 85.30000 | $289-2800-00$ |
| 89.5 | 75.50000 | $289-2751-00$ | $289-2752-00$ | 99.5 | 85.50000 |





Figure 2-3. Plate Cavity Tuning Chart and Control Grid Inductor Spacing
should be positioned from the deck plate (tube socket mounting plate) for each frequency within the 88 to 108 megacycle range. Figure $2-3$ also includes the grid inductance spacing for the same frequency range. Figure 2-4 gives the approximate positioning of the screen neutralizing inductor for the FM band. The plate cavity slider, the grid inductance, and the screen neutralizing inductor may have to be repositioned from positions shown to compensate for individual transmitter characteristics. Refer to the test procedures in paragraph 2.6.1 for tuning details.

### 2.6.1 NEUTRALIZATION PROCEDURE.

The 1000-watt power amplifier will have to be neutralized if the transmitter frequency is changed and may have to be neutralized if power amplifier tube V201 is replaced. If the frequency has changed, roughly adjust the neutralizing inductor to the spacing given in figure 2-4.

Neutralization is best accomplished by using the feedthrough method. As the sensitive voltmeter necessary for use with this method is normally not available at most stations, the following neutralization procedure should be substituted. Proper neutralization can be checked by tuning the plate through resonance and noting a minimum of change in control grid current.
a. Press the FILAMENT ON switch, and turn off the exciter. Allow the transmitter to warm upfor at least 15 minutes. Place the WATTMETER switch in the FORWARD position.
b. Press the PLATE ON switch, and observe the R. F. WATTMETER. If there is an indication on the R. F. WATTMETER, the transmitter is not neutralized. If an indication is present, turn off the plate power, and open the plate cavity. Adjust the neutralizing inductor a SMALL AMOUNT by sliding both shorting blocks in opposite directions. This adjustment is critical. Do not move the shorting blocks a large distance.
c. Turn on the transmitter, and note the new indication on the R. F. WATTMETER. If the indication has increased, readjust the shorting blocks in the opposite direction.
d. If the indication has decreased upon applying plate voltage, repeat steps $b$ and $c$ until the R.F. WATTMETER indication is zero.
e. With the plate voltage on, remove the grid bias fuse, F304, and watch the R. F. WATTMETER for an indication. The PA plate current will start to rise immediately upon removing the grid bias fuse and will continue to rise until the overload relay trips or the plate current is shut off.
f. Repeat steps $b$ and e until no further indication is shown on the R. F. WATTMETER.
g. Turn on the exciter, and retune the power amplifier according to the procedure given in paragraph 2.7.0.


Figure 2-4. Screen Neutralization Inductor Spacing

### 2.7 Starting the Transmitter in a New Installation.

Before starting the transmitter for the first time, read section 2 of the subunit instruction books to become familiar with the location and function of the various transmitter controls. Then, perform the following procedure.

## W ARNING

Voltages are present in this transmitter that are dangerous to life. Observe safety precautions when making any transmitter adjustments. Do not reach inside the rear of the transmitter cabinet when high voltages are applied. Do not depend entirely on door interlocks. Always shut off transmitter power at the external cutout box, and ground all capacitors with the shorting stick in the transmitter cabinet before doing any work inside the rear of the cabinets. When working in the power amplifier cavity, remember that 115 volts a-c is present on one side of the cavity compartment interlock. Keep metal tools and the hands away from all transistor cases.
a. Complete the entire transmitter installation procedure as directed in this instruction book.
b. Close the doors at the rear of the cabinet. Open the doors at the front of the cabinet, and remove the lower front panel so that the entire inside panel is exposed. Check that the plate cavity slider, the grid inductor, and the screen neutralizing inductor conform approximately to the distance specified in figure 2-3 for the operating frequency. These adjustments have been set at the factory and will not normally require further adjustment. Close the cover on the plate cavity compartment.
c. Set the PLATE circuit breaker on the front panel to ON.
d. Press the FILAMENT ON switch. The green indicator lamp at the top left of the cabinet should light. This means that all transmitter tube filaments and cooling-air blowers are operating.
e. Place the proper crystals into the exciter sockets. The 14-megacycle crystal is placed into the Y501 socket. The heterodyning crystal is placed into the Y426 socket. Turn on the exciter, and allow it to warm up for at least 15 minutes.
f. Turn S101 to the 14 MC REF B position, and check M101 for an indication in the B meter range. Turn S101 to the AFC KEY B position, and check M101 for an indication in the $B$ meter range.

## NOTE

When S101 is in the AFC KEY position, the meter pointer will not hold steady but will pulse at the $5-\mathrm{cps}$ keying generator rate. This pulsing is an indication of normal operation.

Turn S101 to the MOD OUTPUT B position, and check M101 for an indication in the $B$ meter range. If all meter indications for the three S101 positions fall within the B meter range, proceed with the following power amplifier tuning procedures. If any of the meter indications fall outside of the $B$ meter range, the modulator and afc discriminator is out of adjustment and will have to be adjusted according to the Modulator and AFC Discriminator Adjustment Procedures in the Maintenance section of this instruction book.
g. Switch S101 to the MIXER GRID A position. Set POWER OUT resistor R454 to its midposition. Using a nonmetallic screwdriver type tuning tool, adjust the BUF TUNE control for a peak indication on meter M101.

## NOTE

The MIX BAL control, R438, should be placed in its midrange position. No further adjustment of this control is then necessary unless the transmitting frequency falls within the range of 97 to 100 megacycles. See step $m$ if the transmitted frequency falls within this range.
h. Switch S101 to the V428B position. Adjust L429 and L430 for a maximum indication on M101.
i. Switch S101 to the V429B position. Adjust L431 and L432 for a maximum indication on M101.
j. Switch S101 to the V430B position. Adjust L433 and L434 for a maximum indication on M101.
k. Switch S101 to the V430C position. Adjust the PA PLATE control for a minimum indication on M101.

1. Set the MULTIMETER switch to GRD FS 4 MA. Remove the grid bias fuse F304. Adjust first the exciter PA MATCH control, then the power amplifier GRD TUNING control for a peak MULTIMETER indication. (The grid tuning capacitor should be near its center when the peak occurs. If not approximately in this position, move the gridinductance in the proper direction, and repeat the above step.) Replace the grid bias fuse and peak PA MATCH control and GRID TUNING control. Set GRID COUPLING control for 0.5 ma of grid current.
m . If the transmitter frequency falls between 97 and 100 megacycles, the following additional step will have to be completed. Place a grid dip meter tuned to 98 megacycles near the exciter output. Adjust the MLX BAL control for a minimum output as indicated on the grid dip meter.
n. Set the POWER OUTPUT ADJUST control fully counterclockwise.
o. Press the PLATE ON switch. The red indicator lamp at the top right of the cabinet should light, and the P. A. PLATE VOLTAGE meter should indicate $2700 \pm 100$ volts.
p. Set the MULTIMETER switch to SCREEN FS 400 VDC. The MULTIMETER should indicate $240 \pm 30$ volts.
q. Adjust the PLATE TUNING control for a dip in the P. A. PLATE CURRENT meter indication.
r. Set the WATTMETER switch to FORWARD. Adjust the OUTPUT COUPLING control for approximately 10 ma of screen current.
$s$. Turn the POWER OUTPUT ADJUST control approximately two-thirds of its maximum clockwise rotation.
$t$. Increase the transmitter coupling a small amount by turning the OUTPUT COUPLING control clockwise until the PA screen current is reduced to approximately 10 ma . Adjust the PLATE TUNING control for a dip in the P. A. PLATE CURRENT meter indication. (The plate tuning capacitor should be near its center yosition when the dip in power amplifier plate current occurs. If the capacitor is not in this position, move the plate slider in the appropriate direction and repeat step t.)
u. Rotate the POWER OUTPUT ADJUST control clockwise a small amount.
v. Repeat steps $t$ and $u$ until the P. A. PLATE CURRENT meter indicates the transmitter output is 1000 watts as measured by the indirect method. At this time, the PA screen current should be not less than 10 ma. or more than 35 ma . The indirect method of measuring power output is:

$$
\text { Power Output }=I_{p} E_{p} K
$$

when $K$ is efficiency, $E_{p}$ plate voltage, and $I_{p}$ plate current.

## NOTE

Determine efficiency by referring to the efficiency chart shown in figure 2-5.
w. Check to be sure that the FM monitor that is connected to the transmitter is properly calibrated. Then, if necessary, adjust the exciter VHF OSC FREQ ADJ control until the monitor indicates that the transmitter operating frequency is within specified operating limits.
x. Apply a $50-\mathrm{cps}$ audio tone to the transmitter input. The input level should be such that the voltage at the transmitter audio input terminals is +10 dbm .
y. Adjust the exciter MOD GAIN control until the monitor indicates 100 percent modulation.
z. Replace the lower front panel on the transmitter cabinet, and close the front doors. The transmitter is now ready for standard broadcast use.

## NOTE

At this point, it is suggested that a record be made of all meter readings for future maintenance and trouble shooting. These meter readings may be recorded in table 5-3.


Figure 2-5. Efficiency Chart

ADDITION TO
APPLICATION FOR FCC TYPE ACCEYTAACE COLLIMS TYPE 83OD-1 FM TRANSMTTTER

### 1.0 POHER CAPABILITIES OF THE COLLINS TYPE 830D-1 FM TRANSATTIER

The power range of the Collins Tyoa 830D-1 FM Transmitter is 250 to 1050 watts. The following graph denonstrates noninal efficiency over the power range.



4
FRODUENCY (MS)

### 3.1 General.

Refer to the subunit instruction books to become familiar with the operation and function of controls on both the jower amplifier and the exciter.
After the transmitter has been placed in operation, it will only be necessary to check meter indications from time to time to be sure the transmitter is operating properly and occasionally to "touch-up" the power amplifier loading and tuning.

### 3.2 Starting the Transmitter in Normal Operation.

The transmitter may be put into operation by two different methods, depending upon the circumstances. For normal operation, press the FILAMENT ON switch to start the power amplifier filament and to warm up the exciter (the exciter POWER switch should be left in the ON position at all times). Check the power amplifier grid current to be sure the exciter is presenting sufficient drive to the power amplifier
before applying plate power. Approximately 3 to 4 minutes after filament power is applied, the PLATE ON switch may be pressed, starting the transmitter.

The alternate method of starting the transmitter consists of pressing the PLATE ON switch only. The power amplifier filament and the exciter will immediately start to warm up. As soon as the power amplifier time delay relay has completed its cycle, the power amplifier plate voltage will come on automatically, starting the transmitter.
To shut down the transmitter it is recommended, but not necessary, to press the PLATE OFF switch, wait a few seconds, and then shut off the filament and exciter power by pressing the FILAMENT OFF switch. It is also possible to press the FILAMENT OFF switch only, which removes plate, filament, and exciter power. Pressing the PLATE OFF switch first allows the plate power supply voltage to discharge through the power amplifier while the filament is at normal operating temperature and, in addition, cools the power amplifier components.

## principles of operation

output is a 14 -me signal frequency modulated by the baseband audio. The deviation of the $14-\mathrm{mc}$ signal is $\pm 75 \mathrm{kc}$ for 100 percent modulation. The oscillator output is coupled through two limiters to remove any amplitude modulation. The limited $14-\mathrm{mc}$ signal is then amplified and coupled to the rate correction frequency discriminator and to the output amplifier. The output of the frequency discriminator is simply the baseband audio detected from the modulated $14-\mathrm{mc}$ signal. This detected audio is coupled back to the baseband input to correct for any nonlinearity in C654.

The output amplifier amplifies the modulated $14-\mathrm{mc}$ signal to a level sufficient to mix with the 74- to 94mc signal (per customer requirement) in the balanced mixer.

A portion of the limiter output is coupled to the afc buffer stage. The afc buffer output, the modulated
$14-\mathrm{mc}$ signal, is coupled to the reference oscillator and afc limiters through a diode switch. The output of the 14 -mc reference oscillator, is also coupled to the reference oscillator and afc limiters through a diode switch. The diode switch is operated by a 5 -cps keying generator. The 5 -cps generator is a unijunction transistor operating as a relaxation oscillator keying a multivibrator.

The diode switch alternately connects the modulated $14-\mathrm{mc}$ signal (afc buffer output) and the $14-\mathrm{mc}$ reference signal. The limiter output is coupled to the afc discriminator. The afc discriminator detects the difference between the $14-\mathrm{mc}$ reference signal and the modulated $14-\mathrm{mc}$ signal. The modulated $14-\mathrm{mc}$ signal will cause a baseband audio output at the discriminator. This is not an error infrequency, so a portion of the baseband audio input is amplified by the baseband canceling amplifier and fed into the output of the frequency discriminator through a diode switch. This diode switch is keyed by the same 5 -cps signal which switched the reference oscillator and afc limiter input. When the modulated $14-\mathrm{mc}$ signal is connected to the reference oscillator and afc limiter input, the baseband canceling signal is switched into the output of the frequency discriminator to cancel the baseband output from the discriminator.

The input signal to the four error signal amplifiers is a 5 -cps square wave. The amplitude of this square wave is proportional to the frequency error in the $F M$ oscillator. The error signal amplifier square wave output is converted to a d-c control signal in the synchronous detector. The synchronous detector is also keyed by the 5 -cps keying signal. The d-c error signal is coupled to C654 to correct the frequency modulation oscillator.

The modulated $14-\mathrm{mc}$ signal from the output amplifier is heterodyned up to the operating frequency in a balanced mixer. The injection frequency is generated in a crystal oscillator. The crystal frequency is 14 mc below the customer's operating frequency. The crystal oscillator output is coupled to a buffer stage and is mixed with the modulated $14-\mathrm{mc}$ signal in the balanced mixer. The balanced mixer output is limited and amplified to the 10 -watt $\mathbf{r - f}$ output level. The output impedance of the A830-2 is between 50 and 70 ohms.

The power supply for the $A 830-2$ is of conventional design and supplies operating voltages for the vacuum tubes and transistors in the A830-2.

The final power amplifier consists of a single ceramictype tetrode tube. The tube is operated as a class $C$ amplifier with a tuned-cavity plate circuit. The output from the power amplifier is fed through a harmonic filter which reduces all output harmonics.

The harmonic filter consists of two series resonant "M-derived" low pass end sections and a "constant $K$ " " $T$ " center section. The harmonic filter starts to attenuate above 110 megacycles and reaches maximum attenuation at the carrier second harmonic. The attenuation pattern then tapers off slowly as the frequency rises. The over-all result of the harmonic filter is in keeping the harmonics attenuated at least 73 db below the carrier frequency.

### 4.3 Control Circuits.

The 230 -volt a-c single phase power input is stepped down to 115 volts a-c by transformer T301. This lower voltage is used to activate relays in the transmitter control circuits and is also fed to the exciter as its primary power source. The control circuits allow power to be applied to the transmitter only in the proper sequency to prevent damage to the final amplifier. These circuits also contain protective devices to prevent damage to components from accidental overloads. All meter circuits are bypassed to eliminate damage from r-f energy.

### 4.4 Plate Contactor and Plate Power Supply.

The plate contactor consists of a heavy duty relay which controls the 230 -volt a-c primary power to the plate power supply. The plate contactor is actuated by the PLATE ON switch through the control circuitry.

The plate power supply consists of a step-up transformer, a full wave bridge rectifier, anda filter. The power supply is capable of delivering 2,700 volts $d-c$ at 850 ma to the power amplifier.

### 4.5 Control Grid Bias Supply.

The control grid bias supply is a conventional halfwave type with an adjustable output. The supply is fused for protection and, in addition, is an aid in neutralization of the power amplifier. The bias supply output is approximately a negative 48 volts.


TABLE 5-1. ABBREVIATED TUNING PROCEDURES

| CONTROL | POSITION | ADJUSTMENT | INDICATING METER | INDICATION | NOTES <br> Allow transmitter to warm up at least 15 minutes before tuning. |
| :---: | :---: | :---: | :---: | :---: | :---: |
| S101 | MIXER GRID | BUF TUNE | M101 | Maximum |  |
| S101 | V428 B | *L429, L430 | M101 | Maximum |  |
| S101 | V429 B | *L431, L432 | M101 | Maximum |  |
| S101 | V430 B | *L433, L434 | M101 | $\frac{\text { Minin: }}{M A X:=}$ |  |
| S101 | V430C B | PA (exciter) <br> PLATE | M 101 | Minimum |  |
| MULTIMETER | $\begin{aligned} & \text { GRID FS } \\ & 4 \text { MA } \end{aligned}$ | PA MATCH GRID TUNING | MULTIMETER | Maximum |  |
| MULTIMETER | GRID FS | GRID COUPLING | MULTIMETER | 0.5 MA (approx) |  |
| WATTMETER | FORWARD | PLATE TUNING <br> OUTPUT <br> COUPLING <br> POWER OUTPUT ADJUST | P. A. PLATE CURRENT <br> R. F. WATTMETER <br> R.F. WATTMETER | Near min. <br> Near 1000 watt indication <br> Near 1000 watt indication | Repeat the adjustment of PLATE TUNING, OUTPUT COUPLING, and POWER OUTPUT ADJUST controls until 1000 watts is achieved by the indirect power measuring method. $P=I_{p} E_{p} K$ |
| *Use slotted nonmetallic screwdriver on these adjustments. <br> ** Check frequency and adjust if necessary with VHF OSC FREQ ADJ control. |  |  |  |  |  |

adjusting tool. Keep all metallic tools and the hands or other parts of the body away from transistor cases. When disabling the afc, and the complete transmitter is in operation, check the station monitor to be sure the center frequency stays within the FCC requirements.
a. Remove transistor Q509, and place a vtvm from TP501 to ground. Tune L505 for a peak indication on the vtvm, tune L504 for a peak indication. Be sure to tune the inductances in the order given to minimize the limiting effect. Check that each stage is limiting when making these adjustments. Limiting will show up as a broad flat peak on the vtivm when tuning. Set the controls midway between the limiter fall-off points shown on the vtvm. Do not replace Q509.
b. Tune the afc discriminator by placing a vtrm from TP501 to ground, and tune the DISCR PRI control, C515, for a maximum indication. Place the vtvm from TP502 to ground, and adjust the DISCR SEC control, C518, for zero on the vtym.
c. Disable the afc by pressing the AFC DISABLE switch, and adjust the OSC FREQ control until the station monitor indicates that the exciter is on frequency. Replace Q509.
d. Remove Q607. Place a vtvm between TP602 and ground. Tune L606 and L603 for a maximum indication on the vtvm. Tune the inductors in the order shown to minimize the effects of the limiter.
e. Remove afc by holding down the AFC DISABLE switch. With the vtvm from TP602 and ground, tune the DISCR PRI control, C639, for a maximum indication on the vtvm. Move the vtvm to TP601 and tune the DISCR SEC control, C644, for a zero indication when the modulation monitor indicates the exciter is approximately on frequency. Repeat the tuning of the DISCR PRI and DISCR SEC controls. Replace Q607.
f. Place a d-c vtvm between TP603 and ground. Set the AMP BIAS control for a 7.5 -volt indication on the vtvm.
g. Place a vtvm between TP504 and ground. Remove Q510. Tune L611 and L608 for a maximum indication on the vtvm. Replace Q510.
h. With a vtum on TP504, adjust the REF LEVEL control for an equal indication on the vtvm with first Q510 removed and then Q509 removed. (This equalizes the modulator oscillator voltage and the 14-megacycle reference voltage.) Replace the transistors.
i. Place an oscilloscope between TP503 and ground. Apply a 150 -cps audio signal on J601. Adjust the MOD

TABLE 5-2. DISTORTION CHECKS

| FREQUENCY | DISTORTION IN PERCENT |  |  |
| :---: | :---: | :---: | :---: |
|  | $25 \%$ MODULATION | $50 \%$ MODULATION | $100 \%$ MODULATION |
| 50 |  |  |  |
| 100 |  |  |  |
| 400 |  |  |  |
| 1,000 |  |  |  |
| 7,000 |  |  |  |
| 7,500 |  |  |  |
| 10,000 |  |  |  |
| 15,000 |  |  |  |

BALANCE control for a minimum 150-cps indication as shown on the oscilloscope.

## NOTE

The MOD BAL control must be adjusted slowly to allow the error signal amplifiers to stabilize between adjustments.

### 5.4. Distortion Testing Procedure.

a. Refer to figure 5-1. Connect an audio frequency signal generator, such as a Hewlett-Packard Model 600 D , to the exciter audio input, terminals 1 and 2 of TB305. (Disconnect the station console audio input leads when making this comnection.) Comect a distortion and noise meter, such as a Hewlett-Packard

Model 330D, to the broadcast monitor. Comnect a 50ohm artificial load to the r-f output connector located on top of the transmitter cabinet. Turn on the transmitter.
b. Apply a 50-cps audio tone to the transmitter input. The input level should be such that the voltage at the transmitter audio input terminals is $+10 \pm 2 \mathrm{dbm}$.
c. Adjust the exciter MOD GAIN control until the monitor indicates 100 percent modulation ( $\pm 75$ kilocycle deviation).
d. Measure the distortion at the frequencies and modulation levels given in table 5-2. The distortion shall be less than 1.5 percent for frequencies between 50 and 100 cps , less than 1.0 percent for frequencies between 100 and $7,500 \mathrm{cps}$, and less than 1.5 percent for frequencies between 7,500 and $15,000 \mathrm{cps}$.


Figure 5-1. Distortion Test Setup

### 5.5 Audio Frequency Response Measurements.

a. Refer to figure 5-2. Connect an audio frequency signal generator, such as a Hewlett-Packard Model 600 D , to terminals 1 and 2 of terminal board TB305. (Disconnect the station console audio input leads when making these measurements.) Connect a vacuum tube voltmeter, such as a Ballantine Model 310A, to the audio output terminals of the audio frequency generator. Connect a 50 -ohm artificial load to the r-f output comnector located on top of the transmitter cabinet. Turn on the transmitter.
b. Check the audio frequency response of the transmitter by modulating the transmitter at $50,100,400$, $1,000,5,000,7,500,10,000$, and $15,000 \mathrm{cps}$ for 25 percent, 50 percent, and 100 percent modulation. Audio frequency response is measured by keeping the percentage of modulation constant and measuring the magnitude of audio, at each frequency given, to give the desired percentage of modulation. The audio frequency response must fall within the limits given in figure 5-3.

## NOTE

When taking audio frequency response measurements, a broadcast monitor, such as a Hewlett-Packard Model 335B, should be used. Do not use an instrument where audio de-emphasis might give a false indication of peak modulation.

### 5.6 FM Noise Measurement.

a. Refer to figure 5-4. Connect an audio frequency signal generator, such as a Hewlett-Packard Model 600D, to terminals 1 and 2 of terminal board TB305. (Disconnect the station console audio input leads when making this measurement.) Connect a vacuum tube voltmeter to the output terminals of the broadcast
monitor. Connect an artificial load to the r-f output connector located on top of the transmitter cabinet. Turn on the transmitter.
b. Modulate the transmitter 100 percent ( $\pm 75$ kilocycles deviation) with 400 cps of audio.
c. Remove the modulating 400 cps , and read the residual $F M$ noise on the vacuum tube voltmeter. The residual FM noise shall be less than -65 db below 100 percent modulation.

### 5.7 AM Noise Measurements.

a. Refer to figure 5-5. Short out terminals 1 and 2 of terminal board TB305. Connect a vacuum tube voltmeter to connector J 3 of the Hewlett-Packard Model 335B broadcast monitor. Comect a $50-\mathrm{ohm}$ artificial load to the r-f output connector located on top of the transmitter cabinet. Turn on the transmitter.
b. Switch the broadcast monitor to measure carrier level.
c. Measure the $A M$ noise in db at J 3 of the broadcast monitor in the following manner. Set modulation monitor to CARRIER LEVEL, and measure the d-c level on the modulation meter ( 100 percent on scale equals 10 volts). Connect the vacuum tube voltmeter to J3, and terminate J3 with a two-megohm resistor. Measure the a-c level on the vacuum tube voltmeter. (The input to the vacuum tube voltmeter should be a shielded cable having less than 100 uuf distributed capacitance.) The AM noise is the direct ratio of the $d-c$ reading and the $a-c$ level. The AM noise shall be not less than -55 db below voltage or $\mathrm{d}-\mathrm{c}$ carrier level.

$$
\text { AM Noise }=20 \log \frac{D-C \text { reading }}{A-C \text { reading }}
$$

### 5.8 Troublar Shooting.

Standard trouble-shooting procedures should be used in finding malfunctions in the transmitter. Meter indications for all functions should be


Figure 5-2. Audio Frequency Response Test Setup


Figure 5-3. Audio Frequency Response Limits

i

Figure 5-4. FM Noise Test Setup
recorded when the transmitter is installed and operating properly. Table $5-3$ is supplied for recording these readings. If some malfunction should occur after the normal meter readings are recorded, it is a simple matter to compare the meter readings of the malfunctioning equipment with the normal meter readings. When trouble shooting and comparing the meter readings, it is advisable to start with the final stage and proceed backwards until normal readings are encountered. The malfunctioning stage will then be the one immediately ahead of the normal meter indications.
As most cases of trouble will be traced to tubes or transistors, it is advisable to first of all replace the tube (or transistor) in the stage in which the trouble is suspected. If the trouble does not clear with tube or transistor replacement, it will become necessary to take resistance or voltage measurements, within the suspected circuit, to determine which component has failed.
When tracing trouble within the power amplifier, it will be helpful to use the "from-to" information given in the D830-1 1000 Watt FM Power Amplifier Unit Instructions.

The "from-to" information gives the actual location of the individual wires within the power amplifier cabinet. Whenused in conjunction with the schematic, the 'from-to' information can be very helpful.
If the transmitter center frequency shifts excessively with modulation, the trouble may be isolated to either the afc circuitry or the modulator circuitry by disabling the afc and noting if the carrier shifts more than 1.8 kc with a change in modulation from zero to 100 percent. If the modulator oscillator shifts more than the 1.8 kc with the afc disabled, the trouble will be within the modulator oscillator circuits. The afc circuitry camnot shift the modulator oscillator frequency more than 1.8 kilocycles. If the carrier shift is under 1.8 kc the trouble will be in the afe circuitry.

If the power amplifier tube, V201, is replaced, the stage may no longer be neutralized. Check neutralization of the stage before going ahead with the neutralization procedure given in paragraph 2.6.1. It may not be necessary to change the neutralization adjustment.

TABLE 5-3. NORMAL TRANSMITTER METER INDICATIONS

| CONTROL | POSITION | METER | INDICATION |
| :---: | :---: | :---: | :---: |
| S101 | BUFFER GRID A | M101 |  |
| S101 | MIXER GRID A | M101 |  |
| S101 | V428 B | M 101 |  |
| S101 | V429 B | M101 |  |
| S101 | V430 B | M101 |  |
| S101 | V430C B | M101 |  |
| S101 | MOD OUTPUT B | M101 |  |
| S101 | AFC KEY B | M101 |  |
| S101 | 14 MC REF B | M101 |  |
| MULTIMETER | SCREEN FS 400 VDC | MULTIMETER |  |
| MULTIMETER | SCREEN FS 40 MA | MULTMETER |  |
| MULTIMETER | GRID FS 4 MA | MULTIMETER |  |
|  |  | P.A. PLATE VOLTAGE |  |
|  |  | P.A. PLATE CURRENT |  |
| WATTMETER | FORWARD 1.5 KW | R.F. WATTMETER |  |
| WATTMETER | REFLECTED 1.5KW | R.F. WATTMETER |  |



Figure 5-5. AM Noise Test Setup

E I M A C
Division of Varian
8183

SAN CARLOS
CALIFORNIA

CERAMIC POWER TETRODE

The EIMAC 8168/4CX1000A is a ceramic and metal, forced-air cooled, radial-beam tetrode with a rated maximum plate dissipation of 1000 watts. It is a low-voltage, high-current tube specifically designed for Class- $\mathrm{AB}_{1}$ rf linear-amplifier or audio-amplifier applications where its high gain and low distortion characteristics may be used to advantage. At its rated maximum plate voltage of 3000 volts, it is capable of producing 1630 watts of peak-envelope output power. Two $8168 / 4 \mathrm{CX} 1000 \mathrm{As}$ operating in Class- $\mathrm{AB}_{1}$ will produce 3260 watts of audio power.

## general characteristics




*In shielded fixture.

## MECHANICAL

Base - - - - - - - - - -
Maximum Operating Temperatures:



AUDIO AMPLIFIER OR MODULATOR<br>Class $A B_{1}$

MAXIMUM RATINGS

| DC PLATE VOLTAGE | - | - | - | 3000 VOLIS |
| :--- | :---: | :---: | :---: | ---: |
| DC SCREEN VOLIAGE | - | - | - | 400 VOLTS |
| DC PLATE CURRENT - | - | - | - | 1.0 AMP |
| PLATE DISSIPATION - | - | - | - | 1000 WATTS |
| SCREEN DISSIPATION | - | - | - | 12 WATTS |
| GRID DISSIPATION - | - | - | - | 0 WATTS |

TYPICAL OPERATION
(Sinusoidal wave, two tubes unless noted)

| DC Plate Voltage | 2000 | 2500 | 3000 | volts |
| :---: | :---: | :---: | :---: | :---: |
| DC Screen Voltage | 325 | 325 | 325 | volts |
| DC Grid Voltage ${ }^{1}$ | -60 | -60 | -60 | volts |
| Zero-Signal DC Plate Current | 500 | 500 | 500 |  |
| Max.-Signal DC Plate Current | 1.78 | 1.77 | 1.75 | amps |
| Zero-Signal DC Screen Current* | 16 | 12 | 10 | mA |
| Max-Signal DC Screen Current* | 70 | 70 | 70 | mA |
| Effective Load, Plate to Plate | 2040 | 2850 | 3680 | ohms |
| Driving Power | 0 | 0 |  | watts |
| Max-Signal Plate Output Power | 1860 | 2600 | 3260 | watts | of plate current without drawing grid current may vary somewhat from the typical values shown.

## APPLICATION

## MECHANICAL

Cooling - Sufficient cooling must be provided for the anode and ceramic-to-metal seals to maintain operating temperatures below the rated maximum values:

$$
\begin{array}{ll}
\text { Ceramic-to-Metal Seals } & 250^{\circ} \mathrm{C} \\
\text { Anode Core } & 250^{\circ} \mathrm{C}
\end{array}
$$

A flow rate of 25 cubic feet per minute will be adequate for operation at maximum rated plate dissipation at sea level and with inlet air temperatures up to $40^{\circ} \mathrm{C}$. Under these conditions, 25 cfm of air flow corresponds to a pressure difference across the tube and socket of 0.2 inch of water column. Experience has shown that if reliable long-life operation is to be obtained, the cooling air flow must be maintained during standby periods when only the heater voltage is applied to the tube.
At higher altitudes and at VHF increased air flow will be required. For example, at an altitude of 10,000 feet, a flow rate of 37 cfm will be required and will be obtained with a pressure drop across tube and socket of 0.3 inch of water column. In selecting a blower for use at high altitudes, care must be taken to assure that the blower is designed to deliver the desired volume of air at the corresponding pressure drop and at the particular altitude.

In cases where there is any doubt regarding the adequacy of the supplied cooling, it should be borne in mind that operating temperature is the sole criterion of cooling effectiveness. Surface temperatures may be easily and effectively measured by using one of the several tempera-ture-sensitive paints or sticks available from various chemical or scientific-equipment suppliers. When these materials are used, extremely thin applications must be made to avoid interference with the transfer of heat from the tube to the air stream, which would cause inaccurate indications.

The 4CX1000A is tested for vibration (noise) from 10 Hz to 500 Hz . Vibration level is 10 G
units peak 28 Hz to 500 Hz . Below 28 Hz vibration double amplitude is .25 inch.

The 4CX1000A is tested for shock, 50 G, 11 ms, three axes, after which the tube must be within specification for grid bias voltage and gas current.

## ELECTRICAL

Heater - The rated heater voltage for the $4 \mathrm{CX1000} \mathrm{~A}$ is 6.0 volts. The voltage, as measured at the socket, should be maintained at this value to minimize variations in operation and to obtain maximum tube life. In no case should the voltage be allowed to exceed $5 \%$ above the rated value.

The cathode and one side of the heater are internally connected.

It is recommended that the heater voltage be applied for a period of not less than 3 minutes before other operating voltages are applied. From an initial cold condition, tube operation will stabilize after a period of approximately 5 minutes.

Control-Grid Operation - The grid dissipation rating of the 4CX1000A is zero watts. The design features which make the tube capable of maximum power operation without driving the grid into the positive region also make it necessary to avoid positive-grid operation.

Although the average grid-current rating is zero, peak grid currents of less than five-milliamperes as read on a five-milliampere meter may be permitted to flow for peak-signal monitoring purposes.

Screen-Grid Operation - Tetrode tubes may exhibit reversed screen current to a greater or lesser degree depending on individual tube design. This characteristic is prominent in the 4CX1000A and, under some operating conditions, indicated negative screen currents in the order of 25 milliamperes may be encountered.

The maximum rated power dissipation for the screen grid in the $4 \mathrm{CX1} 1000 \mathrm{~A}$ is 12 watts and
the screen power should be kept below this level. The product of the peak screen voltage and the indicated dc screen current approximates the screen input power except when the screen current indication is near zero or negative. In the usual tetrode amplifier, where no signal voltage appears between cathode and screen, the peak screen voltage is equal to the dc screen voltage. Experience has shown that the screen will operate within the limits established for this tube if the indicated screen current, plate voltage and drive voltage approximate the "Typical Operation" values.

The screen supply voltage must be maintained constant for any values of negative and positive screen currents that may be encountered. Dangerously high plate currents may flow if the screen power supply exhibits a rising


| DIMENSION DATA |  |  |  |
| :---: | :---: | :---: | :---: |
| AEF. | NOM. | MIN. | MAX. |
| A |  | 3.335 | 3.365 |
| B |  | . 807 | . 817 |
| C |  | 1.870 | 1.900 |
| D |  | 2.250 DIA. | $2.300 \mathrm{D14}$. |
| E |  | 2.195 | 2.380 |
| $F$ |  | 3.410 | 3.550 |
| G |  | 4.600 | 4.800 |
| H |  | . 950 | 1.000 |
| J |  | . 675 | . 725 |
| K |  | 400 | . 450 |
| L |  | . 140 | . 170 |
| M |  | . 020 | . 030 |
| N |  | . 700 | . 800 |
| P |  | . 314 DIA | 326 DIA. |
| A |  | $55^{\circ}$ | $65^{\text {a }}$ |
| 5 |  | 1150 | $125^{\text {a }}$ |
| T |  | . 470 | . 530 |
| U |  | . 023 | . 043 |
| V |  | . 057 DIA. | . 073014. |



NEUTRALIZATION: Turn off the exciter; remove the bias fuse (F304) and observe that the plate current rises to full scale. The overload should cut the power off. Also observe that as the plate current rises that no power output is observed on the watt meter. If the overcead does not cut off the power, immediately push the plate off button $A \mathbb{A}$ maximum current. If power is observed it means the PA has a tendency to oscillate; therefore, move the nfokeiization bars apart slightly. Repeat as above. If the PA is stable, but the grid drive drops off considerably as the plate voltage is applied, it indicates over. neutralization and the bars should be moved slightly closer together. Repeat as necessary for optimum neutralization.

Screen Diode Protection Modification

$0.47 \mu f^{\prime}$ booVolt capacilon is jied across p509 (15R2)

0

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## SECTIONI GENERA DESCRIPTION

### 1.1 PURPOSE OF INSTRUCTION BOOK.

Unit Instructions TD-567 provides information about D830-1 1000-Watt FM Power Amplifier. Information which is furnished covers a general description of the equipment, principles of operation, maintenance procedures, and a parts list.

### 1.2 PURPOSE OF EQUIPMENT.

The D830-1 1000-Watt FM Power Amplifier can be used for continuous monaural or multiplex and SCA FM broadcast service on a single frequency, in the range from 88 to 108 megacycles with an exciter input of 10 watts and an output power of 1000 watts.

### 1.3 DESCRIPTION OF EQUIPMENT.

### 1.3.1 PHYSICAL DESCRIPTION.

The D830-1 1000-Watt FM Power Amplifier, shown in figure $1-1$, is contained in a single cabinet that is 38 inches wide, 76 inches high, 27 inches deep, and weighs approximately 767 pounds. All D830-1 operating controls are located behind the doors on the front of the cabinet. The filament and plate on-off controls and four monitoring meters are located at the top front of the cabinet. The meters may be observed easily while operating the tuning controls. The D830-1 uses one tube (the power amplifier), plus semiconductors for voltage rectification and regulating. The power amplifier tube is accessible from the front of the transmitter. The bottom front of the D830-1 cabinet is removable to allow access to components on the bottom of the inside panel.

Large doors at the upper rear of the cabinet (see figure 1-2) allow access to the upper part of the D830-1 for servicing and maintenance. Operating personnel are protected by both electrical andmechanical interlocks on the rear doors and panel. These interlocks remove the D830-1 plate voltage and ground the high voltage circuits when the doors are opened or the panel is removed. The power amplifier plate-tuning cavity is located in an interlocked compartment at the front of the D830-1.

Inside the D830-1, heavy iron-core components are at the bottom of the cabinet. A standard 19 -inch rack is provided for mounting the 10 -watt exciter within the confines of the D830-1. Room is also provided on the standard 19 -inch rack, for the mounting of SCA generators. An optional harmonic filter which can be connected to the D830-1 output is suspended from the top of the cabinet.


Figure 1-1. D830-1 1000-Watt FM Power Amplifier, Over-all View

Cooling air for the D830-1 is drawn through a permanent-type air filter at the rear of the cabinet by a low-speed, high-volume fan, and exhausted through a shielded opening in the top of the cabinet. A pressure blower supplies cooling air directly to the power amplifier tube.


Figure 1-2. D830-1 1000-Watt FM Power Amplifier, Rear View with Bottom Panel Removed

### 1.3.2 ELECTRICAL DESCRIPTION.

The D830-1 1000-Watt FM Power Amplifier consists of a single air-cooled, power amplifier tube capable of being driven to full power by a 10 -watt exciter. All associated power supply and control circuitry is included for operation of the D830-1. D830-1r-finput impedance is 50 ohms nominal, unbalanced. D830-1 output power is at least 1000 watts over the frecuency range of 88 to 108 mc into a $250-\mathrm{ohm}$ load with an swr not exceeding $2: 1$.

Line power input required is $50 / 60$-cycle, singlephase, with primary taps on all power transformers to compensate for line voltage variations from 200 to 250 volts. Circuit breakers in the input side of the line are provided for primary currentoverload protection. The control circuits, the final amplifier filament, and the central grid bias supply are fused. A time delay relay provides protection of the power amplifier tube during warmup. Remote control canbe exercised over
filament on, filament off, plate on, and plate off functions of the power amplifier. Instruction books covering the exciters used in conjunction with D830-1 1000-Watt FM Power Amplifier are listed in table 1-1.

TABLE 1-1
ASSOCIATED EQUIPMENT INSTRUCTION BOOKS

| ASSOCIATED <br> EQUIPMENT | INSTRUCTION BOOK <br> NUMBER |
| :--- | :---: |
| A830-2 10 W Wide-Band <br> FM Broadcast Exciter <br> 786M-1 Stero <br> Generator | TD-536 |

### 1.9 EQUIPMENT SUPPLIED.

Table 1-2 lists equipment that is supplied as part of D830-1 1000-Watt FM Power Amplifier.

TABLE 1-2
EQUIPMENT SUPPLIED WITH D830-1 1000-WATT FM POWER AMPLIFIER

| EQUIPMENT | COLLINS <br> PART NUMBER |
| :---: | :---: |
| D830-1 1000 -Watt FM Power <br> Amplifier | $522-2948-00$ |

### 1.5 EQUIPMENT REQUIRED BUT NOT SUPPLIED.

Table 1-3 lists equipment that is required for operation of D830-1 1000-Watt FM Power Amplifier but not supplied as part of the power amplifier.

TABLE 1-3
EQUIPMENT REQUIRED BUT NOT SUPPLIED AS PART OF D830-1 1000-WATT FM POWER AMPLIFIER

| EQUIPMENT | COLLINS <br> PART NUMBER |
| :---: | :---: |
| A830-2 10 W Wide-Band FM <br> Broadcast Exciter | $522-2714-00$ |

### 1.6 ACCESSORY EQUIPMENT.

Table 1-4 lists accessory equipment that is available for use with D830-1 1000-Watt FM Power Amplifier.

TABLE 1-4 ACCESSORY EQUIPMENT

| EQUPMENT | COLLINS <br> PART NUMBER |
| :--- | :---: |
| 250-Watt/1-Kw Harmonic Filter <br> (used only if the D830-1 is fed <br> directly to an antenna and not to <br> a higher power amplifier). | $549-2010-000$ |

### 1.7 EQUIPMENT SPECIFICATIONS.

### 1.7.1 MECHANICAL.

Weight . . . . . . . 767 pounds maximum.

Size . . . . . . . . 38 inches wide, 76 inches | high, 27 inches deep. |
| :--- |

Ventilation . . . . . . One ventilating fan, one

blower. | Ambient temperature |
| :--- |
| range . . . . . |

Altitude . . . . . . . 0 to 6000 feet.

### 1.7.2 ELECTRICAL.

Power source . . . . 200 to 250 volts, 50/60cycle, single-phase. (When using the D830-1 as a driver, connect TB301-1
and TB301-2 across one phase of the three phase primary power source.)

| Maximum power requirements - | 3300 watts. |
| :---: | :---: |
| R-f input power | 10 watts nominal. |
| Input impedance | 50 ohms nominal, unbalanced. |
| Power output | 1000 watts nominal. |
| Output impedance | 50 ohms, unbalanced. Maximum swr 2:1. |
| Frequency range | 88 to 108 mc . Exact operating frequency determined by frequency of exciter. |

### 1.8 TUBE AND SEMICONDUCTOR COMPLEMENT.

Table 1-5 lists the tube and semiconductor complement supplied as part of the D830-1 1000-Watt FM Power Amplifier.

TABLE 1-5
TUBE AND SEMICONDUCTOR COMPLEMENT

| QUANTITY | TYPE | FUNCTION |
| :---: | :--- | :--- |
| 1 | 4CX1000A | R-f power amplifier |
| 1 | 50 M 140 ZB 5 | Voltage regulation |
| 80 | 1N1566 | H.V. rectifier diodes |
| 1 | 1N1492 | Bias rectifier |
| 1 | 1N1492 | Grid current diode |
| 1 | 1N1566 | Meter protection diode |

## SECTION II <br> PRINCIPLES OF OPERATION

### 2.1 GENERAL.

The D830-1 1000-Watt FM Power Amplifier consists of a power amplifier and associated control circuitry, for the amplification of 10 watts of exciter drive to 1000 watts of r-f power. The D830-1 output can be used to drive a higher power amplifier or fed through a harmonic filter to an antenna.

Refer to figure 2-1, a block diagram of D830-1 1000Watt FM Power Amplifier. The 230 volts a-c is fed to a control circuit power transformer which reduces the input voltage to 115 volts a-c for use by the power amplifier blower and control circuits. A second regulated transformer located within the control circuitry reduces the 230 volts a-c to 6.3 volts for use by the
power amplifier filament. The 230 volts a-c is also fed to a transformer which supplies the power amplifier plate and screen voltages. The primary power to the plate power supply is controlled by the plate contactor. The control circuits provide cabinet interlocks for protection of personnel from all high voltage, local or remote filament on and filament off controls, local or remote plate on and plate off controls, and a time delay to prevent the application of high voltage before the power amplifier filament has heated sufficiently. Provisions are also available within the control circuitry for connection to the control circuitry of a higher power amplifier. The higher power amplifier could then control the D830-1 plate and filament power. Exciter input power is applied directly to the power amplifier where it is raised to 1000 -watts of $r$-f power.


Figure 2-1. D830-1 1000-Watt FM Power Amplifier, Block Diagram

The power amplifier consists of a forced air cooled, ceramic tetrode, V201. The plate of V201 is connected to a tuned cavity (foreshortened $\frac{\lambda}{4}$ coaxial line resonator). The output from the tuned cavity can then be fed to a higher power amplifier or to a harmonic filter and antenna if high power amplification is not necessary. A sample of the power amplifier output is taken from the plate tuned cavity for monitoring purposes.

Metering circuits are provided for the power amplifier plate current and voltage, screen current and voltage, grid current, and output power.

### 2.1.1 CONTROL CIRCUITS.

Refer to figure 2-2, a simplified schematic diagram of the control circuits of D830-1 1000-Watt FM Power Amplifier. The primary purpose of the control circuitry is to provide filament and plate on and off control. The power amplifier filament is turned on in the following manner: When the momentary FILAMENT ON switch, S112, is pressed, a ground is placed on filament control relay K301. As 115 volts a-c is present on terminal 2 of relay K301 from 115-volt a-c supply transformer T301, the filament control relay will be energized. This closes contacts 3 and 4 holding the relay in the energized position after the momentary FILAMENT ON switch, S112, is released. The green filament indicator lamp, DS301, will light. The filament control relay, K 301 , when energized, will close relay contacts which will start cabinet fan B103, supply 230 volts to the regulated filament transformer, supply 115 volts a-c to the 10 -watt exciter, and supply 115 volts a-c to the power amplifier bias supply. When blower B301 comes up to speed, the air interlock switch, S314, will close, applying 115 volts a-c to time delay K307. In 4 to 5 minutes (time for power amplifier filament V201 to warm up) time delay K307 will
close, energizing auxiliary relay K 306 , which through contacts 3 and 5 supplies 115 volts a-c to plate contactor relay K304. The plate can now be turned on by pressing momentary PLATE ON switchS113 (providing all interlocks are closed). Pressing the momentary PLATE ON switch places a ground on plate control relay K302. Relay K302 will then energize, closing contacts 3 and 4 holding K302 in the energized position. Contacts 9 and 10 will also close providing a ground path energizing plate contactor K304, and lighting the red plate on indicator. When the plate contactor closes, the primary a-c supply will be applied to the plate and screen supply transformer T303. The plate power supply will then furnish approximately 2700 volts d-c to power amplifier tube V201.

The momentary FILAMENT OFF switch, S111, removes power from the control and power amplifier circuits. It accomplishes this by opening the 115 -volt a-c lead to filament control relay K301 and plate control relay K302. This de-energizes plate contactor K304 and blower interlock S314, removing filament, plate, and screen voltage. The time delay relay will begin to recycle and will always cause the correct time delay necessary for proper filament heating.

The momentary PLATE OFF switch, S114, removes only plate and screen voltage from power amplifier V201. This is accomplished by momentarily opening the ground lead of plate control relay K302, which deenergizes plate contactor K304, removing 230-volt a-c power from the plate and screen supply.

Plate overload protection is provided by overload relay K305. As the plate current increases, the current through K305 will increase until overload relay K305 energizes, opening the ground lead of plate control relay K302, removing plate voltage. Potentiometer



R304 adjusts the point at which excess plate current will activate overload relay K 305 .

### 2.1.2 POWER AMPLIFIER CIRCUITS.

Refer to figure 2-3, a simplified schematic of the power amplifier circuit of D830-1 1000-Watt FM Power Amplifier. The power amplifier consists of a forced air-cooled, tetrode amplifier, operating over the standard frequency modulated broadcast band of 88 to 108 mc .

The power amplifier plate and screen voltages are obtained from a full-wave bridge rectifier circuit. Transformer T303 increases the 230 -volt single-phase $50 / 60$-cps primary input to approximately 3000 volts rms across terminals 8 and 9 (T303 secondary) winding. The primary winding of transformer T303 has six adjustable taps to compensate for line voltages from 200 to 250 volts. The output voltage of the secondary winding of T303 is fed to a conventional full-wave bridge and filter circuit rectifier consisting of diodes CR301 through CR380, inductances L301 and L302, and capacitors C301 and C302. Resistors R340 through R419, and capacitors C320 through C399 are used to equalize the forward currents of all diodes. Resistor R320, R327, R328, R331, and R332, and capacitor C303 form a transient suppressor network which helps suppress transient voltages formed when power is first applied to transformer T303 and when power is switched off. The 2750 -volt d-c output from the power supply is fed through P.A. PLATE CURRENT meter M303, through an r-f filter consisting of C210, C217, C222, L202 and R336 to the plate of V201. Plate voltage and plate current are read directly from P.A. PLATE VOLTAGE meter M302 and P.A. PLATE CURENT meter M303 respectively.
The screen voltage is obtained from the 2750 -volt plate supply. The plate supply is fed to a combination bleeder resistor and voltage divider consisting of resistors R305, R324, R309, R315, R317, R318, and R319. Voltage regulation for the screen supply of V201 is accomplished by passing most of the bleeder current through two Zener diodes placed in parallel with R309. If the screen current should fluctuate, the screen voltage is held to 250 volts by the two Zener diodes, CR381 and CR382. Capacitor C212 is placed from the screen grid to ground to shunt any r-f energy present on the screen grid. Screen voltage, and thus the power output of V201, is made variable by POWER OUT PUT ADJUST control R317. A protective device consisting of two carbon blocks (which will arc over if the screen voltage exceeds 400 volts) is provided to protect the screen supply Zener diodes in the event of a screen-to-plate short.

Power amplifier V201 is a grounded cathode tetrode, using fixed and grid leak bias. The control gridcircuit of V201 consists of a parallel-tuned resonant tank circuit, Z202 and C213. R201, R205, R206, and R207 placed in parallel with C213 present the proper load to the exciter and also provide a broad-band lowimpedance input to the control grid. Z 202 consists of two parallel rods, forming a shorted stub which is tuned by a shorting bar between the two rods. Control
grid fine tuning is accomplished by capacitor C213. Negative fixed bias for the control grid is obtained from a half-wave power supply consisting of diode CR383, capacitors C383 and C384, and resistors R325, R326, and R330. The fixed bias is made adjustable by adjustable resistor R326. Power to operate the bias supply is obtained from T301 through the filament control relay, K301. Grid leak bias is obtained from the voltage drop across R329 due to grid current. Diode CR384 is placed across R329 to prevent the possibility of the grid going positive because of grid emission.

The plate cavity is formed by a short section of coaxial transmission line resonating with the plate capacity of V201 and plate tuning capacitor C209. The coaxial transmission line is roughly tuned initially by adjusting a shorting plate which lengthens or shortens the physical length of the coaxial transmission line. A resistor, R208, is attached to the tank slider to provide parasitic swamping. Fine plate tuning is accomplished by C209. The output coupling network, formed by C208 and L203, is coupled to the coaxial transmission line adjacent to the plate of V201 for correct impedance matching. Inductance L203 and capacitor C220 act as an L section low-pass filter for frequencies above 130 mc to provide a measure of harmonic suppression. A monitor output is provided in the plate cavity for use by the station program monitor.

Neutralization of V201 is accomplished by two small adjustable bars which are connected in series with the screen by-pass capacitor connected to the tube socket. These bars form two parallel inductances that adjust the over-all screen reactance, bringing the tube internal reactances into balance for neutralization.

A MULTIMETER is located on the front panel of the power amplifier to enable the station operator to monitor screen voltage, screen current, and grid current. The MULTIMETER functions are selected by the multimeter switch located within the power amplifier cabinet. Screen voltages are determined by connecting multimeter M301 and meter multiplier resistor R322 across the regulated screen supply with switch S307. Screen current is determined by placing M301 and shunt resistor R307 in series with the screen voltage line. Grid current is determined by placing M301 and shunt in series with the control grid bias input.

### 2.2 CONTROL FUNCTIONS.

The following paragraphs describe all the functions of controls in D830-1 1000-Watt FM Power Amplifier. Refer to figure 2-4 for control locations.

The controls located directly on the front panel under the meters include the FILAMENT ON, FILAMENT OFF, PLATE OFF, and PLATE ON controls. The filament on and plate on indicators are placed in line with the above mentioned controls. The FILAMENT ON switch, S112, energizes the power


Figure 2-4. D830-1 1000-Watt FM Power Amplifier, Control Locations
amplifier filament, the power amplfier and cabinet blowers, and supplies power to the 10 -watt exciter. The FILAMENT OFF switch, S111, de-energizes all transmitter circuits. The PLATE ON switch, S113, energizes the plate control relay, K302, which in turn could (if the FILAMENT ON switch has not been pressed) energize the filament control relay, K301, starting the power amplifier in sequence.

The PLATE OFF switch, S114, removes plate and screen voltage. The green filament indicator light, DS301, lights when the FILAMENT ON switch is pressed and indicates that voltage is available to the filament control relay. The filament control relay starts the PA blower which activates the PA blower interlock, energizing the power amplifier filament.

The red plate on indicator light, DS302, indicates the plate contacter is receiving voltage.
The following controls are located directly under the left front door on the power amplifier panel. The POWER OUTPUT ADJUST potentiometer, R317, adjusts the power amplifier screen voltage thus changing the output power. The WATTMETER switch, S308, connects the R. F. WATTMETER to either the reflected power or forward power section of the directional coupler. The WATTMETER switch is normally left in the FORWARD position. The MULTIMETER switch, S307, selects either screen voltage, screen current or grid current for the MULTIMETER, M301. Table 3-1 lists the MULTIMETER switch positions and typical indications for each of the three meter circuits.

The following controls are located directly behind the left front door on the power amplifier compartment. The OUTPUT COUPLING control, C208, adjusts the coupling of the load to the plate cavity. The PLATE TUNING control, C209, tunes the plate cavity to resonance and is set very near the minimum indication on the P.A. PLATE CURRENT meter, M303. At this point the power output should be at the peak as indicated on the R.F. WATTMETER, M304. The GRID TUNING control, C213, tunes the grid tank and is set for maximum indication on the MULTIMETER, M301, with the MULTIMETER switch, S307, set to GRID FS 4 MA. The GRID COUPLING control, C221, adjusts the coupling of the grid tank to the exciter output and is normally set for 0.5 ma indication on the MULTIMETER. The following controls are located on the power supply panel directly behind the bottom front panel of the power amplifier cabinet. The LOCAL-REMOTE switch, S302, allows the power amplifier to be operated from a remote position or from the power amplifier. With S302 in the REMOTE position, filament on, filament off, plate on, and plate off functions may be selected from either a remote position or at the power amplifier. With S302 in the LOCAL position, filament on, filament off, plate on and plate off functions may be selected only at the
power amplifier cabinet. The PLATE circuit breaker, CB301, is a protective device which monitors the plate supply transformer primary current. The circuit breaker will activate if the transformer primary current exceeds 12 amperes. The control circuit fuses protect the control circuits from overloads. The two 5 -ampere fuses protect the control transformer primary while the 4 -ampere fuse protects the control transformer secondary. A fourthfuse, $1 / 8$ ampere, protects the control grid bias supply.

The wattmeter adjusting potentiometer, R321, is located directly below the R.F. WATTMETER when the upper switch and meter panel is raised. The wattmeter adjusting potentiometer is set at the factory and does not normally require adjustment.

The overload adjusting potentiometer, R304, is located inside the right rear door on the relay panel. The overload adjusting potentiometer is set at the factory and does not normally require adjustment. The bias adjust control, R326, is located on the rear of the power amplifier panel on the bias supply chassis. The bias adjusting control adjusts the fixed control grid bias. This control is set at the factory and does not normally require adjustment.

## SECTION III <br> MAINTENANCE

### 3.1 GENERAL.

This section contains information concerning the maintenance of D830-1 1000-Watt FM Power Amplifier.

## WARNING

Voltages present in this equipment are dangerous to life. Observe safety precautions when performing any maintenance. Do not reach inside the D830-1 cabinet whenever high voltage is applied. Do not depend entirely on door interlocks. Always shut down the D830-1 before doing any work inside the D830-1 cabinet. Immediately upon opening the rear doors short out all highvoltage points using the shorting stick located inside the left rear door.

### 3.2 PREVENTIVE MAINTENANCE.

Most service interruptions in equipment of this type are caused by dirt and corrosion. Corrosion is accelerated by the presence of moisture and dust. Dust should be removed periodically with a soft
brush or a dry, oil-free air jet. Remove dust as often as a perceptible quantity accumulates at any point in the power amplifier.
When the D830-1 is operated near salt water or in other corrosive atmospheres, inspect and clean interlock switches, cable connectors, tube prongs, and other metal parts more frequently to keep the equipment in good operating condition.

### 3.2.1 AIR FILTER CLEANING.

At least once each month, or more often if needed, clean the air filter according to the following procedure.
a. Remove the air filter from the D830-1 cabinet by loosening the two thumb screws located above the air filter. Slide the air filter to the extreme right and pull the left side of the air filter out as soon as the filter clears the panel. Slide the air filter to the left and remove.
b. Mark with an arrow the direction of air flow.
c. Wash by passing a fine spray of hot water through the filter in the direction opposite that of the airflow. Gently shake the water out of the filter. d. Dip the filter in a water-soluble oil, such as Filter-kote M available from Collins Radio Company, Service Parts Department, Cedar Rapids, Iowa (Collins part number 005-0609-00).
e. Remove the filter from the oil, lay the filter face down until oil ceases to drip from the filter.
f. Replace the filter into the lower rear panel with the air flow arrow (marked when the filter was removed) pointing in the direction of the air flow. Tighten the two thumb screws.
g. Replacement filters are Collins part number 009-1069-00.

### 3.2.2 PA TUBE CLEANING.

The power amplifier tube depends uponastream of air passing through the fins to cool the anode. When these fins become dirty, the air-flow is reduced and the tube life is shortened. The radiator fins should be cleaned as follows:
a. Remove the r-f amplifier tube as described in paragraph 3.2.2.1.

## CAUTION

Special care must be used in removing or -installing the power amplifier tube to prevent damage to the tube.
b. Direct a low-pressure ( 50 psi ) air stream through the fins in the direction opposite to the normal airflow until all dust is removed.
c. Replace the r-f amplifier tube as described in paragraph 3.2.2.1.

### 3.2.2.1 PA TUBE REMOVAL.

## WARNING

Voltages present within the plate cavity are dangerous to life. Shut down the D830-1 doing any work inside the cavity. Short the plate to ground immediately on opening the plate cavity door. Do not depend entirely on the door interlock.

The power tube may be removed as follows:
a. Open the power amplifier cavity and loosen the clamp holding the coaxial transmission line and the power amplifier tube anode. Also loosen the plate slider clamp.
b. Grasp the center coaxial transmission line and lift until the center coaxial transmission line stops.
c. Turn the anode of V201 approximately $1 / 6$ turn (counterclockwise) until the tube clears the tube socket.
d. Remove tube V201 from the socket.
e. Replacement is the reverse of the removal procedure.

## NOTE

It may be necessary to move the OUTPUT COUPLING capacitor to the right so the center coaxial transmission line will clear the capacitor when the center transmission line is moved upwards.

### 3.2.3 INSPECTION.

Once each week check and clean the three interlock switches and the two shorting switches at the rear of the D830-1 cabinet to be sure they are in good working order.

Once each month check all connections in the D830-1. Tighten any nuts, bolts, or screws that may be loose. Check cable connections to see that they are clean and mechanically secure. Check moving parts such as tuning controls for excessive wear. Check the plate cavity for corrosion around the cavity contact strip. Check and clean (by lightly brushing) the screen grid voltage protector blocks located within the plate cavity.

### 3.2.4 LUBRICATION.

The PA blower is to lubricated once every six months with two drops of SAE no. 20 oil in each bearing. The cabinet fan has bearings that are lubricated for the life of the equipment. No other lubrication of the D830-1 is required.

### 3.2.5 TUBE MAINTENANCE.

The power amplifier, V201, should be inspected once each week to ensure that an accumulation of dust does not build up on the radiator fins. If dust is present, clean as described in paragraph 3.2.2. When tuning the D830-1, care should be taken not to exceed the maximum plate current shown in table 3-1.

### 3.3 TROUBLE SHOOTING.

The most common cause of trouble will probably be traced to tube failure. If the power amplifier tube is suspected of failure, replace it with a tube of known quality, retune, and note any change in performance. A small loss in emission of V201 can be compensated for by a change in the setting of the POWER OUTPUT ADJUST potentiometer. If no screen grid voltage is present, the trouble may lie in the screen grid protector blocks. These should then be cleaned or replaced.

Four meters are located on the D830-1 front panel to assist in locating any trouble which may occur. Table 3-1 contains typical meter indications. These average indications are obtained from several production power amplifiers, and the indications of some D830-1 may vary slightly outside the given limits without affecting the power amplifier performance. A list of panel meter indications for each

TABLE 3-1. TYPICAL METER INDICATIONS

| METER | METER SWITCH POSITION | INDICATION |
| :--- | :--- | :--- |
| MULTIMETER | SCREEN FS 400 VDC | 210 to 260 |
| MULTIMETER | SCREEN FS 40 MA | 25 ma |
| MULTIMETER | GRID FS 4 MA | 0.5 ma |
| PA PLATE VOLTAGE |  | 2650 to 2750 |
| PA PLATE CURRENT | Forward | Approx 600 ma |
| RF WATTMETER | Reflected | 1000 watts |
| RF WATTMETER |  | Less than 100 watts |

individual power amplifier should be taken when the D830-1 is operating properly in its particular installation. Any abnormal deviation from these values will then be apparent during a check of meter indications.

### 3.4 CABLE CHART.

Table 3-2 contains from-to information for cables installed in D830-1 1000-Watt FM Power Amplifier. The table is useful in locating point to point wiring
within the D830-1 cabinet. The FROM column is listed in alphabetical and numerical order. To find a particular wire, establish the point on the D830-1 from which wire tracing is to be initiated. Find this point in the FROM column of table 3-2 and the TO column will give the location of the other end of that particular wire. The WIRE CODE column gives the type and color of wire used in each case. Refer to the inside back cover of this manual for the wire code explanation. When the wire code CBSJ is encountered, the letters $S J$ mean shield with jacket.

TABLE 3-2. CABLE FROM-TO INFORMATION

| FROM | TO | WIRE CODE | FROM | TO | WIRE CODE |
| :---: | :---: | :---: | :---: | :---: | :---: |
| C205 | T302-7 | TC91 | [301 | R3C5-1 | KE O |
| C 206 | T302-5 | TCOl | E 301 | S309 | KEO |
| C 211 | E $21 t$ | RC4 | E 301 | TB308-1 | KEO |
| C 217 | ¢3C3-2 | LE9 | E 304 | R 209-1 | RC96 |
| C218 | 5302-11 | RC95 | [. 304 | S3C7-4 | RC 96 |
| C 219 | TE 3C4-3 | RC95 | E 306 | E 313 | SHIELD |
| C 301-1 | L201-2 | KEO | E 306 | M304-1 | SHIELD |
| C 302-2 | C3C1-2 | RC90 | + 307 | K $301-\epsilon$ | RC 91 |
| C 301-2 | C302-2 | RC90 | [307 | те3C7-1 | RC91 |
| C 301-2 | $\mathrm{F} 3 \mathrm{C} 4-3$ | RC90 | E 307 | TE3C9-1 | RC92 |
| C 301-2 | TH31C-3 | RC30 | [ 307 | XF3C4-1 | RC91 |
| C 307-1 | T3C2-4 | RC925 | E 308 | K202-3 | RC913 |
| C 303-1 | Te316-1 | KEO | F. 308 | K 3C5-5 | RC913 |
| C 307-2 | T30<-5 | RC926 | E 308 | S113-1 | RC913 |
| C4301-1 | Tr 3 Cl -1 | VG90 | E 308 | тe3C4-7 | RC913 |
| Ce3Cl-1 | $\mathrm{xF}^{\text {3 }}$ C1-1 | VG 90 | [ 310 | E 309 | VGg |
| CB301-2 |  | RE90 | [309 | E31C | VGg |
| C8301-3 | Te3Cl-2 | vGO | E 210 | E311 | RC9 |
| Ce301-3 | XF3C2-1 | VGO | E 310 | E 312 | RC9 |
| CB3Cl-4 | K3C4-3 | KE 95 | E310 | тв301-3 | VG9 |
| CR383-2 | XF3C4-2 | RC95 | E31C | 13C1-8 | RC9 |
| CC301-3 | E113 | SHIELD | E 310 | TE3C3-2 | RC9 |
| DC301-3 | S208E-4 | CBSJ905 | E 310 | Te3c8-7 | RC9 |
| CC3Cl-4 | [313 | SHIELD | E311 | E31C | RC9 |
| OC301-4 | 53c8a-11 | CBSJ903 | [312 | F.31C | RC9 |
| E 301 | L 3CE-2 | KEO | E 312 | Te313-1 | RC9 |
| E 301 | M $202-1$ | KEO | E 313 | [C3C1-3 | SHIELD |

TABLE 3－2．CABLE FROM－TO INFORMATION（Cont）

| FROM | TO | WIRE CODE | FROM | TO | WIRE CODE |
| :---: | :---: | :---: | :---: | :---: | :---: |
| E 313 | DC 3C1－4 | SHIELD | M 302－2 | TE303－8 | RC902 |
| E313 | E 3 C c | SHIELD | M 302－2 | TB308－6 | RC912 |
| E 313 | TE323－3 | SHIELD | M303－1 | E301 | KEO |
| E 313 | TB323－3 | SHIELD | M303－2 | C217 | LES |
| E 314 | S3C7－11 | RC93 | M304－1 | E 306 | SHIELD |
| E316 | Cく11 | RC4 | M 304－1 | TR308－7 | RC9 |
| E 316 | S307－10 | RCl | M304－2 | R321－1 | CHSJ906 |
| E 317 | TE303－1C | RC905 | K304－1 | K305－1 | RC915 |
| E 317 | TE3C7－2 | RC＇ | R 304－3 | C301－2 | RC90 |
| E 317 | Te3cs－2 | RC9 | R304－3 | R 322 － | RC90 |
| E 317 | S112－1 | RC9 | R 305－1 | E3C1 | KEO |
| ［ 322 | K 301－3 | RC9 | R 309－1 | E 204 | RC96 |
| E 322 | K3C2－6 | RC9 | R309－1 | R324－1 | RC6 |
| 」 305－1 | K3C1－12 | RCC | R309－2 | R $217-3$ | RC90 |
| J 305－1 | 1202－1 | RCC | R315－1 | R317－1 | RC903 |
| J305－2 | K201－9 | RC90 | R317－1 | R $315-1$ | RCSO3 |
| J305－2 | T302－2 | RC 90 | R 217－3 | K305－2 | RCSO |
| K 301－1 | K3C1－4 | RC916 | R317－3 | R309－2 | RC90 |
| K301－1 | S $302-3$ | RC96 | R321－1 | N304－2 | CBSJ906 |
| K 301－1 | S112－2 | RC96 | R321－3 | S302e－2 | CBSJ902 |
| K301－2 | K 302－2 | RC902 | R223－1 | R 204－3 | RC90 |
| K301－2 | S111－4 | RC902 | R323－2 | K 305－2 | RCS |
| K 301－2 | S314－1 | RCSI | R 324－1 | R305－1 | RC6 |
| K 301－3 | E 222 | RC9 | R 330 | S307－5 | RCY2 |
| K 301－4 | K3C1－1 | RC916 | S111－3 | TE3C2－8 | RC906 |
| ＜301－4 | K3C2－7 | RC916 | S111－4 | K201－2 | RC902 |
| K 301－6 | E3C7 | RC91 | 5111－4 | xCS 3C1－1 | RCS02 |
| K 301－7 | XF303－2 | RC902 | S112－1 | E317 | KC9 |
| K301－9 | J3C5－2 | RCGO | S112－2 | K 301－1 | RC96 |
| K301－10 | T3C1－6 | RC93 | S112－2 | TB308－9 | RC96 |
| K 301－12 | J3C5－1 | RCO | S113－1 | E308 | RC913 |
| K 301－13 | 1 301－1 | RC92 | S113－2 | K 302－1 | RC916 |
| K302－1 | K 302－4 | RC 916 | 5114－3 | S $313-2$ | RC915 |
| K302－1 | K3C2－9 | RC 91 | S114－4 | K 20503 | RC916 |
| K 302－1 | S112－2 | RC916 | S302－1 | TB3C3－1 | RCSl |
| K302－2 | K 301－2 | RC902 | S 302－3 | TE3C4－6 | RC96 |
| ト302－2 | 113C4－8 | RC 902 | S 302－3 | K201－1 | RC96 |
| K 302－3 | ㄷ3C 8 | RC913 | S 302－4 | TE303－5 | RC912 |
| k302－4 | K3C2－1 | RCgl6 | S302－4 | TE3C4－11 | RC912 |
| k 302－6． | E 322 | RC9 | S 302－5 | S 211 －1 | RC90 |
| K302－7 | K 3C1－4 | RC 316 | S 302－5 | TE303－4 | RC90 |
| K 302－9 | K3C2－1 | RC 91 | S 302－7 | TB 303－3 | RC92 |
| K 302－9 | T23C4－2 | RC923 | 5302－9 | Te3C4－1 | RC915 |
| K302－10 | K 304－1 | RC 906 | S 302－10 | TE3C2－8 | RC93 |
| K．302－10 | TH3CE－11 | RC916 | S 302－11 | C213 | RC95 |
| K 304－1 | K3C2－10 | RC906 | S 302－11 | TR302－9 | RC 95 |
| K．304－？ | TE3C4－5 | RC925 | 5307－7 | v？ 1 1－1 | RC 903 |
| K 304－2 | T¢313－3 | RC725 | S 307－4 | E304 | RC96 |
| K304－2 | x1；S 1C－2－1 | RC 923 | S307－5 | R 330 | RC92 |
| K304－3 | CH3C1－4 | RE 95 | S 307－8 | m301－2 | RC902 |
| 1．304－4 | 1 3 C3－1 | RE93 | 5307－10 | E316 | KCl |
| r3）4－5 | Cf？Cl－2 | RE90 | 5307－11 | E314 | RC93 |
| R．304－6 | T3Cs－5 | REGI | S308A－6 | TR233－5 | RAS95 |
| K305－1 | ？3C4－1 | RC915 | S 30EA－11 | IJC3C1－4 | CHSJ903 |
| $k 205-7$ | ＋217－3 | RC90 | S 30 BB －2 | R321－3 | CBSJ902 |
| K 30 5－2 | k323－2 | RC9 | S $3 \cap 8 B-4$ | CC3C1－3 | CBSJ905 |
| K305－3 | S114－4 | RC916 | S $309 \mathrm{BB}-5$ | T6333－4 | RAS90 |
| K305－5 | E $30: 3$ | RCY13 | S 309 | E301 | K E O |
| L 301－1 | IC315－1 | KEO | S3C9 | S310 | KF 0 |
| L 301－2 | C201－1 | KEO | 5315 | S 300 | KEC |
| L301－2 | L3C ${ }^{\text {c－1 }}$ | KR．C | S $311-1$ | S302－5 | RC＇O |
| L302－1 | L．301－2 | KEO | S311－2 | S312－1 | RC912 |
| L 302－2 | E 301 | K EO | 5312－1 | S311－2 | KC 912 |
| 4301－1 | S307－2 | RC 903 | S312－2 | 5313－1 | RCOL3 |
| Y301－2 | 53C7－8 | HC 302 | 5313－1 | S312－2 | RCH13 |
| －307－1 | T＋3C，8－5 | 4C902 | 5313－2 | 5114－3 | RCS15 |

TABLE 3-2. CABLE FROM-TO INFORMATION (Cont)


## SECTION IV

## PARTS LIST

This section contains a list of all replaceable electrical, electronic, and critical mechanical parts for the D830-1 1000-Watt FM Power Amplifier 522-2948-000. The manufacturers' codes appearing in the MFR CODE column of the parts list are listed in numerical order at the end of the parts list. The code list pro-
vides manufacturers' names and addresses as shown in the Federal Supply Code for Manufacturers, Handbook H4-1. Manufacturers not listed in Handbook H4-1 are assigned a five-letter code and will appear first in the code list.

ILLUSTRATIONS . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . 17
PARTS LIST . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . 27


Figure 4-1. D830-1 FM Power Amplifier (Sheet 1 of 10)


Figure 4-1. D830-1 FM Power Amplifier (Sheet 2 of 10)


Figure 4-1. D830-1 FM Power Amplifier (Sheet 3 of 10)


Figure 4-1. D830-1 FM Power Amplifier (Sheet 4 of 10)


Figure 4-1. D830-1 FM Power Amplifier (Sheet 5 of 10)


Figure 4-1. D830-1 FM Power Amplifier (Sheet 6 of 10)


Figure 4-1. D830-1 FM Power Amplifier (Sheet 7 of 10)


DEIALID

Figure 4-1. D830-1 FM Power Amplifier (Sheet 8 of 10)


DEALILE


Figure 4-1. D830-1 FM Power Amplifier (Sheet 9 of 10)


DEACILF


Figure 4-1. D830-1 FM Power Amplifier (Sheet 10 of 10 )


| SYMBOL | DESCRIPTION | MANUFACTURER＇S PART NUMBER | MFR CODE | COLLINS <br> PART NUMBER |
| :---: | :---: | :---: | :---: | :---: |
| C315 | SAME AS C．314 |  |  |  |
| C716 | CAME AS C3：4 |  |  |  |
| C．317 | NOT USED |  |  |  |
| C318 | NOT USED |  |  |  |
| C319 | NOT USED |  |  |  |
| C320 | CAPACITOR，FXD，CERAMIC 10.000 UUF．PLUS 20\％MINUS 20\％． 1000 VDC | DAO49－182CE | 71590 | 913－3183－000 |
| C321 |  |  |  |  |
| THROLKH | SAME AS C320 |  |  |  |
| C399 |  |  |  |  |
| C8301 | CIRCUIT BREAKER <br> 50－AMP CURRENT RATING |  |  | 260－0243－000 |
| CR301 | SEMICONDUCTOR DEVICE，DIICDE | 1N1576 | 07688 | 353－1736－000 |
| CR302 |  |  |  |  |
| THROUGH | SAME AS CR301 |  |  |  |
| CR380 |  |  |  |  |
| CR 381 | SEMICONDUCTOR DEVICE，SET | SOM140285 | 04713 | 353－6015－000 |
| CR3日2 | SAME AS CR382 |  |  |  |
| CR383 | SEMICONDUCTOR DEVICE，DIODE | IN1492 | 07688 | 353－1661－000 |
| CR384 | SAME AS CR383 |  |  |  |
| CR385 | SAME AS CR30： |  |  |  |
| CR386 | SEMICONDUCTOR DEVICE，DIODE | 1 N270 | 81349 | 353－2016－000 |
| CR387 | SEMICONDUCTOR DEVICE，DIODE | 1N30168 | 07688 | 353－3121－000 |
| CR388 | SEMICONDUCTOR DEVICE，DIODE | 1 N963A | 07688 | 353－3220－000 |
| DS301 | LAMP，BULB <br> 0．Cこ7 AMP， 125 VOLTS | 356－5 | 24446 | 262－3310－000 |
| DS302 | SAME AS DS301 |  |  |  |
| F．3n1 | ARRESTOR，ELECTRICAL |  |  | 549－2453－002 |
| E302 | LENS，INDICATOR GREEN | 75A101GRN | 72765 | 262－0258－000 |
| E303 | LENS，INDICATOR RED | 75A101RED | 72765 | 262－0259－000 |
| E304 | PUSHBUTTON |  |  | 548－3590－003 |
| E305 | CONTACT，ELECTRICAL |  |  | 549－2317－002 |
| E306 | CONTACT．SHORTING |  |  | 542－1773－002 |
| E 307 | SLIDER，CONTACT |  |  | 549－2413－002 |
| E308 | SAME AS E307 |  |  |  |
| F301 | FUSE 4 CARTRIDGE 5 AMPS． 250 VDC | F03A2SOV5AS | A 1340 | 264－0361－000 |
| F30？ | SAME $A S$ F301 |  |  |  |
| F303 | FUSE，CARTRIDGE 4 AMPS， 125 VDC | MDX4 | 71400 | 264－0217－000 |
| F304 | FUSE，CARTRIDGE <br> 1／日 AMP， 250 VOLTS | F02B250V1－8AS | 81349 | 264－4230－000 |
| $J 201$ | CONNECTOR，ELECTRICAL 1 CONTACT | UG 10940 | 80058 | 357－9183－000 |
| $J 202$ | CONNECTOR，ELECTRICAL 1 CONTACT | UG58AU | 80058 | 357－900コ－000 |
| J203 | SAME AS J201 |  |  |  |
| J301 | CONNECTOR，ELECTRICAL 1 CONTACT | UG58．4U | 80058 | 357－9003－000 |
| J．3n？ | CAME $\triangle S$ JTnt |  |  |  |
| $J 303$ $J 304$ | CONNECTOR CONNECTOR，ELECTRICAL | $2261$ | 84147 <br> 94375 | $013-1215-000$ $357-9248-000$ |
| J304 | CONNECTOR，ELECTRICAL <br> 1 CONTACT | 100日3000C75 | $94375$ | 357－9248－000 |
| 1305 | CONNECTOR，ELECTRICAL 3 CONTACTS |  |  | 368－0014－000 |
| K301 | RELAY，ARMATURE 4C CONTACT ARRANGEMENT | 83－3544 | 04221 | 970－1933－000 |
| K3n2 | SAME AS K301 |  |  |  |
| K303 | SEMICONDUCTOR DEVICE． RECTIFIER |  |  | 549－2463－003 |
| K3n4 | RELAY，ARMATURE 2 CONTACTS | 702LP0092 | 01121 | －0゙5－ 0674 － 10 |
| K305 | RELAY，ARMATURE <br> IC CONTACT ARRANGEMENT | 95062 | 78277 | 408－1114－000 |
| K306 | RELAY：ARMATURE 2C CONTACT ARRANGEMENT | 45－2446 | 04221 | 972－1347－000 |
| $K 307$ | RELAY，THERMAL <br> IA CONTACT ARRANGEMENT | GT4574 | 93929 | 402－0．388－000 |
| L201 | NOT USED |  |  |  |
| L202 | COIL．RADIO FREQUENCY | LT7K194 | 81349 | 240－0178－000 |


| SYMBOL | DESCRIPTION | MANUFACTURER'S PART NUMBER | MFR CODE | COLLINS PART NUMBER |
| :---: | :---: | :---: | :---: | :---: |
|  | 4.7 UH. 0.60 OHMS. 950 MA DC |  |  |  |
| L203 | COIL, RADIO FREQUENCY |  |  | 549-2373-002 |
| L301 | ```CHOKE 10 H. 60 OHMS: 0.850 AMP``` | 292-5822GI | 49956 | 668-0022-000 |
| L. 302 | SAME AS L301 |  |  |  |
| L. 303 | COIL |  |  | 549-2042-002 |
| L304 | SAME AS L303 |  |  |  |
| ᄂ305 | COIL |  |  | 549-2043-002 |
| L306 | SAME AS L305 |  |  |  |
| M301 | AMMETER <br> 100 OHMS, PLUS 5\% MINUS 10\% METER RESISTANCE | 56-0443-0000 | 80145 | 458-0649-000 |
| M302 | VOLTMETER <br> 0 TO 1 MA METER RANGE, 50 OHMS | 56-0273-0000 | 80145 | 458-0610-000 |
| M303 | AMMETER <br> 0 TO 800 MA METER RANGE, 0.5 OHMS | 56-9623-0000 | 80145 | 458-0611-000 |
| M304 | WATTMETER 900 OHMS, 2\% ACCURACY | 56-4752-0000 | 80145 | 458-0652-000 |
| MP301 | KNOB <br> BLACK PHENOLIC |  |  | 546-1 293-002 |
| MP3n? |  |  |  |  |
| THPOUGH MP305 | SAMF AS MP301 |  |  |  |
| MP306 | KNOB <br> BLACK PHENOLIC |  |  | 547-8792-003 |
| MP307 | CONTACT. ASSEMBLY <br> INTERLOCK SWITCH | N4050 | 85107 | 260-4050-000 |
| MP308 | SAME AS MP307 |  |  |  |
| MP309 | LATCH, MAGNETIC ALUMINUM, STRIKER PLATE STEEL. FERRITE MAGNET | ADPL 1 OODCST 1 | 84792 | 015-0899-000 |
| MP310 | SAME AS MP309 |  |  |  |
| MP311 | ROD ASSEMELY, SHORTING日RASS, WITH PLAGTIC HANDLF. 24.938 INCHES LONG |  |  | 549-2186-003 |
| MP312 | SLIDER, TUNING CAVITY Aluminum |  |  | 549-2424-004 |
| MP313 | SHAFT. STRAIGHT, SHORT GLASS MELAMINE ROD |  |  | 549-2436-002 |
| MP314 | SHAFT. STRAIGHT, LONG GLASS MELAMINE ROD |  |  | 549-2437-002 |
| MP315 | SLIDER, CONTACT RRASS |  |  | 549-2413-002 |
| MP316 | SAME AS MP315 |  |  |  |
| MP317 | GFAR, SPUR <br> ALUMINUM, 48 TE゙FTH |  |  | 542-7422-002 |
| MP318 | HINDOW, METER ORSERVATION GLASS, 3/16 INCHES THICK |  |  | 548-3567-002 |
| MP319 | SPRING, SHORTING COPPER, 0.032 INCHES THICK |  |  | 549-2374-002 |
| MP320 | COUPLING <br> BRASS | 2105 | 74887 | 015-0257-000 |
| MP32 1 | BEARING, SLEEVE | F346MILL6085A | 70417 | 309-0086-000 |
| Mp322 | NOT USED |  |  |  |
| MP323 | CONDUCTOR, ROD. LONG |  |  | 549-2372-002 |
| MP324 | CONDUCTOR, ROD, SHORT |  |  | 549-2371-002 |
| MP325 | CLAMP. HALF, ELECTRICAL, LOWER |  |  | 549-2367-002 |
| MP326 | CLAMP, HALF: ELECTRICAL, UPPER |  |  | 549-2.366-002 |
| MP327 | PLATE, CAPACITOR |  |  | 549-2379-003 |
| P301 | CONNECTOR, ELECTRICAL 1 CONTACT | MS35168-88E | 96906 | 357-9292-000 |
| P302 | SAME AS P301 |  |  |  |
| P303 | CONNECTOR, ELECTRICAL 1 CONTACT | UG 11854 L | 81349 | 357-9326-000 |
| P304 | SAME AS P303 |  |  |  |


| SYMBOL | DESCRTPTION | MANUFACTURER'S PART NUMBER | MFR CODE | COLLINS PART NUMBER |
| :---: | :---: | :---: | :---: | :---: |
| P305 | SAME AS P303 |  |  |  |
| P306 | NOT USED |  |  |  |
| P307 | SAME AS P3O3 |  |  |  |
| P308 | NOT USED |  |  |  |
| P309 | CONNECTOR, ELECTRICAL |  |  | 368-0013-000 |
| P310 | SAME AS P303 |  |  |  |
| R201 | RESISTOR, FXD, COMPOSITION 3300 OHMS, $10 \%$ TOL. 2 WATTS | RC42GF332K | 81349 | 745-5673-000 |
| R202 | RESISTOR: FXD, COMPOSITION 100 OHMS, $10 \%$ TOL, 2 WATTS | RG42GF101K | 81349 | 745-5610-000 |
| 8203 | RESISTOR, FXD, WIRE WOUND 470 OHMS, 5\% TOL, 6.5 watts | RW67V471 | 81349 | 747-5527-000 |
| R204 | NOT USED |  |  |  |
| R205 | SAME AS R201. |  |  |  |
| R296 | GAMF AS R2O1 |  |  |  |
| R207 | SAME AS R201 |  |  |  |
| R200 | RESISTOR, FXD, COMPOSITION 47 OHMS, $20 \%$ TOL, 15 WATTS | 7705P6 | 10646 | 712-0014-000 |
| R301 | RESISTOR, FXD, COMPOSITION <br> 1 K OHMS. $10 \%$ TOL. 2 WATTS | RC42GF 102 K | 81349 | 745-5652-000 |
| R302 | SAME AS R391 |  |  |  |
| R303 | NOT USED |  |  |  |
| R304 | RESISTOR, VAR, WIRE WOUND 50 OHMS. $10 \%$ TOL. 2 WATTS |  | 12697 | 377-0619-000 |
| R305 | RESISTOR, FXD, WIRE WOUND 2OK OHMS, 5\% TOL. 210 WATTS | RW4 7U203 | 81349 | 746-6723-000 |
| R306 | RESISTOR, FXD, COMPOSITION 10 OHMS, $10 \%$ TOL. 2 WATTS | RC42GF 100 K | 81349 | 745-5568-000 |
| R307 | RESISTOR. FXD, WIRE WOUND 2.56 OHMS. $1 \%$ TOL, 2.5 WATTS | RS 10-10001H | 91637 | 746-9448-000 |
| R308 | NOT USED |  |  |  |
| R309 | RESISTOR. FXO, WIRE WOUND 15K OHMS, 5\% TOL. 20 WATTS | 0217 | 49655 | 710-4782-000 |
| R310 | RES:STOR. FXD, FILM 1000K OHMS, $1 \%$ TOL. 2 WATTS | RNEOB 1004 F | 81349 | 705-4254-000 |
| R311 | SAME AS R310 |  |  |  |
| R312 | SAME AS R310 |  |  |  |
| R313 | RESISTOR, FXD, COMPOSITION 10K OHMS, $10 \%$ TOL 1 WATT | RC32GF 10.3 K | 81349 | 745-3394-000 |
| R314 | ```RESISTOR. FXD, FILM 7500 OHMS. 5% TOL. 1 WATT``` | RL.32S752J | 81349 | 745-3994-000 |
| R315 | PESISTOR, FXD, WIRE WOUND 27.0 OHMS, 5\% TOL, 55 UATTS | Rw35V2 70 | 91349 | 747-2815-000 |
| R316 | RESISTOR, FXD, COMPOSITION 33 OHMS, $10 \%$ TOL. 1 WATT | RC32GF330K | 81349 | 795-3289-000 |
| R317 | RHEOSTAT, WIRE WOUND 75 OHMS. $10 \%$ TOL. 100 watTS | ```R100-75REARSHAFT EXT``` | 94310 | 735-4000-000 |
| R318 | SAME AS R315 |  |  |  |
| R319 | RESISTOR, FXD, WIRE WOUND 10 OHMS. 5\% TOL, 11 WATTS | RW29U100 | 81349 | 746-6040-000 |
| R320 | RESISTOR, FXD, COMPOSITION 1200 OHMS. $10 \%$ TOL, 2 WATTS | RC42GF122K | 81349 | 745-5656-000 |
| R32 1 | RESISTOR, VAR, WIRE WOUND IOK OHMS, $20 \%$ TOL, 2 WATTS | RV4LAYSA103B | B 1349 | 380-2757-000 |
| R322 | RESISTOR. FXD. FILM <br> 4OPK OHMS. $1 \%$ TOL, I WATT | RN75B4023F | A1349 | 705-328 7-000 |
| R323 | RESISTOR, FXD, WIRE WOUND 3.9 OHMS, 5\% TOL, 11 WATTS | Rw29V2R7 | 81349 | 746-6115-000 |
| R324 | SAME AS R305 |  |  |  |
| R325 | RESISTOR, FXD, COMPOSITION 100 OHMS. $10 \%$ TOL. 2 WATTS | RC42GF 101 K | 81349 | 745-5610-000 |
| R326 | RESISTOR, ADJUSTAELE |  |  | 716-0013-000 |


|  | SYMBOL | DESCRIPTION | PART NUMBER | CODE | PART NUMBER |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | 1OK OHMS, 10\% TOL. 10 WATTS |  |  |  |
|  | R327 | SAMF AS R320 |  |  |  |
|  | R328 | SAME AS R320 |  |  |  |
|  | R329 | RESISTOR, FXD. COMPOSITION 27K OHMS. 10X TOL, 2 WATTS | RC42GF273K | 81349 | 745-5712-000 |
|  | R330 | RESISTOR. FXD. WIRE WOUND 310 OHMS. 5\% TOL. 11 WATTS | Rw29V222 | 81349 | 746-6087-000 |
|  | R331 | SAME AS R320 |  |  |  |
|  | R332 | SAME AS R320 |  |  |  |
|  | R333 | SAME AS R310 |  |  |  |
|  | R334 | SAME AS R306 |  |  |  |
|  | R335 | SAME AS R306 WIRE WOUND |  |  |  |
| * | R336 | RESISTOR. FXD, WIRE WOUND 12 OHMS, 5\% TOL, 14 WATTS | RW3IV120 | 81349 | 747-0726-000 |
|  | R337 | RESISTOR, FXD, COMPOSITION 680 OHMS. $10 \%$ TOL, 2 WATTS | RCa 2 GF681K | 81349 | 745-5645-000 |
| * | R338 | RESISTOR, FXD. COMPOSITION 8200 OHMS. $10 \%$ TOL. 1 WATT | RC32GF822K | 81349 | 745-3391-000 |
|  | R339 | NOT USED |  |  |  |
|  | R340 | RESISTOR, FXD, COMPOSITION 0.19 MEGOHMS. $10 \%$ TOL. 1 WATT | RC32GF 184 K | 81349 | 745-3447-000 |
|  | R 341 |  |  |  |  |
|  | THROUGH <br> R410 | SAME AS R.740 |  |  |  |
|  | R4>0 | NOT USED |  |  |  |
|  | R421 | RESISTOR. FXD. FILM <br> 1800 OHMS. $5 \%$ TOL. 1 WATT | PL32S182J | 81349 | 745-3956-000 |
|  | R4.22 | RESISTOR. FXD. FILM <br> 4700 OHMS. 5\% TOL. 1 WATT | RL32S472J | 81349 | 745-3981-000 |
|  | R423 | RESISTOR, FXD, WIRE WOUND 100 OHMS, 5\% TOL. 6.5 WATTS | RW67V101 | 81349 | 747-5440-000 |
|  | R424 | RESISTOR. FXD. WIRE WOUND 220 OHMS. $10 \%$ TOL. 6.5 WATTS | RW67V221 | 81349 | 747-5447-000 |
|  |  |  |  |  |  |
|  | THROUGH $\text { S } 110$ | NOT USED |  |  |  |
|  | $5111$ | SWITCH. PUSH OPST CONTACT ARRANGEMENT | B2BL | 04000 | 260-2020-000 |
|  | $5112$ | SAME AS S:II |  |  |  |
|  | S113 | SAME AS SIII |  |  |  |
|  | S 114 | CAME AS S111 |  |  |  |
|  | 5201 | ARRESTOR. ELECTRICAL |  |  | $549-245.3-002$ |
|  | S202 | SWITCH, INTFRLOCK SPDT CONTACT ARRANGEMENT | $3 \mathrm{AC5}$ | 91929 | $266-8013-000$ |
| '. | $\begin{array}{r} 5301 \\ 5302 \end{array}$ | NOT USED <br> SWITCH. ROTARY |  |  |  |
|  | S302 | SWITCH. ROTARY <br> ROTARY WAFER SWITCH |  |  | 259-1564-000 |
|  | S303 |  |  |  |  |
| 4 | $\begin{aligned} & \text { THROUGH } \\ & 5306 \end{aligned}$ | NOT USED |  |  |  |
|  | 5307 | SWITCH, ROTARY 8 CONTACTS |  |  | 259-1565-000 |
|  | $5308$ | SWITCH, ROTARY ROTARY WAFER SWITCH | 228556-F1E | 76854 | 259-1806-000 |
|  | $5.309$ | NOT USED |  |  |  |
|  | 5310 | NOT USED |  |  |  |
|  | S3114 | CONTACT ASSEMBLY, ELECTRICAL 5 CONTACTS | N4050 | 85107 | 260-4050-000 |
|  | S311R | CONTACT ASSEMBLY, ELECTRICAL 2 CONTACTS | N4040 | 85107 | 260-4040-000 |
|  | S312A | SAME AS S311A |  |  |  |
|  | S312n | SAME AS S311日 |  |  |  |
|  | S313A | SAME AS S311A |  |  |  |
|  | 53138 | SAME AS S311日 |  |  |  |
|  | S314 | SPDT CONTACT ARDANGEMENT | 4000 | 82877 | 266-930 7-000 |






# 830E-1A <br> 5 KW FM Broadcast Transmitter 

system instructions

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# UNIT INSTRUCTIONS 

TD No.
Title

536 A830-2 10 W Wide-Band FM Broadcast Exciter
539

538 B830-1 250-Watt FM Power Amplifier 5-Kw FM Power Amplifier


Figure 1-1. 830E-1A 5-Kw FM Broadcast Transmitter, Over-all View

## SECTION I <br> GENERAL DESCRIPTION

### 1.1 PURPOSE OF INSTRUCTION BOOK.

This instruction book is a guide for installing, adjusting, operating, and maintaining 830E-1A 5-Kw FM Broadcast Transmitter.

### 1.2 PURPOSE OF EQUIPMENT.

The 830E-1A 5-Kw FM Broadcast Transmitter is used for continuous monophonic or optional stereophonic FM broadcast service on a single frequency in the range from 88 to 108 megacycles with an output power of 5000 watts.

### 1.3 DESCRIPTION OF EQUIPMENT.

### 1.3.1 PHYSICAL DESCRIPTION.

The 830E-1A 5-Kw FM Broadcast Transmitter, shown in figure $1-1$, is contained in two interconnected cabinets that, together, are 76 inches wide, 76 inches high, 27 inches deep, and weigh approximately 1800 pounds. All transmitter operating controls are located behind the doors on the front of the cabinets. The filament and plate on-off controls and eight monitoring meters are located at the top front of the cabinets. The meters may be observed easily while operating the tuning controls. A monitoring meter is also provided with the wide-band exciter. The transmitter uses 15 tubes and 20 transistors, most of which are accessible from the front of the transmitter. The bottom front of the transmitter cabinets are removable to allow access to components on the bottom of the inside panels.

Large doors at the upper rear of the cabinets allow access to the upper part of the transmitter for servicing and maintenance. The lower rear half of the transmitter cabinets are covered by removable panels that contain ventilating fans and permanent-type air filters. Operating personnel are protected by both electrical and mechanical interlocks on the rear doors and panels. These interlocks ground the transmitter high-voltage circuits when the doors are opened or the panels are removed. The power amplifier platetuning cavity is located in an interlocked compartment at the front of the transmitter.

Inside the transmitter, heavy iron-core components are at the bottom of the cabinets. The exciter portion of the transmitter and the 250 -watt driver are contained in one cabinet. The other cabinet contains a 5000-watt power amplifier and harmonic filter.

Cooling air for the transmitter is drawn through permanent air filters at the rear of the cabinets by lowspeed, high-volume fans, and exhausted through shielded openings in the tops of the cabinets. Individual blowers supply cooling air directly to the driver and power amplifier tubes.

Room is provided in the 250 -watt amplifier cabinet for mounting a stereo generator and SCA equipment if multiplex operation is desired.

### 1.3.2 ELECTRICAL DESCRIPTION.

The 830E-1A 5-Kw FM Broadcast Transmitter is composed of three electrically-connected subunits: (1) a wide-band exciter that furnishes a 10 -watt FM output to drive (2) a 250 -watt amplifier that, in turn, drives (3) a 5000-watt power amplifier. Instruction books covering the exciter and power amplifiers used in the transmitter are listed in table 1-1. These three books are supplied following section $V$ of this system instruction book. The unit instruction books contain detailed descriptions of the three transmitter subunits.

TABLE 1-1
SUBASSEMBLY INSTRUCTION BOOKS

| PUBLICATION | INSTRUCTION BOOK <br> NUMBER |
| :--- | :---: |
| A830-2 10 W Wide-Band <br> FM Broadcast Exciter <br> B830-1 250-Watt FM Power <br> Amplifier <br> $5-K w ~ F M ~ P o w e r ~ A m p l i f i e r ~$ | TD-536 |

### 1.4 EQUIPMENT SUPPLIED.

Table 1-2 lists equipment that is supplied as part of 830E-1A 5-Kw FM Broadcast Transmitter.

TABLE 1-2
EQUIPMENT SUPPLIED

| EQUIPMENT | COLLINS <br> PART NUMBER |
| :--- | :---: |
| A830-2 10 W Wide-Band FM <br> Broadcast Exciter <br> B830-1 250-Watt FM Power <br> Amplifier <br> E830-1 5-Kw FM Power <br> Amplifier <br> 5-Kw Harmonic Filter | $522-2714-00$ |

### 1.5 ACCESSORY EQUIPMENT.

Table 1-3 lists accessory equipment that is available for use with 830E-1A 5-Kw FM Broadcast Transmitter. Information on $786 \mathrm{M}-1$ Stereo Generator will be found in Unit Instructions TD-537.

TABLE 1-3
ACCESSORY EQUIPMENT

| EQUIPMENT | COLLINS <br> PART NUMBER |
| :---: | :---: |
| $786 \mathrm{M}-1$ Stereo Generator | $522-2914-00$ |

### 1.6 EQUIPMENT SPECIFICATIONS.

### 1.6.1 MECHANICAL.



### 1.6.2 ELECTRICAL.

Power source. . . . 200 to 250 volts, 60 -cycle, 3 -phase.

Maximum power
requirements . . . . 12.25 kilowatts.
Power output . . . . 5000 watts nominal.
Output impedance . . 50 ohms, unbalanced. Maximum swr 2:1.

Frequency range . . 88 to 108 mc . Exact operating frequency determined by frequency of exciter.

Excitation source . . Crystal - controlled highstability oscillator using a plated, nontemperaturecontrolled crystal, controlling an LC modulation oscillator to provide automatic frequency control. Then the
modulation oscillator is heterodyned up to the operating frequency of the station by a second highstability crystal-controlled oscillator.

## Carrier-frequency

stability . . . . . . Within $\pm 2000 \mathrm{cps}$ of specified carrier frequency over ambient temperature range from $+20^{\circ} \mathrm{C}\left(68^{\circ} \mathrm{F}\right)$ to $+45^{\circ} \mathrm{C}$ ( $113^{\circ} \mathrm{F}$ ) and line-voltage variations of $\pm 5$ percent.

Harmonic and
spurious radiation
Any emission appearing on a frequency removed from the carrier by between 120 kc and 240 kc , inclusive, is at least 30 db below the level of the unmodulated carrier.

Any emission appearing on a frequency removed from the carrier by more than 240 kc and up to and including 600 kc is at least 40 db below the level of the unmodulated carrier.

Any emission appearing on a frequency removed from the carrier by more than 600 kc is at least 80 db below the level of the unmodulated carrier.

Modulation
characteristics . . . Direct-frequency modulation. Standard audio preemphasis is incorporated in modulator.

Audio input
impedance . . . . . 600 ohms, balanced.
Audio input level . . $+10 \mathrm{dbm} \pm 2 \mathrm{db}$.
Audio frequency
response. . . . . . Complies with standard FCC 75-microsecond preemphasis curve.
Audio frequency
distortion . . . . . 50 to $100 \mathrm{cps}, 1.5$ percent maximum.
100 to 7500 cps, 1.0 percent maximum.
7500 to $15,000 \mathrm{cps}, 1.5$ percent maximum.

FM noise level . . . Not less than 65 db below 100 percent modulation ( $\pm 75$ kc ).

AM noise level (rms) Not less than 55 db below equivalent 100 percent AM .

# SECTION II INSTALLATION 

### 2.1 UNPACKING AND INSPECTING.

Be careful when uncrating the transmitter and components to avoid damaging the equipment. Inspect the transmitter carefully for scratches, dents, or other physical damage. Check for loose screws and bolts. Inspect all controis, such as switches, for proper operation as far as can be determined without applying power to the transmitter. Examine cables and wiring, making sure that all connections are tight and clear of each other and the chassis. File any damage claims promptly with the transportation company. If such claims are to be filed, retain all packing material.

### 2.2 TRANSMITTER LOCATION.

Plan transmitter and wiring placement carefully before starting installation work. Refer to figure 2-1, the transmitter installation diagram. This diagram shows the location of all wiring openings in the transmitter cabinets. As will be noted in figure 2-1, several alternate wiring arrangements can be used. Select the combination that most nearly suits the station requirements.

Allow adequate clearance both in front and back of the transmitter. There should be a minimum clearance of $3-1 / 2$ feet behind the transmitter to provide sufficient room for service work.

If desired, an air duct may be placed over the exhaustair opening in the top of the 5 -kw amplifier cabinet to carry heat away from the transmitter.

### 2.3 INTERCABINET CONNECTIONS.

Place the two transmitter cabinets beside each other in their permanent location so that, when viewed from the front, the 250 -watt driver cabinet is on the left and the $5-\mathrm{kw}$ amplifier cabinet is on the right. Then connect the two cabinets with the appropriate cables. The intercabinet cables consist of (1) an r-f coaxial cable, and (2) a laced control and power cable. Both cables are furnished with the transmitter.

Connect the coaxial cable from the r-f output of the 250 -watt driver to the $r$-f input of the $5-\mathrm{kw}$ power amplifier. These connections are made on the tops of the cabinets. Run the control and power cable through the circular openings near the bottom rear of the cabinets on the sides where the cabinets adjoin. Table 2-1 shows the terminal locations of the laced control and power cable and the cable wire code. An explanation of the wire code is given inside the back cover of the instruction book. When the transmitter is received, one end of the laced control and power cable will be connected to the terminal boards of the $5-\mathrm{kw}$ power amplifier. The other end must be connected to the 250 -watt driver at the station site.

TABLE 2-1. CABINET INTERCONNECTIONS

| WIRE CODE | 5-KW POWER AMPLIFIER TERMINAL <br> BOARD CONNECTIONS | 250-WATT DRIVER TERMINAL <br> BOARD CONNECTIONS |
| :--- | :---: | :---: |
| RE2 | TB402-1 | TB301-1 |
| RE5 | TB402-2 |  |
| RE9 | TB402-3 | TB301-2 |
| RC90 | TB403-1 | TB301-3 |
| RC923 | TB403-2 | TB304-1 |
| RC91 | TB403-3 | TB403-5 |
| RC4 | TB403-6 | TB304-2 |
| RC93 | TB403-7 | TB304-3 |
| RC92 | TB403-8 | TB304-6 |
| RC0 | TB403-10 | TB403-11 |
| VE9 |  | TB304-7 |
| RC91 |  | TB304-8 |

### 2.4 REPLACEMENT OF COMPONENTS REMOVED FOR SHIPPING.

Several of the transmitter components have been removed from the cabinets and packed separately for safety during shipping. These include the plate transformer and fragile units such as power amplifier tubes, mercury vapor tubes, and crystals. These units should not be replaced in the cabinets until the transmitter is in its permanent location. Wires and cables that were discomected before shipping have been tagged to facilitate recomection. Refer to the photographs in section VI of TD-538 and TD-539 for assistance in replacing these components in the transmitter.

### 2.5 EXTERNAL CONNECTIONS.

Refer to figure 2-1 for assistance in making the following external connections.
a. Connect the audio input to the transmitter. Bring the audio signal through the bottom (or any one of the optional cable input locations) of the 250 -watt driver cabinet on a shielded twisted pair. Connect the two audio leads to terminals 1 and 2 of TB305 if monaural operation is specified. If optional stereophonic operation is employed, the left audio leads are connected to TB305-1 and 2 and the right audio leads to TB305-4 and 5. TB305 is located about half way up the cabinet on the left side as viewed from the rear of the cabinet. Connect the shield to terminal 3 of TB305.
b. Connect the FM monitor to the monitor output on the top of the 5 -kw amplifier cabinet. Refer to figure 2-1. Use type RG-58U coaxial cable to make this connection.
c. Connect the antenna transmission line to the $r-f$ output located on top of the $5-\mathrm{kw}$ amplifier cabinet. The r-f connection is for a standard E1A 3-1/8-inch flange.

## CAUTION

Before making this antenna connection, be sure that the transmission line and antemna present a nominal impedance of 50 ohms and an swr of not more than 2:1 at the transmitter operating frequency. If the transmitter output is improperly matched, the transmitter will not operate properly and may be damaged. THIS IS IMPORTANT.
d. Connect the power input cable to the transmitter. This power cable should be brought from an external fused cut-out box rated for 40 amperes. Use type $R$ or $T$ AWG \#6 wire to make these connections. Connect the three wires to terminals 1,2 , and 3 of TB401, located at the bottom left side of the $5-\mathrm{kw}$ amplifier cabinet. The power cable may be brought into the transmitter through holes in either the top or bottom of the $5-\mathrm{kw}$ amplifier cabinet.

The 3 -phase power input must be connected properly. To check for proper phasing of the 3 -phase power input, turn off the plate circuit breaker and supply 3 -phase power to the transmitter control circuits. Check the 5-kw PA blower for clockwise rotation. If the blower rotation is not clockwise, reverse any two of the incoming power leads. ALWAYS SHUT OFF THE TRANSMITTER 3-PHASE POWER AT THE EXTERNAL FUSED CUTOUT BOX BEFORE MAKING ANY ADJUSTMENTS TO THE TRANSMITTER. Recheck the PA blower for clockwise rotation.

### 2.6 INTERNAL CONNECTIONS.

The 830E-1A 5-Kw FM Broadcast Transmitter plate, screen, filament, and control circuit transformers are fitted with adjustable taps to compensate for line variations from 200 to 250 volts in 10 -volt steps. To adjust transformers T 301, T303, T401, T403, T404, and T 405 perform the following steps.
a. Measure the line voltage at the transmitter fused cut-out box for each of the three phases. The three voltage readings should be nearly equal.
b. In turn remove the wire from the tapped portion of each of the transformers and move to the transformer tap whose voltage rating most closely corresponds to the voltage measured in step a. Refer to figure 2-2 for transformer terminal numbers and the input voltage which should be applied to each terminal. Do not move the solder lug from transformer terminal 5 of T301 or T401 as this terminal supplies 230 volts to the cabinet fans, the grid bias supply, and the highvoltage power supply filaments.
c. Tighten all transformer terminal connections.

Check that a jumper wire is installed from TB404-7 to TB404-8 to ensure that 115 -volt power will be applied to the transmitter control circuitry.

If the optional stereo generator is installed any time after the initial $830 \mathrm{E}-1 \mathrm{~A}$ installation, the $18-\mathrm{db}$ audio pad will have to be removed from the audio circuitry of the exciter. Refer to Unit Instructions TD-536 for the location of the $18-\mathrm{db}$ audio pad.

### 2.7 REMOTE CONTROL.

Remote control of $830 \mathrm{E}-1 \mathrm{~A} 5-\mathrm{Kw}$ FM Broadcast Transmitter, can easily be accomplished by connection to terminal boards TB-302 and TB303 located in the 250 -watt power amplifier cabinet and by placing a jumper between TB404-4 and TB404-9. Terminal board TB404 is located within the 5 -kw power amplifier cabinet. Table 2-2 lists the terminal board connections and the remote functions of each pair of terminals. Remote "on' switches should be the normally




Figure 2-2. Transformer Details
open momentary type. Remote "off" switches should be the normally closed momentary type. For remote operation, the LOCAL-REMOTE switch within the 250 -watt power amplifier cabinet, should be in the REMOTE position. When in the REMOTE position, it is possible to control the transmitter from the transmitter panel switches or from the remote point.

TABLE 2-2
REMOTE CONTROL CONNECTIONS

| FUNCTION | TERMINALS |  |
| :--- | :---: | :---: |
|  | TB302 | TB303 |
| FILAMENT ON |  | 1 and 2 |
| FILAMENT OFF | 8 and 9 |  |
| PLATE OFF |  | 2 and 4 |
| PLATE ON |  | 2 and 3 |

For simplified operation the FILAMENT ON and PLATE OFF switches could be eliminated. The PLATE ON switch starts a sequence of operations which turns the filaments on and the plate voltage on after the filament time delay is completed. The FILAMENT OFF switch shuts down all transmitter functions.

Equipment is available that will completely control and monitor transmitter operation from a remote location through standard telephone pairs. When such remote control equipment is used, necessary instaliation and connection information will be supplied with the remote equipment.

If an optional stereo generator is employed in the $830 \mathrm{E}-1 \mathrm{~A}$, remote control of the stereo mode may be accomplished by a ground on TB302-7. If the ground is present the transmitter will be in the stereo mode. If the ground is removed the transmitter will switch to the monaural mode. Local control of the stereo mode is also available at the transmitter.

### 2.8 FREQUENCY CHANGE.

If the transmitter operating frequency is changed, the following components will have to be changed or adjusted. The components are (1) the hetrodyning crystal, Y426, (2) the driver plate cavity slider, (3) the power amplifier grid tank slider, (4) the power amplifier plate tank slider, and (5) the neutralizing bars of the power amplifier.

Table 2-3 lists the channel frequency versus crystal frequency and the Collins part number for each crystal. Figure 2-3 shows the distance the driver plate cavity slider should be positioned from the deck plate (tube socket mounting plate) for each operating frequency between 88 and 108 megacycles. Figure 2-4 shows the distance the power amplifier plate cavity and grid cavity sliders should be positioned from the deck plate for the operating frequencies.

TABLE 2-3. CRYSTAL PART NUMBERS

| CHANNEL FREQ (mc) | $\begin{gathered} \text { CRYSTAL } \\ \text { FREQ } \\ \text { (mc) } \end{gathered}$ | COLLINS <br> PART NUMBER | $\begin{gathered} \text { CHANNEL } \\ \text { FREQ } \\ \text { (mc) } \end{gathered}$ | $\begin{gathered} \text { CRYSTAL } \\ \text { FREQ } \\ \text { (mc) } \end{gathered}$ | COLLINS PART NUMBER |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 88.1 | 74.10000 | 289-2744-00 | 91.1 | 77.10000 | 289-2759-00 |
| 88.3 | 74.30000 | 289-2745-00 | 91.3 | 77.30000 | 289-2760-00 |
| 88.5 | 74.50000 | 289-2746-00 | 91.5 | 77.50000 | 289-2761-00 |
| 88.7 | 74.70000 | 289-2747-00 | 91.7 | 77.70000 | 289-2762-00 |
| 88.9 | 74.90000 | 289-2748-00 | 91.9 | 77.90000 | 289-2763-00 |
| 89.1 | 75.10000 | 289-2749-00 | 92.1 | 78.10000 | 289-2764-00 |
| 89.3 | 75.30000 | 289-2750-00 | 92.3 | 78.30000 | 289-2765-00 |
| 89.5 | 75.50000 | 289-2751-00 | 92.5 | 78.50000 | 289-2766-00 |
| 89.7 | 75.70000 | 289-2752-00 | 92.7 | 78.70000 | 289-2767-00 |
| 89.9 | 75.90000 | 289-2753-00 | 92.9 | 78.90000 | 289-2768-00 |
| 90.1 | 76.10000 | 289-2754-00 | 93.1 | 79.10000 | 289-2769-00 |
| 90.3 | 76.30000 | 289-2755-00 | 93.3 | 79.30000 | 289-2770-00 |
| 90.5 | 76.50000 | 289-2756-00 | 93.5 | 79.50000 | 289-2771-00 |
| 90.7 | 76.70000 | 289-2757-00 | 93.7 | 79.70000 | 289-2772-00 |
| 90.9 | 76.90000 | 289-2758-00 | 93.9 | 79.90000 | 289-2773-00 |

TABLE 2-3. CRYSTAL PART NUMBERS (Cont)

| $\begin{gathered} \text { CHANNEL } \\ \text { FREQ } \\ \text { (mc) } \end{gathered}$ | $\begin{aligned} & \text { CRYSTAL } \\ & \text { FREQ } \\ & \text { (mc) } \end{aligned}$ | COLLINS <br> PART NUMBER | CHANNEL FREQ (mc) | $\begin{aligned} & \text { CRYSTAL } \\ & \text { FREQ } \\ & \text { (mc) } \end{aligned}$ | COLLINS <br> PART NUMBER |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 94.1 | 80.10000 | 289-2774-00 | 101.1 | 87.10000 | 289-2809-00 |
| 94.3 | 80.30000 | 289-2775-00 | 101.3 | 87.30000 | 289-2810-00 |
| 94.5 | 80.50000 | 289-2776-00 | 101.5 | 87.50000 | 289-2811-00 |
| 94.7 | 80.70000 | 289-2777-00 | 101.7 | 87.70000 | 289-2812-00 |
| 94.9 | 80.90000 | 289-2778-00 | 101.9 | 87.90000 | 289-2813-00 |
| 95.1 | 81.10000 | 289-2779-00 | 102.1 | 88.10000 | 289-2814-00 |
| 95.3 | 81.30000 | 289-2780-00 | 102.3 | 88.30000 | 289-2815-00 |
| 95.5 | 81.50000 | 289-2781-00 | 102.5 | 88.50000 | 289-2816-00 |
| 95.7 | 81.70000 | 289-2782-00 | 102.7 | 88.70000 | 289-2817-00 |
| 95.9 | 81.90000 | 289-2783-00 | 102.9 | 88.90000 | 289-2818-00 |
| 96.1 | 82.10000 | 289-2784-00 | 103.1 | 89.10000 | 289-2819-00 |
| 96.3 | 82.30000 | 289-2785-00 | 103.3 | 89.30000 | 289-2820-00 |
| 96.5 | 82.50000 | 289-2786-00 | 103.5 | 89.50000 | 289-2821-00 |
| 96.7 | 82.70000 | 289-2787-00 | 103.7 | 89.70000 | 289-2822-00 |
| 96.9 | 82.90000 | 289-2788-00 | 103.9 | 89.90000 | 289-2823-00 |
| 97.1 | 83.10000 | 289-2789-00 | 104.1 | 90.10000 | 289-2824-00 |
| 97.3 | 83.30000 | 289-2790-00 | 104.3 | 90.30000 | 289-2825-00 |
| 97.5 | 83.50000 | 289-2791-00 | 104.5 | 90.50000 | 289-2826-00 |
| 97.7 | 83.70000 | 289-2792-00 | 104.7 | 90.70000 | 289-2827-00 |
| 97.9 | 83.90000 | 289-2793-00 | 104.9 | 90.90000 | 289-2828-00 |
| 98.1 | 84.10000 | 289-2794-00 | 105.1 | 91.10000 | 289-2829-00 |
| 98.3 | 84.30000 | 289-2795-00 | 105.3 | 91.30000 | 289-2830-00 |
| 98.5 | 84.50000 | 289-2796-00 | 105.5 | 91.50000 | 289-2831-00 |
| 98.7 | 84.70000 | 289-2797-00 | 105.7 | 91.70000 | 289-2832-00 |
| 98.9 | 84.90000 | 289-2798-00 | 105.9 | 91.90000 | 289-2833-00 |
| 99.1 | 85.10000 | 289-2799-00 | 106.1 | 92.10000 | 289-2834-00 |
| 99.3 | 85.30000 | 289-2800-00 | 106.3 | 92.30000 | 289-2835-00 |
| 99.5 | 85.50000 | 289-2801-00 | 106.5 | 92.50000 | 289-2836-00 |
| 99.7 | 85.70000 | 289-2802-00 | 106.7 | 92.70000 | 289-2837-00 |
| 99.9 | 85.90000 | 289-2803-00 | 106.9 | 92.90000 | 289-2838-00 |
| 100.1 | 86.10000 | 289-2804-00 | 107.1 | 93.10000 | 289-2839-00 |
| 100.3 | 86.30000 | 289-2805-00 | 107.3 | 93.30000 | 289-2840-00 |
| 100.5 | 86.50000 | 289-2806-00 | 107.5 | 93.50000 | 289-2841-00 |
| 100.7 | 86.70000 | 289-2807-00 | 107.7 | 93.70000 | 289-2842-00 |
| 100.9 | 86.90000 | 289-2808-00 | 107.9 | 93.90000 | 289-2843-00 |

### 2.8.1 NEUTRALIZATION PROCEDURE.

## WARNING

Voltages are present in this transmitter that are dangerous to life. Observe safety precautions when performing any inspection or work within the cabinet or plate cavity. Do not depend entirely on the interlocks. Always shut down the transmitter before doing any work inside the transmitter cabinet. Remember 115 volts is present on one side of the door interlocks.

Power amplifier neutralization must be completed after any change in frequency to provide the proper degree of amplifier stability. Neutralization of the power amplifier is accomplished as follows: Set the neutralization sliding contacts to the position corresponding to the station frequency as shown in figure 2-5. Construct a detector as shown in figure 2-6 (parts are available from Collins Radio Company, Cedar Rapids, Iowa) and connect to the station monitor output located on the top of the 5 -kw power amplifier cabinet. Turn on the exciter and driver, and tune the exciter and driver according to the procedure given in paragraph 2.10. Tune the power amplifier grid tank. Adjust the power amplifier neutralization sliding contacts


PLATE CAVITY


Figure 2-3. Driver Plate Cavity Tuning Chart

©

Figure 2-4. Power Amplifier Plate and Grid Cavity Tuning Chart


Figure 2-5. Neutralization Tuning Chart


| PART | TYPE NO. OR <br> VALUE | COLLINS <br> PART NO. |
| :---: | :---: | :---: |
| CAPACITOR | $5-25 U U F$ | $917-1073-00$ |
| DIODE | IN67A | $353-0147-00$ |
| RESISTOR | IOK | $745-1394-00$ |
| INDUCTANCE | $3.9 U H$ | $240-1575-00$ |

Figure 2-6. Neutralization Detector
for a minimum indication on the detector meter. The neutralization can be checked by noting the grid reaction as the plate is tuned through resonance. Minimum grid current reaction indicates the neutralization is close. Neutralization of the 5 -kw power amplifier is normally not critical and should be easily adjusted.

### 2.9 FINAL INSTALLATION PROCEDURE.

a. Check that all tubes, both in the exciter and power amplifiers are securely in place.
b. Check all exciter transistors for proper placement. The transistor location tab must be pointing to the transistor socket locating mark placed adjacent to the transistor socket.
c. Check fuses on both the exciter and power amplifiers.
d. Insert the crystals into the exciter crystal sockets. The 14 -mc crystal should be placed into Y 501 and the hetrodyning crystal into Y426.
e. Carefully inspect all rear door and panel interlocks to be certain that they are operating correctly. Do this by pressing the contact block, located on the doors and lower rear panels, until the spring is completely compressed, then releasing the block. If the block does not spring back to its original position adjust until it operates properly.
f. Recheck all cables and wiring in the transmitter to be sure that all connections are made securely and properly.
g. Check that the sliders in the tuning cavities of both the driver and the power amplifier are in proper position for the station frequency.
h. Check the final power amplifier tube for proper seating. This tube is pressed into place and may have jarred loose during installation.
i. Replace the lower rear panels on the transmitter cabinets. Insert the ventilating fan plugs into the sockets provided.

### 2.10 STARTING THE TRANSMITTER IN A NEW INSTALLATION.

Before starting the transmitter for the first time, read section III of the subunit instruction books to become familiar with the location and function of the various transmitter controls. Then, perform the following procedures.

## WARNING

Voltages are present in this transmitter that are dangerous to life. Observe safety precautions when making any transmitter adjustments. Do not reach inside the rear of the transmitter cabinets whenever high voltages are applied. Do not depend entirely on door interlocks. Always shut off transmitter power at the external cutout box and ground all capacitors with the shorting stick in the transmitter cabinets before doing any work inside the rear of the cabinets. When working in the power amplifier cavities, remember that 115 volts a-c is present on one side of the cavity compartment interlocks. Keep metal tools and all parts of the body away from transistor cases.
a. Complete the entire transmitter installation procedure as directed in earlier paragraph of this instruction book.
b. Close the doors at the rear of the cabinets. Open the doors at the front of the cabinets and remove the lower front panels so that the entire inside panel is exposed. Check to be sure that the covers on the cavity compartments of both the driver and power amplifier are securely closed and latched.
c. Set the PLATE circuit breaker in the driver and the PLATE and FILAMENT breakers in the power amplifier to ON.
d. Press the FILAMENT ON switch on either cabinet. The green indicator lamps at the top left of both cabinets should light. This means that all transmitter tube filaments and cooling-air blowers are operating.
e. Set the right cabinet MULTIMETER switch to FIL V 8V FS. Adjust the FILAMENT VOLTAGE ADJUST control until the right cabinet MULTIMETER indicates 7.5 volts. If this meter indication cannot be reached, readjust the taps on the filament transformer, T403, to the next lower voltage tap.

## NOTE

The mercury-vapor rectifier tubes in the 5-kw amplifier (right cabinet) should be run with only the filaments lighted for at least one-half hour before applying plate voltage to the amplifier. This is done to remove any mercury coating that may be on tube elements.

This operation is necessary only for new tubes or used tubes that have been inverted or agitated. While this aging process is being completed, the exciter and 250 -watt amplifier may be tuned by performing the following steps.
f. Set the POWER switch on the exciter to ON. Allow the exciter oscillator to warm up for 15 minutes. g. Turn S101 to the 14MC REF B position and check M101 for an indication in the B meter range. Turn S101 to the AFC KEY B position and check M101 for an indication in the $B$ meter range.

## NOTE

When S101 is in the AFC KEY B position the meter pointer will not hold steady but will pulse at approximately 5 cps , the keying generator rate. This pulse is an indication of normal operation.

Turn S101 to the MOD OUTPUT B position and check M101 for an indication in the B meter range. If all meter indications for the three S101 positions fall within the $B$ meter range, proceed with the following tuning procedures. If any of the meter indications fall outside of the B meter range, the modulator and afc diseriminator are out of adjustment and will have to be adjusted according to the Modulator and AFC Discriminator Adjustment Procedures in the Maintenance section of this instruction book.
h. Switch S101 to the MIXER GRID A position. Set POWER OUT resistor R454 to its midposition. Using a nonmetallic screwdriver type tuning tool adjust the BUF TUNE control for a peak indication on meter M101.

## NOTE

The MIX BAL control, R438, should be placed in its midrange position. No further adjustment of this control is then necessary unless the transmitting frequency falls within the range of 97 to 100 mc . See step n if the transmitted frequency falls within this range.
i. Switch S101 to the V428B position. Adjust L429 and L430 for a maximum indication on M101.
j. Switch S101 to the V429B position. Adjust L431 and L432 for a maximum indication on M101.
k. Switch S101 to the V430B position. Adjust L433 and L434 for a maximum indication on M101.

1. Switch S101 to the V430C position. Adjust the PA PLATE control for a minimum indication on M101.
m. Set the driver MULTIMETER switch to GRID FS 40 MA. Adjust first the exciter PA MATCH control, then the power amplifier GRID TUNING control for a peak MULTIMETER indication.
n. If the transmitter frequency falls between 97 and 100 mc the following step will have to be completed. Place a grid dip meter tuned to 98 me near the exciter output. Adjust the MIX BAL control for a minimum output as indicated on the grid dip meter.
o. Set the driver POWER OUTPUT ADJUST control fully counterclockwise. Connect a 50 -ohm 250 -watt dummy load to the rf output connector of the driver.
p. Press the driver PLATE ON switch. The red indicator lamp at the top right of the cabinet should light, and the P.A. PLATE VOLTAGE meter should indicate $2100 \pm 100$ volts.
q. Set the driver MULTIMETER switch to SCREEN FS 400 VDC. The MULTIMETER should indicate $300 \pm 35$ volts.
r. Adjust the driver PLATE TUNING control for a dip in the P.A. PLATE CURRENT meter indication.
s. Set the driver WATTMETER switch to FORWARD. Adjust the OUTPUT COUPLING control for approximately 5 ma of screen current.
t. Turn the driver POWER OUTPUT ADJUST control approximately $2 / 3$ of its maximum clockwise rotation.
u. Increase the driver coupling a small amount by turning the OUTPUT COUPLING control clockwise until the PA screen current is reduced to approximately 10 ma. Adjust the PLATE TUNING control for a dip in the P.A. PLATE CURRENT meter indication. (The plate tuning capacitor should be near its center position when the dip in the driver plate current occurs. If the capacitor is not in this position move the plate slider in the appropriate direction and repeat step u.)
v. Rotate the POWER OUTPUT ADJUST control clockwise a small amount.
w. Repeat steps $u$ and $v$ until the P.A. PLATE CURRENT meter indicates a minimum indication and the WATTMETER indicates 250 watts. At this time the driver screen current should be not less than 5 ma nor more than 20 ma .
x . Turn off the driver plate power and disconnect the dummy load from the driver. Reconnect the coaxial cable from the driver to the power amplifier. Set the driver POWER OUTPUT ADJUST potentiometer fully counterclockwise.
y. Set the driver WATTMETER switch to REFLECTED. Adjust the power amplifier GRID TUNING control for minimum reflected power.

## NOTE

If a low value of reflected power cannot be obtained, change the setting of the slider in the power amplifier (the right cabinet) grid cavity slightly, adjust the position of the input tap and repeat step $y$.
z. Turn the driver POWER OUTPUT ADJUST control until the proper PA grid current is flowing (approximately 20 ma ).
aa. Set the power amplifier POWER OUTPUT ADJUST control in the maximum counter clockwise direction. Check the power amplifier filament voltage
and if necessary readjust the FILAMENT VOLTAGE ADJUST control for 7.5 volts as indicated on the power amplifier MULTIMETER. If the transmitter filaments have been energized for at least one-half hour, press the power amplifier PLATE ON switch. When this switch is pressed, the red indicator lamp at the top right of the power amplifier cabinet should light and the P.A. PLATE VOLTAGE should indicate $6400 \pm 200$ volts.
ab. Adjust the power amplifier PLATE TUNING control for a dip in the P.A. PLATE CURRENT meter indication. (The power amplifier plate tuning capacitor should be near its center position when the dip in the power amplifier plate current occurs. If the capacitor is not in this position move the power amplifier plate slider in the appropriate direction and repeat step ab.)

## CAUTION

The power amplifier P.A. PLATE CURRENT meter indication should never exceed 1.6 amperes (power supply rating).
ac. Set the power amplifier WATTMETER switch to REFLECTED. The swr must be less than $2: 1$ at all times. If the indication of reflected power is excessive, check the antenna and its associated r-f cable for possible troubles.
ad. Set the power amplifier WATTMETER switch to FORWARD. Adjust the OUTPUT COUPLING control for approximately 10 ma of screen current.
ae. Turn the power amplifier POWER OUTPUT ADJUST control to approximately its midpoint position. Recheck the driver power output for proper PA grid current.

## NOTE

The values of screen current given in the steps are only approximate. The tube dissipation will allow for greater variations in screen current without adversely affecting tube operation.
af. Increase the power amplifier coupling a small amount by turning the OUTPUT COUPLING control
clockwise until the PA screen current is reduced to approximately 15 ma . Adjust the PLATE TUNING control for a dip in the P.A. PLATE CURRENT meter indication.
ag. Rotate the power amplifier POWER OUTPUT ADJUST control clockwise a small amount.
ah. Repeat steps af and ag until the power amplifier P.A. PLATE CURRENT meter indicates the transmitter output is 5 kw as measured by the indirect method. At this time the PA screen current should be not less than 10 ma nor more than 30 ma . The indirect method of measuring power output is:

$$
\text { Power output }=I_{p} E_{p} K
$$

When $K$ is efficiency, $E_{p}$ is plate voltage, and $\mathrm{I}_{\mathrm{p}}$ is plate current.

## NOTE

$K$ or efficiency for determining the transmitter power output is obtained from the production test data supplied with the transmitter.
If the final does not tune up properly the fault may lie in improper neutralization. If faulty neutralization is suspected, neutralize the final according to the procedure given in paragraph 2.8.1.
ai. Check to be sure that the FM monitor that is connected to the transmitter is properly calibrated. Then, if necessary, adjust the exciter VHF OSC FREQ ADJ control until the monitor indicates that the transmitter operating frequency is within specified operating limits.
aj. Apply a $50-\mathrm{cps}$ audio tone to the transmitter input. The input level should be such that the voltage at the transmitter audio input terminals is +10 dbm .
ak. Adjust the exciter MOD GAIN control until the monitor indicates 100 percent modulation.
al. Replace the lower front panel on the transmitter cabinet and close the front doors. The transmitter is now ready for standard broadcast use.

## NOTE

At this point it is suggested that a record be made of all meter readings for future maintenance and trouble shooting. These meter readings may be recorded in table 5-3.

# SECTION III <br> OPERATION 

### 3.1 GENERAL.

Refer to the unit instruction books to become familiar with the operation and function of controls on the power amplifier the driver, and the exciter.

After the transmitter has been placed in operation it will only be necessary from time to time to check meter indications to be sure the transmitter is operating properly and to occasionally touch-up the power amplifier loading and tuning.

### 3.2 STARTING THE TRANSMITTER IN NORMAL OPERATION.

The transmitter may be put into operation by two different methods, depending upon the circumstances. For normal operation, press either FILAMENT ON switch to start the driver and power amplifier filaments and to warmup the exciter (the exciter POWER switch should be left in the ON position at all times). Check the driver grid current to be sure the exciter is presenting sufficient drive to the driver before applying plate power. Approximately thirty seconds after filament power is applied, press the driver PLATE ON
switch. Check the operation of the driver and, if operating properly, press the power amplifier PLATE ON switch, starting the transmitter.

The alternate method of starting the transmitter consists of pressing the power amplifier PLATE ON switch only. The power amplifier filament, the driver filament, and the exciter will immediately start to warm up. As soon as the driver time delay relay has completed its cycle, the driver and power amplifier plate voltages will come on automatically, starting the transmitter.

To shut down the transmitter it is recommended, but not necessary, to press the driver PLATE OFF switch, wait a few seconds, and then shut off the filament and exciter power by pressing either FILAMENT OFF switch. It is also possible to press either FILAMENT OFF switch which removes plate, filament, and exciter power. Pressing the driver PLATE OFF switch first allows the plate power supply voltages to discharge through the driver and power amplifier while the filament is at normal operating temperature and, in addition, cools the power amplifier components.

## SECTION IV PRINCIPLES OF OPERATION

### 4.1 GENERAL.

Refer to figure 4-1, a block diagram of 830E-1A 5-Kw FM Broadcast Transmitter. The transmitter can be broken down into four main subassemblies; an exciter, a driver, a power amplifier, and a harmonic filter. Refer to section II of the unit instruction books for a complete explanation of the exciter, the driver, and the power amplifier.

### 4.2 A830-2 10 W WIDE-BAND FM BROADCAST EXCITER.

Figure 4-1 is the block diagram of A830-2 10 W Wide-Band FM Exciter. The baseband audio is coupled to the A830-2 through a pre-emphasis network, and through an 18 -db pad to J101 and two baseband amplifiers. The baseband output is coupled to a voltagesensitive capacitor, C654. C654 is a diode which varies in capacity in proportion to the voltage across it. The FM oscillator is tuned to 14 mc . The capacity of C654 varies in proportion to the baseband audio and
therefore the output is a $14-\mathrm{mc}$ signal frequency modulated by the baseband audio. The deviation of the $14-\mathrm{mc}$ signal is $\pm 75 \mathrm{kc}$ for 100 percent modulation. The oscillator output is coupled through two limiters to remove any amplitude modulation. The limited $14-\mathrm{mc}$ signal is then amplified and coupled to the rate correction frequency discriminator and to the output amplifier. The output of the frequency discriminator is simply the baseband audio detected from the modulated $14-\mathrm{mc}$ signal. This detected audio is coupled back to the baseband input to correct for any nonlinearity in C654.

The output amplifier amplifies the modulated $14-\mathrm{mc}$ signal to a level sufficient to mix with the 74- to $94-\mathrm{mc}$ signal (per customer requirement) in the balanced mixer. A portion of the limiter output is coupled to the afc buffer stage. The afc buffer output, the modulated $14-\mathrm{mc}$ signal, is coupled to the reference oscillator and afc limiters through a diode switch. The output of the $14-\mathrm{mc}$ reference oscillator, is also coupled to the reference oscillator and afc
limiters through a diode switch. The diode switch is operated by a 5 -cps keying generator. The 5 -cps generator is a unijunction transistor operating as a relaxation oscillator keying a multivibrator.

The diode switch alternately connects the modulated $14-\mathrm{mc}$ signal (afc buffer output) and the $14-\mathrm{mc}$ reference signal. The limiter output is coupled to the afc discriminator. The afc discriminator detects the difference between the $14-\mathrm{mc}$ reference signal and the modulated $14-\mathrm{mc}$ signal. The modulated 14 -mc signal will cause a baseband audio output at the discriminator. This is not an error in frequency, so a portion of the baseband audio input is amplified by the baseband canceling amplifier and fed into the output of the frequency discriminator through a diode switch. This diode switch is keyed by the same 5 -cps signal which switched the reference oscillator and afc limiter input. When the modulated $14-\mathrm{mc}$ signal is connected to the reference oscillator and afc limiter input, the baseband canceling signal is switched into the output of the frequency disciminator to cancel the baseband output from the discriminator.

The input signal to the four error signal amplifiers is a 5 -cps square wave. The amplitude of this square is proportional to the frequency error in the FM oscillator. The error signal amplifier square wave output is converted to a d-c control signal in the synchronous detector. The synchronous detector is also keyed by the 5 -cps keying signal. The d-c error signal is coupled to C 654 to correct the frequency modulation oscillator.

The modulated $14-\mathrm{mc}$ signal from the output amplifier is heterodyned up to the operating frequency in a balanced mixer. The injection frequency is generated in a crystal oscillator. The crystal frequency is 14 mc below the customer's operating frequency. The crystal oscillator output is coupled to a buffer stage and is mixed with the modulated $14-\mathrm{mc}$ signal in the balanced mixer. The balanced mixer output is limited and amplified to the 10 -watt r-f output level. The output impedance of the A830-2 is between 50 and 70 ohms.

The power supply for the $\mathbf{A 8 3 0 - 2}$ is of conventional design and supplies operating voltages for the vacuum tubes and transistors in the A830-2.

The driver consists of a single ceramic-type tetrode tube. The tube is operated as a class $C$ amplifier with a tuned-cavity plate circuit. The output from the driver is fed to a 5 -kw final amplifier.

The final power amplifier is also made up of a single ceramic-type tetrode tube. The tube is operated as a
grounded screen, class $C$ amplifier, using screen grid neutralization. The plate works into a tuned cavity with a tuned cavity control grid tank. The plate tuned cavity output is fed through a harmonic filter which reduces all output harmonics into the antenna.

The harmonic filter consists of two series resonant M -derived low-pass end sections and two constant K $T$ center sections. The harmonic filter starts to attenuate above 110 mc and reaches maximum attenuation at the carrier second harmonic. The attenuation pattern then tapers off slowly as the frequency rises. The over-all result of the harmonic filter is in keeping the harmonics attenuated at least 80 db below the carrier frequency.

### 4.3 CONTROL CIRCUITS.

One phase of the 230 -volt 3 -phase power is stepped down to 115 volts a-c by transformers T301 and T401. This lower voltage is used to activate relays in the transmitter control circuits and is also fed to the exciter as its primary power source. The control circuits allow power to be applied to the transmitter only in the proper sequence to prevent damage to the driver and final amplifier. These circuits also contain protective devices to prevent damage to components from accidental overloads.

### 4.4 PLATE CONTACTORS AND POWER SUPPLIES.

4.4.1 The driver plate contactor consists of a heavy duty relay which controls the 230 -volt a-c primary power to the plate power supply. The driver plate contactor is actuated by the driver PLATE ON switch through the control circuitry.

The driver plate power supply consists of a step-up transformer, a full-wave rectifier, and a filter. The power supply is capable of delivering 2100 volts of $\mathrm{d}-\mathrm{c}$ at 250 ma to the driver.
4.4.2 The power amplifier plate contactors are heavyduty relays similar to the type used in the driver. The plate contactors control the 3 -phase power to the plate power supply and are activated by the power amplifier PLATE ON switch. The two plate contactors are energized in sequence to prevent transients from entering the power supply when power is first applied.

The power amplifier plate power supply is a conventional 3-phase full-wave mercury vapor power supply capable of providing 6400 volts at 1.6 amp . The screen power supply is a full-wave 3 -phase silicon diode power supply capable of providing 600 volts at 1.8 amp. The control grid bias supply places the control grid at about cutoff and is set at the factory.


# SECTION V MAINTENANCE 

### 5.1 GENERAL.

The following paragraphs contain information concerning maintenance of $830 \mathrm{E}-1 \mathrm{~A} 5-\mathrm{Kw}$ FM Broadcast Transmitter.

## WARNING

Voltages are present in this transmitter that are dangerous to life. Observe safety precautions when performing any maintenance. Do not reach inside the transmitter cabinets whenever high voltage is applied. Do not depend on door interlocks. Always shut down the transmitter before doing any work inside the transmitter cabinets. Immediately upon opening the rear cabinet doors, short out the power supply capacitors with the shorting sticks provided with the transmitter.

Refer to the applicable unit instructions for specific maintenance procedures for each subassembly.

### 5.2 NORMAL TUNING PROCEDURES.

The following are tuning procedures which should be observed after the transmitter has been installed and tuned according to the installation procedures given in section II. Table 5-1 presents abbreviated tuning instructions to be used with the following procedures. Table 5-1 can be detached from this instruction book, placed within one of the cabinet doors or adjacent to the transmitter and used separately when the operator becomes familiar with the transmitter.
a. Open the doors at the front of the cabinets.
b. Press either FILAMENT ON switch and allow the exciter to warm up for at least 15 minutes.
c. Set S101 on the exciter to MIXER GRID A. Adjust BUF TUNE control for a peak on MiOl.
d. Set S101 to V428 B and adjust L429 and L430 for a peak on M101.
e. Set S101 to V429 B and adjust L431 and L432 for a peak on M101.
f. Set S101 to V430B and adjust L433 and L434 for a peak on M101.
g. Set S101 to V430C B and adjust the PA PLATE control for a minimum indication on M101.

## NOTE

Convenient marked ranges are available on the meter which correspond to switch positions.

These ranges give an approximate requirement for min-max readings for each switch position.
h. Tune the exciter coupling by setting the MULTIMETER switch to GRID FS 40 MA and adjusting the exciter PA MATCH control for maximum coupling. Tune the driver GRID TUNING control for a peak MULTIMETER indication.
i. Set the driver WATTMETER to REFLECTED.
j. Press the driver PLATE ON switch. Adjust the power amplifier GRID TUNING for a dip in the R.F. WATTMETER indication. Set the driver WATTMETER to FORWARD. In turn, adjust the driver PLATE TUNING control for minimum indication and increase the OUTPUT COUPLING control and the POWER OUTPUT ADJUST control, by small amounts, until the proper value of PA grid current in flowing (approximately 20 ma ). The driver screen current may be 0 but should not be more than 20 ma when the driver is properly tuned.
k. Press the power amplifier PLATE ON switch.

1. Adjust the FILAMENT VOLTAGE ADJUST control for 7.5 voits as shown on the MULTIMETER.
m . Adjust the power amplifier GRD TUNING control for a maximum indication on the MULTIMETER. (This should be near the minimum reflected power on the driver wattmeter.)
n. In turn, adjust the power amplifier PLATE TUNING control for minimum plate current indication and increase the OUTPUT COUPLING control and the POWER OUTPUT ADJUST control, by small amounts, until the transmitter output power is 5 kw as measured by the indirect method.

$$
\text { Power output }=\mathrm{Ip}_{\mathrm{p}} \mathrm{EK}
$$

The power amplifier screen current should not be less than 10 ma or more than 30 ma when the power amplifier is properly tuned. Normal operation is achieved when the R.F. WATTMETER indication is maximum for near minimum power amplifier plate current.
o. Adjust the VHF OSC FREQ ADJ control until the FM monitor indicates the transmitter is operating within the specified operating limits.

### 5.3 MODULATOR AND AFC DISCRIMINATOR ADJUSTMENT PROCEDURES.

The broadband exciter is designed to be exceptionally stable and will require few adjustments over a long period of time. The following adjustment procedures should only be followed if the exciter is not operating within limits upon installation, or if any of the transistors (Q503, Q511, Q601, or Q604) are replaced.

TABLE 5-1. ABBREVIATED TUNING PROCEDURES

|  | CONTROL | POSITION | ADJUSTMENT | INDICATING METER | INDICATION | NOTES <br> Allow transmitter to warm up at least 15 minutes before tuning. |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | S101 | MIXER GRID A | BUF TUNE | M101 | Maximum |  |
|  | S101 | V428 B | *L429, L430 | M101 | Maximum |  |
|  | S101 | V429 B | *L431, L432 | M101 | Maximum |  |
|  | S101 | V430B | *L433, L434 | M101 | Minimum |  |
|  | S101 | V430C B | PA (exciter) PLATE | M101 | Minimum |  |
| $\begin{aligned} & \text { 㽞 } \\ & \text { 品 } \\ & \end{aligned}$ | MULTIMETER | GRID FS 40 MA | PA MATCH GRID TUNING | MULTIMETER | Maximum |  |
|  | WATTMETER | FORWARD | PLATE TUNING <br> OUTPUT COUPLING <br> POWER OUTPUT ADJUST | P.A. PLATE CURRENT <br> R.F. WATTMETER <br> R.F. WATTMETER | Minimum <br> Maximum <br> 250 watts or less depending on drive requirements | Repeat the adjustment of PLATE TUNING, OUTPUT COUPLING, and POWER OUTPUT ADJUST controls until proper power amplifier drive is obtained. |
|  | WATTMETER | FORWARD | PLATE TUNING <br> OUTPUT COUPLING <br> POWER OUTPUT AD.JUST | R.F. WATTMETER <br> R.F. WATTMETER | Near minimum <br> Near 5-kw indication <br> Near 5-kc indication | Repeat the adjustment of PLATE TUNING, OUTPUT COUPLING, and POWER OUTPUT ADJUST controls until 5 kw is achieved by the indirect power measing method. $P=I_{p} E_{p} K$ |
| *Use slotted nonmetallic screwdriver on these adjustments. |  |  |  |  |  |  |

Replacement of other components should not normally require a change in the adjustments given in this section. The transmitter will have to be energized when performing steps $c$ and $e$. Use ana-c vtvm such as a Ballantine Model 310A when making adjustments. A d-c vtvm such as a Heath type may be used in step $f$.

## NOTE

When adjusting the modulator and afc section of the exciter, use a nonmetallic hex type adjusting tool. Keep all metallic tools and the hands or other parts of the body away from transistor cases. When disabling the afc, and the complete transmitter is in operation, check the station monitor to be sure the center frequency stays within the FCC requirements.
a. Remove transistor Q509 and place a vtvm from TP501 to ground. Tune L505 for a peak indication on the vtvm and tune $L 504$ for a peak indication. Be sure to tune the inductances in the order givento minimize the limiting effect. Check that each stage is limiting when making these adjustments. Limiting will show up as a broad flat peak on the vtivm when tuning. Set the controls midway between the limiter fall off points shown on the vtvm. Do not replace Q 509 at this time.
b. Tune the afc discriminator by placing a vtvm from TP501 to ground and tune the DISCR PRI control, C515, for a maximum indication. Place the vtem from TP502 to ground and adjust the DISCR SEC control, C518, for 0 on the vtum.
c. Disable the afc by pressing the AFC DISABLE switch and adjust the OSC FREQ control until the station monitor indicates that the exciter is on frequency. Replace Q509.
d. Remove Q607. Place a vtvm between TP602 and ground. Tune L606 and L603 for a maximum indication on the vtvm. Tune the inductors in the order shown to minimize the effects of the limiter.
e. Remove afc by holding down the AFC DISABLE switch. With the vtvm from TP602 and ground, tune the DISCR PRI control, C639, for a maximum indication on the vtvm. Move the vtvm to TP601 and tune the DISCR SEC control, C644, for a 0 indication when the modulation monitor indicates the exciter is approximately on frequency. Repeat the tuning of the DISCR PRI and DISCR SEC controls. Replace Q607.
f. Place a d-c vtvm between TP603 and ground. Set the AMP BIAS control for a 7.5 -volt indication on the vtvm.
g. Place a vtvm between TP504 and ground. Remove Q510. Tune L611 and L608 for a maximum indication on the vtvm. Replace Q510.
h. With a vtvm on TP504 adjust the REF LEVEL control for an equal indication on the vtum with first Q510 removed and then $\mathcal{Z} 09$ removed. (This equalizes the modulator oscillator voltage and the $14-\mathrm{mc}$ reference voltage.) Replace the transistors.
i. Place an oscilloscope between TP503 and ground. Apply a 150 -cps audio signal on J601. Adjust the MOD BAL control for a minimum $150-\mathrm{cps}$ indication as shown on the oscilloscope.

## NOTE

The MOD BAL control must be adjusted slowly to allow the error signal amplifiers to stablize between adjustments.

### 5.4 DISTORTION TESTING PROCEDURE.

a. Refer to figure 5-1. Connect an audio frequency signal generator, such as a Hewlett-Packard Model


Figure 5-1. Distortion Test Setup

600D to the exciter audio input, terminals 1 and 2 of TB305. (Disconnect the station console audio input leads when making this connection.) Connect a distortion and noise meter, such as a Hewlett-Packard Model 330D, to the broadcast monitor. Comnect a 50 -ohm artificial load to the r-f output connector located on top of the transmiiter cabinet. Turn on the transmitter. b. Apply a 50 -cps audio tone to the transmitter input. The input level should be such that the voltage at the transmitter audio input terminals is $+10 \pm 2 \mathrm{dbm}$.
c. Adjust the exciter MOD GAIN control until the monitor indicates 100 percent modulation $( \pm 75-\mathrm{kc}$ deviation).
d. Measure the distortion at the frequencies and modulation levels given in table 5-2. The distortion shall be less than 1.5 percent for frequencies between 50 and 100 cps , less than 1.0 percent for frequencies between 100 and 7500 cps, and less than 1.5 percent for frequencies between 7500 and $15,000 \mathrm{cps}$.

TABLE 5-2. DISTORTION CHECKS

| FREQUENCY | DISTORTION IN PERCENT |  |  |
| :---: | :---: | :---: | :---: |
|  | $25 \%$ MODULATION | $50 \%$ MODULATION | $100 \%$ MODULATION |
| 50 |  |  |  |
| 100 |  |  |  |
| 400 |  |  |  |
| 1000 |  |  |  |
| 5000 |  |  |  |
| 7500 |  |  |  |
| 10,000 |  |  |  |
| 15,000 |  |  |  |

### 5.5 AUDIO FREQUENCY RESPONSE MEASUREMENTS.

a. Refer to figure 5-2. Connect an audio frequency signal generator, such as a Hewlett-Packard Model 600D, to terminals 1 and 2 of terminal board TB305. (Disconnect the station console audio input leads when making these measurements.) Connect a vacuumtube voltmeter, such as a Ballantine Model 310A, to the audio output terminals of the audio frequency generator. Connect a 50 -ohm artificial load to the r-f output connector located on top of the transmitter cabinet. Turn on the transmitter.
b. Check the audio frequency response of the transmitter by modulating the transmitter at $50,100,400$, $1000,5000,7500,10,000$, and 15,000 cps for 25 percent, 50 percent, and 100 percent modulation. Audio frequency response is measured by keeping the percentage of modulation constant and measuring the magnitude of audio, at each frequency given, to give the desired percentage of modulation. The audio frequency response must fall within the limits given in figure 5-3.

## NOTE

When taking audio frequency response measurements a broadcast monitor, such as a

Hewlett-Packard Model 335B, should be used. Do not use an instrument where audio deemphasis might give a false indication of peak modulation.

### 5.6 FM NOISE MEASUREMENT.

a. Refer to figure 5-4. Connect an audio frequency signal generator, such as a Hewlett-Packard Model 600D, to terminals 1 and 2 of terminal board TB- 305. (Disconnect the station console audio input leads when making this measurement.) Connect a vacuum-tube voltmeter to the output terminals of the broadcast monitor. Connect an artificial load to the r-f output connector located on top of the power amplifier cabinet. Turn on the transmitter.
b. Modulate the transmitter 100 percent $( \pm 75-\mathrm{kc}$ deviation) with 400 cps of audio.
c. Remove the modulating 400 cps and read the residual $F M$ noise on the vacuum-tube voltmeter. The residual $F M$ noise shall be less than -65 db below 100 percent modulation.

### 5.7 AM NOISE MEASUREMENTS.

a. Refer to figure 5-5. Short out terminals 1 and 2 of terminal board TB305. Connect a vacuum-tube


Figure 5-2. Audio Frequency Response, Test Setup


Figure 5-3. Audio Frequency Response Limits


Figure 5-4. FM Noise Test Setup


Figure 5-5. AM Noise Test Setup
voltmeter to connector J3 of the Hewlett-Packard Model 335 B broadcast monitor. Connect a 50 -ohm artificial load to the r-f output comnector located on top of the power amplifier cabinet. Turn on the transmitter.
b. Switch the broadcast monitor to measure carrier level.
c. Measure the AM noise in db at J 3 of the broadcast monitor in the following manner. Set modulation monitor to Carrier Level and measure the d-c level on the modulation meter ( 100 percent on scale equals 10 volts). Connect the vacuum-tube voltmeter to J3 and terminate J 3 with a 2 -megohm resistor. Measure the a-c level on the vacuum-tube voltmeter. (The input to the vacuum-tube voltmeter should be a shielded cable having less than 100 uuf distributed capacitance.) The AM noise is the direct ratio of the d-c reading and the a-c level. The AM noise shall be not less than -55 db below voltage or d -c carrier level.

$$
\text { AM noise }=20 \log \frac{d-c \text { reading }}{a-c \text { reading }}
$$

### 5.8 TROUBLE SHOOTING.

Standard trouble-shooting procedures should be used in finding malfunctions in the transmitter. As is suggested in TD-536, TD-538, and TD-539, meter indications for all functions should be recorded when the transmitter is installed and operating properly. Table $5-3$ is supplied for recording these readings. If some malfunction should occur after the normal meter readings are recorded, it is a simple matter to compare the meter readings of the malfunctioning equipment with the normal meter readings. When
trouble-shooting and comparing the meter readings it is advisable to start with the final stage and proceed backwards until normal readings are encountered. The malfunctioning stage will then be the one immediately ahead of the normal meter indications.

As most cases of trouble will be traced to tubes or transistors, it is advisable to first of all replace the tube (or transistor) in the stage in which the trouble is suspected. If the trouble does not clear with tube or transistor replacement, it will become necessary to take resistance or voltage measurements, within the suspected circuit, to determine which component has failed.

When tracing trouble within the power amplifier it will be helpful to use the "from-to" information given in unit instructions, TD-528 and TD-539. The "fromto" information gives the actual location of the individual wires within the transmitter cabinets. When used in conjunction with the schematics, the "from-to" information can be very helpful.

If the transmitter center frequency shifts excessively with modulation, the trouble may be isolated to either the afc circuitry or the modulator circuitry of the exciter by disabling the afc and noting if the carrier shifts more than 1.8 kc with a change in modulation from 0 to 100 percent. If the modulator oscillator shifts more than $1.8-\mathrm{kc}$ with the afc disabled, the trouble will be within the modulator oscillator circuits. The afc circuitry cannot shift the modulator oscillator frequency more than 1.8 kc . If the carrier shift is under 1.8 kc , the trouble will be in the afc circuitry.

TABLE 5-3. NORMAL TRANSMITTER METER INDICATIONS

|  | CONTROL | POSITION | METER | INDICATION |
| :---: | :---: | :---: | :---: | :---: |
|  | S101 | BUFFER GRID A | M101 |  |
|  | S101 | MIXER GRID A | M101 |  |
|  | S101 | V428 B | M101 |  |
|  | S101 | V429 B | M101 |  |
|  | S101 | V430 B | M101 |  |
|  | S101 | V430C B | M101 |  |
|  | S101 | MOD OUTPUT B | M101 |  |
|  | S101 | AFC KEY B | M101 |  |
|  | S101 | 14MC REF B | M101 |  |
| $\begin{aligned} & \text { 品 } \\ & \text { 品 } \\ & \text { 1 } \end{aligned}$ | MULTIMETER | SCREEN FS 400 VDC | MULTIMETER |  |
|  | MULTIMETER | SCREEN FS 40 MA | MULTIMETER |  |
|  | MULTIMETER | GRID FS 40 MA | MULTIMETER |  |
|  |  |  | P.A. PLATE VOLTAGE |  |
|  | $3$ |  | P.A. PLATE CURRENT |  |
|  | WATTMETER | FORWARD | R.F. WATTMETER |  |
|  | WATTMETER | REFLECTED | R.F. WATTMETER |  |
|  | MULTIMETER | BIAS V400V FS | MULTIMETER |  |
|  | MULTIMETER | GRID I 40MA FS | MULTIMETER |  |
|  | MULTIMETER | SCREEN I 80MA FS | MULTIMETER |  |
|  | MULTIMETER | EXTERNAL NO 1 | MULTIMETER |  |
|  | MULTIMETER | EXTERNAL NO 2 | MULTIMETER |  |
|  |  |  | P.A. PLATE VOLTAGE |  |
|  |  |  | P.A. PLATE CURRENT |  |
|  | WATTMETER | FORW ARD | R.F. WATTMETER |  |
|  | WATTMETER | REFLECTED | R.F. WATTMETER |  |

0

## E830-1 5-KW FM Power Amplifier

## unit instructions

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## SECTION: GENERAL DESCRUPTON

### 1.1 PURPOSE OF INSTRUCTION BOOK.

Unit Instructions TD-539 provides information about E830-1 5-Kw FM Power Amplifier. Information which is furnished covers a general description of the equipment, principles of operation, maintenance procedures, and a parts list.

### 1.2 PURPOSE OF EQUIPMENT.

The E830-1 5-Kw FM Power Amplifier is used for continuous monaural or stereophonic FM broadcast service on a single frequency, in the range from 88 to 108 megacycles with an output power of 5000 watts.

### 1.3 DESCRIPTION OF EQUIPMENT.

### 1.3.1 PHYSICAL DESCRIPTION.

The E830-1 5-Kw FM Power Amplifier, shown in figure 1-1, is contained in a single cabinet that is 38 inches wide, 76 inches high, 27 inches deep, and weighs approximately 1140 pounds. All power amplifier operating controls are located behind the doors on the front of the cabinet. The filament and plate on-off controls and four monitoring meters are located at the top front of the cabinet. The meters may be observed easily while operating the tuning controls. The power amplifier uses 1 r-f amplifier tube and 6 rectifier tubes. The r-£ amplifier tube is accessible from the front of the power amplifier. The bottom front of the power amplifier cabinet is removable to allow access to components on the bottom of the inside panel.

Large doors at the upper rear of the cabinet (see figure 1-2) allow access to the upper part of the power amplifier for servicing and maintenance. The lower rear half of the power amplifier cabinet is covered by a removable panel that contains a ventilating fan and a permanent type air filter. Operating personnel are protected by both electrical and mechanical interlocks on the rear doors and panel. These interlocks remove the plate voltage and ground the high-voltage circuits when the doors are opened or the panel is removed. The power amplifier plate tuning and grid tuning cavities are located in an interlocked compartment at the front of the transmitter.
Inside the E830-1, heavy iron-core components are at the bottom of the cabinet. An optional harmonic filter can be attached to the E830-1 within the confines of the rear cabinet compartment.

Cooling air for the power amplifier is drawn through a permanent type air filter at the rear of the cabinet by a low-speed, high-volume fan, and exhausted through a shielded opening in the top of the cabinet. A single
high-volume blower supplies cooling air directly to the power amplifier tube.

### 1.3.2 ELECTRICAL DESCRIPTION.

The E830-1 5-Kw FM Power Amplifier consists of a single, air-cooled power amplifier tube capable of being driven to full power by a 250 -watt driver. All


Figure 1-1. E830-1 5-Kw FM Power Amplifier, Over-all View
the actual remote control available from one source for both units. Instruction books covering the exciter and driver used in conjunction with the E830-15-Kw FM Power Amplifier are listed in table 1-1.

TABLE 1-1
ASSOCIATED EQUIPMENT INSTRUCTION BOOKS

| ASSOCIATED EQUIPMENT | INSTRUCTION BOOK |
| :--- | :---: |
| A830-2 10 W Wide-Band <br> FM Broadcast Exciter | TD-536 |
| $786 \mathrm{M}-1$ Stereo Generator | TD- 537 |
| B830-1 250 -Watt FM <br> Power Amplifier | TD- 538 |

### 1.4 EQUIPMENT SUPPLIED.

Table 1-2 lists equipment that is supplied as part of the E830-1 5-Kw FM Power Amplifier.

TABLE 1-2
EQUIPMENT SUPPLIED

| EQUIPMENT | COLLINS PART <br> NUMBER |
| :---: | :---: |
| E830-1 5-Kw FM Power <br> Amplifier <br> $5-K w ~ H a r m o n i c ~ F i l t e r ~$ | $549-2009-00$ |

### 1.5 EQUIPMENT REQUIRED BUT NOT SUPPLIED.

Table 1-3 lists equipment that is required for operation of E830-1 5-Kw FM Power Amplifier, but is not supplied as part of the power amplifier.

TABLE 1-3
EQUIPMENT REQUIRED BUT NOT SUPPLIED

| EQUIPMENT | COLLINS PART <br> NUMBER |
| :---: | :---: |
| A830-2 10 W Wide-Band <br> FM Broadcast Exciter | $522-2714-00$ |
| B830-1 250-Watt FM <br> Power Amplifier | $549-2008-00$ |

associated power supply and control circuitry is included for operation of the E830-1. The E830-1 input impedance is 50 ohms nominal, unbalanced. The E830-1 output power is at least 5000 watts over the frequency range of 88 to 108 megacycles into a 50 -ohm load with an swr not exceeding 2:1.

Line power input required is 60 cycle, 3 phase with primary taps on all power transformers to compensate for line voltage variations from 200 to 250 volts. Circuit breakers in the input side of the plate and control circuits are provided for primary current overload protection. The control circuit auxiliary power supply transformer secondary and the control grid bias supply are fused. Time delay circuitry for protection of the high voltage rectifier tubes and the power amplifier tube during warmup is provided, with the actual time delay control received from the driver time delay relay. Circuits are provided for possible remote control tie-in with the driver remote control circuits, with


Figure 1-2. E830-1 Rear View with Doors Open and Lower Panel Removed

### 1.6 EQUIPMENT SPECIFICATIONS.

Output impedance . . 50 ohms , nominal unbalanced.

### 1.6.1 MECHANICAL.



Frequency range . . 88 to 108 megacycles. Exact operating frequency determined by frequency of exciter.

### 1.7 TUBE AND SEMICONDUCTOR COMPLEMENT.

Table 1-4 lists the tube and semiconductor complement supplied as part of the E830-1 5-Kw FM Power Amplifier.

TABLE 1-4
TUBE AND SEMICONDUCTOR COMPLEMENT

| QUA NTITY | TYPE | FUNCTION |
| :---: | :--- | :--- |
| 6 | $872 \mathrm{~A} / 872$ | Plate voltage rectifiers |
| 1 | $4 \mathrm{CX5000A}$ | Power amplifier |
| 8 | 1 N540 | Control grid bias <br> rectifiers |
| 4 | 1 N540 | Multimeter rectifiers <br> MR326 |
| Screen grid voltage <br> rectifiers |  |  |

## SECTION II PRINCIPLES OF OPERATION

### 2.1 GENERAL.

The E830-1 5-Kw FM Power Amplifier contains a power amplifier and associated circuitry for the amplification of 250 watts of $r-f$ drive to 5 kilowatts of r-f power. The E830-1 output can be fed through a harmonic filter, for the attenuation of spurious radiations, to an antema or to a higher power amplifier.

Refer to figure 2-1, a block diagram of the E830-1 5-Kw FM Power Amplifier. The 230 -volt, $60-\mathrm{cps}$, 3 -phase line input is fed to the plate and screen step start contactors, where plate and screen voltage onoff functions are controlled by the control circuits. The 230 -volt 3 -phase line input is also fed to an auxiliary power supply, T401, where part of the 230volt input is reduced to 115 volts single phase. The single phase output of T401 is then fed to the control circuits. The control circuits turn on and off the plate, screen, filament, and control grid bias supply. The control circuits also feed 230 -volt, 60 -cps, 1 -phase power to the driver. Provisions are made within the control circuitry to connect to the filament on-off and plate on-off functions of a higher power amplifier
if one is available. The higher power amplifier could then control the filament on-off and plate on-off functions.

The power amplifier consists of a single, ceramic type, forced air-cooled, grounded screen tube working into a resonant quarter wave-length cavity resonator. The control grid circuit consists of a resonant cavity with a swamping resistor in parallel to provide a low impedance broadband load to the control grid. Metering circuits are included to measure filament voltage, screen grid bias voltage, control grid current, and screen current.

### 2.1.1 CONTROL CIRCUITS.

Refer to figure 2-2, a simplified schematic diagram of the control circuits of E830-1 5-Kw FM Power Amplifier. The control circuits perform the function of applying or removing filament, plate, screen, and grid bias voltages, turn on and off the cabinet fan and power amplifier blowers, and also provide overload protection for all power amplifier circuits. These functions will be discussed in the following paragraphs.


Figure 2-1. E830-1 5-Kw FM Power Amplifier, Block Diagram

The 230 -volt 3 -phase power enters the cabinet on TB401 and is fed to two circuit breakers, CB401 and CB402. The PLATE circuit breaker, CB402, protects only the plate and screen supplies, while the CONTROL circuit breaker, CB401, protects the remaining power amplifier circuitry. The 3 -phase output from CB402 is fed to two contactors to provide a voltage stepping action when plate and screen power is applied. This stepping action reduces the transients generated in the plate and screen power supplies when the E830-1 is turned on or off. Switch S 411 is placed in the screen supply 3 -phase input to provide an aid when trouble shooting the transmitter. Screen voltage can be shut off, further isolating possible trouble.

The 3-phase output from the control circuit breaker, CB401, is split into single phase components and fed to the filament contactor, the blower relay, the auxiliary power supply T401, and to the 250 -watt driver.

Power to operate the filament and blower relays is obtained from the 115 -volt secondary of T401. A CONTROL indicator, DS403, is lighted when the 115volt power is present at the secondary of T401. This power is fed through the FILAMENT OFF switch to the 250 -watt driver control circuitry at TB403-3. The driver control circuitry can interrupt power at this point if filament off functions are to be controlled from the driver. Filament and blower control power is routed through the driver and re-enters the E830-1 at TB403-8. The power then is fed directly to the filament and blower control relays.

If the FILAMENT ON switch, S112, is pressed on either the driver or the E830-1, a ground is placed on the blower relay, K401, causing the relay to close. This action starts the power amplifier blower B401,
and closes contacts K401-7 and K401-8. Closing of these contacts holds a ground on K401 when the momentary FILAMENT ON switch is released. When PA blower B401 comes up to speed, air pressure activated switch 5407 closes, causing filament relay K402 to energize. The green filament-on light, DS401 will come on, the cabinet blower will start, and filament power will be fed to the tubes. The control grid bias supply will also be energized from the primary of transformer T401.

Power for operation of the plate circuits is obtained from the same source as the filament power transformer, T401. To start the plate-on sequence, PLATE ON switch S 113 is pressed actuating the plate hold relay K403. This closes contacts 10 and 12 to hold K403 energized when the PLATE ONswitch is released. The ground is obtained from the driver control circuitry to allow the driver to interrupt the ground for the plate-off function. Contacts 7 and 9 will also close, turning on the driver plate power ifdriver plate power has not been turned on previously. With K403 energized, contacts 1 and 3 will make, transferring the control voltage to the time delay transfer relay K404. K404 will actuate when the driver time delay relay has completed its cycle allowing the filaments to warm up to operating temperature. With K404 energized, 115 volts will be transferred to the plate-on contactor, K405. A fraction of a second later (as soon as K405-11 and 12 close) the step-start contactor will actuate applying 230 volts 3 phase to the plate and screen power supplies. The red plate-on indicator, DS402, will light indicating that plate and screen power is applied to the power amplifier.

If the PLATE OFF switch, S114, is pressed, plate and screen voltage will only be removed from the


power amplifier and will not normally affect the driver. Pressing the PLATE OFF switch, S114, opens the ground to plate-hold relay K403, releasing K403. Step-start contactor, K406, will then release removing direct 3 -phase power from plate and screen power supplies and momentarily throwing dropping resistors R401, R402, and R403 into the plate-screen circuits. Moments after step-start contactor K406 opens, plateon contactor K 405 will open, removing all power from the plate and screen power supplies. This stepping action limits transients that would normally be introduced into the power supply from the sudden shutting off of power.

If FILAMENT OFF switch S111 is pressed, power will momentarily be removed from all circuits within the driver and power amplifier causing the holding relays to drop out, shutting off the power amplifier and driver.

### 2.1.2 POWERAMPLIFIER CIRCUITS.

Refer to figure 2-3, a simplified schematic of the power amplifier circuitry of E830-1 5-Kw FM Power Amplifier. The power amplifier consists of a single ceramic type, forced air-cooled tube working into a tuned cavity over the standard frequency modulated broadcast band of 88 to 108 megacycles.

Power amplifier V 207 is a fixed bias, class C , grounded screen tetrode. The input from the $250-$ watt driver is tied into a resonant tuned cavity (a foreshortened $\frac{\lambda}{4}$ resonator) and fed to the control grid. Resistor R468 swamps out much of the driver power which is fed into the tuned cavity and, in addition, presents a low impedance broadband load to the control grid. Initial tuning of the grid cavity is accomplished by a slider which physically lengthens or shortens the grid cavity. Grid cavity tuning is accomplished by adjusting variable capacitor C436. Fixed capacitors C437, C438, C439, C440, C441, and C442 in parallel with C436 couple the input tank to pa grid. As the power amplifier is a grounded screen amplifier necessitating the filament be below ground potential, the grid bias supply is floating and is connected in series with the screen grid power supply giving negative control grid bias with respect to the filament. Resistor R473 and capacitor C434 form a parasitic suppressor (parasitics around 200 megacycles) while L405 blocks r-f from the grid bias supply.

The plate works into a tuned cavity (foreshortened $\frac{\lambda}{4}$ resonator which is similar to the grid cavity. The plate cavity consists of a short piece of coaxial transmission line which resonates with the plate capacity of V407 and PLATE TUNING capacitor C445. Initial tuning is similar to the grid cavity, and is accomplished by a shorting slider on the transmission line which physically lengthens or shortens the transmission line. Parasitic suppressors located within the cavity (C449, LA10, R484, and L411) suppress resonances around 200 and 400 megacycles. The parasitic suppressor capacity is formed by the distributed capacity of resistors, R474 and R484, and the cavity walls.

R-f output coupling is achieved by a movable plate within the cavity, forming a capacity between the coaxial transmission line and the movable metal plate. This capacitor is C444. R-f output from C444 is then fed to a directional coupler if a higher power amplifier is fed, or to a harmonic filter and through the directional coupler if an antenna is fed directly. Suppressor L408, R471, and C446 effectively damps the high order resonances of the plate tank circuit.

As stated previously, the power amplifier, V407, operates as a grounded screen amplifier. Actually the screen is slightly above r-f ground to provide screen neutralization of the tube. Neutralization is accomplished by balancing the capacitance bridge formed by the plate to control grid capacitance, the plate to screen grid capacitance, and the screen to control grid capacitance. This is accomplished by adding a small variable inductance, L406, to form an r-f voltage from the screen to controlgrid 180 degrees out of phase with the plate to control grid r-f. This additional impedance balances the bridge to neutralize the tube for a fixed operating frequency.

Plate voltage is obtained from a conventional 3-phase mercury vapor power supply, which, when connected in series with the screen power supply, forms the plate voltage. The power supply transformer is connected in a delta primary and a wye secondary configuration. Six mercury vapor rectifier tubes, V401 through V406, are arranged in a 3 -phase full wave bridge rectifier circuit. A P.A. PLATE CURRENT meter, M402, is placed in the plate supply side of ground to indicate only plate current. As the screen supply is in series with the plate supply, the P.A. PLATE VOLTAGE meter, M403, is placed across both the screen and plate supplies to indicate the filament to plate voltage. The plate supply output is approximately 5700 volts d-c at 1.6 amp . When the plate supply voltage is added to the screen supply voltage, the total, 6350 volts, equals the total plate voltage.

Screen voltage is obtained from a conventional 3-phase semiconductor power supply. Resistors R443 through R466 are placed in parallel with the diodes to equalize the diode currents. A transient suppressor, R439 and C411, reduces transients formed when the power supply is turned on or off. The screen supply output is approximately 650 volts at 1.8 amp .

The control grid bias supply is a single phase, semiconductor power supply. A resistor and capacitor placed in parallel with each diode equalizes each diode. A transient suppressor, C412 and R430, reduces transients when power is turned on or off. The positive end of the grid bias supply is connected to the fixed side of POWEROUTPUT ADJUST potentiometer, R405. This permits R405 to control both the bias voltage and screen voltage when R405 is adjusted. The negative end of the bias supply is tied into the control grid through a meter shunt resistor and a filtering network. The control grid bias supply output is approximately 300 volts at 300 ma . Bias voltage is made adjustable by R417.

The filament of power amplifier V 407 is below d-c ground because of the grounded screen configuration. The filament to ground potential will then be the screen voltage.

Metering circuits are provided to measure the filament voltage, control grid bias voltage, control grid current, and screen current. In addition, two external metering positions are available for connection to the multimeter. The customer may employ these extra meter positions for any use that he may desire.

The filament voltage metering circuit employs a full wave bridge rectifier to change the a-c filament voltage to d-c. The bridge output is filtered by R418 and C410, and the complete circuit is calibrated by R415. The grid bias voltage circuit is a voltage measuring device which measures the control grid bias voltage directly through the use of meter multiplier resistor R416. Grid current is metered by shunt resistor R428. Screen current is measured by placing the MULTIMETER across shunt resistors R413 and R414. As the screen is grounded, only screen current will flow through these resistors. External meter readings may be made by connecting to the proper terminals on TB405 (not shown on simplified schematic, see figure $5-1$ ). The MULTIMETER is shunted with capacitor C450, to prevent any stray r-f from damaging the meter movement.

### 2.2 CONTROL FUNCTIONS.

The following paragraphs describe the functions of all controls in E830-1 5-Kw FM Power Amplifier. Refer to figure 2-4 for control locations.

The controls located on the front panel directly under the meters include the FILAMENT ON, FILAMENT OFF, PLATE OFF, and PLATE ON controls. The green "filament on" indicator and red "plate on" indicator are placed in line with the above mentioned controls. The FILAMENT ON switch, S112, energizes the filaments, the blowers, the bias supply, and will turn on the driver filament and the exciter if connected. The FILAMENT OFF switch, S111, de-energizes all transmitter circuits. The PLATE ON switch, S113, energizes the plate and screen power supplies and the driver plate supply. The PLATE OFF switch, S114, removes plate and screen voltage. The green filament indicator light, DS401, lights when the FILAMENT ON switch is pressed and the PA blower has activated the PA blower interlock. DS401 indicates that voltage is available to the filament control contactor. The filament control contactor starts the cabinet blower, the bias supply, and supplies the necessary voltage to the E830-1 filaments. The red "plate-on'" indicator light, DS402, indicates plate voltage has been applied to the power amplifier.

The following controls are located directly under the right front door on the power amplifier panel. The POWER OUTPUT ADJUST potentiometer, R405,
adjusts the power amplifier screen potential thus changing the output power. The FILAMENT VOLTAGE ADJUST potentiometer, R404, adjusts the filament transformer input voltage thus changing the filament voltage of V407. The WATTMETER switch, S406, connects the R.F. WATTMETER to either the reflected power or forward power section of the directional coupler. The WATTMETER switch is normally left in the FORWARD position. The MULTIMETER switch, S405, selects either filament voltage, bias voltage, control grid current, or screen current. In addition, two external positions are available for use by the customer. Table 3-1 lists the MULTIMETER switch positions and typical indications for each of the four meter positions.

The following controls are located directly behind the left front door on the power amplifier cavity. The PLATE TUNING control, C445, tunes the plate cavity to resonance and is set near the minimum indication on the P.A. PLATE CURRENT meter, M402. At this point, the power output should be at the peak as indicated on the R.F. WATTMETER, M404. The OUTPUT COUPLING control, C444, adjusts the coupling of the load to the plate cavity. The GRID TUNING control, C436, tunes the grid cavity.

The following controls are located on the power panel directly behind the bottom front panel of the power amplifier cabinet. The PLATE circuit breaker, CB402, is a protective device which monitors the plate and screen supply transformer primary currents. The circuit breaker will open when the current exceeds 30 amp. The CONTROL circuit breaker, CB401, monitors the total control circuit current. This current includes the power fed to the driver and exciter. The circuit breaker will open when the control circuit current exceeds 8 amp . The 5 -amp CONTROL CIRCUIT fuse, F401, protects the control circuits from overloads. The $1 / 2$ amp BIAS VOLTAGE fuse, F402, protects the bias supply from overloads.

The following adjustments are located directly below the P.A. PLATE CURRENT and R.F. WATTMETER indicators when the upper switch and meter panel is raised. The left potentiometer, located behind the panel, is the filament calibrating potentiometer, R415. This potentiometer is approximately 600 volts below ground potential and should only be adjusted with an insulated screwdriver. The filament metering circuit calibrating adjustment is also set at the factory and will not normally require adjustment. The center potentiometer is the overload adjusting potentiometer, R408. The overload adjustment is normally set for a plate current of 1.8 amp . The right potentiometer is the wattmeter adjusting resistor, R419. This potentiometer is normally set at the factory and should not require adjustment. The control grid bias adjustable resistor, R417, is located in the left bottom rear of the power amplifier cavity. The resistor is normally set for approximately 150 to 200 ma of plate current with no drive to the power amplifier.


Figure 2-4. E830-1 Control and Adjustment Locations

# SECTION III MAINTENANCE 

### 3.1 GENERAL.

This section contains information concerning the maintenance of E830-1 5-Kw FM Power Amplifier.

## WARNING

Voltages present in this equipmentare dangerous to life. Observe safety precautions when performing any maintenance. Do not reach inside the E830-1 cabinet when high voltage is applied. Do not depend entirely ondoor interlocks. Always shut down the E830-1 before doing any work inside the E830-1 cabinet. Immediately upon opening the rear doors, short out all high-voltage points using the shorting stick located inside the left rear door.

### 3.2 PREVENTIVE MAINTENANCE.

Most service interruptions in equipment of this type are caused by dirt and corrosion. Corrosion is accelerated by the presence of moisture and dust. Dust should be removed periodically with a soft brush or a dry, oil-free air jet. Remove dust as often as a perceptible quantity accumulates at any point in the power amplifier.

When the E830-1 is operated near salt water or in other corrosive atmospheres, inspect and clean interlock switches, cable connectors, tube prongs, and other metal parts more frequently to keep the equipment in top operating condition.

### 3.2.1 AIR FILTER CLEANING.

At least once each month, or more often if needed, clean the air filter according to the following procedure:
a. Remove the air filter from the E830-1 cabinet by loosening the two thumb screws located above the air filter. Slide the air filter to the extreme right, and pull the left side of the air filter out as soon as the filter clears the panel. Slide the air filter to the left and remove.
b. Mark with an arrow the direction of airflow.
c. Wash by passing a fine spray of hot water through the filter in the direction opposite that of the airflow. Gently shake the water out of the filter.
d. Dip the filter in a water-soluble oil, such as Filter-kote " $\mathrm{M}^{\prime \prime}$ ' available from Collins Radio Company

Service Parts Department, Cedar Rapids, Iowa (Collins part number 005-0609-00).
e. Remove the filter from the oil; lay the filter face down until oil ceases to drip from the filter.
f. Replace the filter into the lower rear panel with the airflow arrow (marked when the filter was removed) pointing in the direction of the airflow. Tighten the two thumb screws.
g. Repiacement filters are Collins part number 009-1069-00.

### 3.2.2 PA TUBE CLEANING.

The power amplifier tube depends upon a stream of air passing through the fins to cool the anode. When these fins become dirty, the airflow is reduced, and the tube life is shortened. The radiator fins should be cleaned as follows:
a. Remove the r-f amplifier tube as described in paragraph 3.2.2.1.

## CAUTION

Special care must be used in removing or installing the power amplifier tube.
b. Direct a low-pressure ( 50 psi ) air stream through the fins in the direction opposite to the normal airflow until all dust is removed.
c. Replace the r-f amplifier tube as described in paragraph 3.2.2.1.

### 3.2.2.1 PA TUBE REMOVAL

## WARNING

Voltages present within the plate cavity are dangerous to life. Shut down the E830-1 before doing any work inside the cavity.

The power tube may be removed as follows:
a. Open the power amplifier cavity.
b. Grasp the center transmission line, and lift straight up until the transmission line clears the tube. (In some instances where the operating frequency is low, the slider clamp surrounding the center transmission line may have to be loosened. Before loosening the clamp, mark the plate slider position on the cavity sides for reference when replacing the slider.)
c. Grasp the handles located on the tube, and lift the tube from the tube socket. When the tube clears the tube socket chimney, the tube may be removed.
d. Tube replacement is the reverse of removal. After tube replacement, check the plate slider for proper distance from the deck plate for the station frequency. See System Instructions, Frequency Change for this approximate distance. Be sure the tube is seated firmly in the tube socket before replacing the center transmission line.

### 3.2.3 INSPECTION.

Once each week check and clean the three interlock switches at the rear of the E830-1 cabinet to be sure they are in good working order. Once each month check all connections in the E830-1. Tighten any nuts, bolts, or screws that may be loose. Check cable connections to see that they are clean and mechanically secure. Check moving parts, such as tuning controls, for excessive wear. Check the plate cavity slider for oxidation around ground springs.

### 3.2.4 LUBRICATION

The cabinet fan and PA blower have bearings that are lubricated for the life of the equipment. No lubrication of the $\mathrm{E} 830-1$ is required.

### 3.2.5 TUBE MAINTENANCE.

The power amplifier, V201, should be inspected (tube in place) once each week to ensure that an accumulation of dust does not build up on the radiator fins. If dust is present, clean as described in paragraph 3.2.2. When tuning the E830-1, care should be taken not to exceed the maximum plate current shown in table 3-1.

### 3.3 TROUBLE SHOOTING.

The most common cause of trouble will probably be traced to tube failure. If a tube is suspected of failure, replace it with a tube of known quality, and note any change in performance. A small loss in emission of V201 can be compensated for by a change in the setting of the POWER OUTPUT ADJUST potentiometer.

Spare, preaged, mercury-vapor rectifier tubes should be available for immediate replacement. To ready these tubes for emergency use, place them in the power amplifier during off-the-air hours, and run them for approximately twenty minutes with only the filaments lighted. This will remove the mercury coating from the tube elements. Then carefully remove the tubes from the power amplifier, and store them in an upright position where they will not be inverted or agitated. When these preaged tubes are placed in the E830-1, handle them carefully to avoid the twentyminute warmup period that will be required if mercury comes in contact with the tube elements. Never apply plate voltage to mercury-vapor rectifier tubes that have not been aged long enough to remove all mercury from the tube elements.

Four meters are located on the E830-1 front panel to assist in locating any trouble which may occur. Table 3-1 contains typical meter indications. These average indications are obtained from several production power amplifiers. The indications of some E830-1 mayvary slightly outside the given limits without affecting power amplifier performance. A list of panel meter indications for each individual power amplifier should be taken when the E830-1 is operating properly in its particular installation. Any abnormal deviation from these values will be apparent during a check of meter indications.

TABLE 3-1. TYPICAL METER INDICATIONS

| METER | METER SWITCH POSITION | INDICATION |
| :---: | :---: | :---: |
| MULTIMETER | Filament | 7 F 5 volts $7.0-7.3$ |
| MULTIMETER | Bias V 400 V FS |  |
| MULTIMETER | Grid I 40 MA FS | 5-15 ma |
| MULTIMETER | Screen I 80 MA FS | $5-40 \mathrm{ma} \quad 25-30$ |
| PA PLATE VOLTAGE |  | 6100 to 6300 volts |
| PA PLATE CURRENT |  | Approx. 1.25 A |
| R.F. WATTMETER | Forward | 5 Kw |
| R.F. WATTMETER | Reflected | Less than 500 watts |

### 3.4 CABLE CHART.

Table 3-2 contains from-to information for cables installed in E830-1 5-Kw FM Power Amplifier. The table is useful in locating point-to-point wiring within the E830-1 cabinet. The from column is listed in alphabetical and numerical order. To find a particular
wire, establish the point on the E830-1 from which wire tracing is to be initiated. Find this point in the from column of table 3-2, and the to column will give the location of the other end of that particular wire. The wire code column, located inside the rear cover of this instruction book, gives the type and color of wire used in each case.

TABLE 3-2. CABINET FROM-TO INFORMATION

| WIRE CODE | FROM | TO | WIRE CODE | FROM | то |
| :---: | :---: | :---: | :---: | :---: | :---: |
| RD90 | B401-1 | K401-2 | RC 9 | E401 | K403-6 |
| RD92 | B401-2 | K401-4 | RC9 | E401 | K404-7 |
| RD95 | B401-3 | K401-6 | RC9 | E401 | K405-2 |
| v C905 | C401-1 | L401-2 | RC9 | E401 | K406-10 |
| VC902 | C401-2 | TB406-12 | RC925 | E401 | M404-2 (-) |
| RB91 | C401-2 | T402-4 | VE905 | E401 | R413-1 |
| V C92 | C402-1 | R417-1 | RC9 | E401 | S112-1 |
| VC906 | C402-1 | L402-2 | VB9 | E401 | S405-15 |
| VC91 | C402-2 | R431-2 | RC9 | E401 | T401-9 |
| VE92 | C403-1 | R405-1 | RE9 | E402 | E401 |
| VE902 | C403-1 | S413-B | RE9 | E402 | TB401-4 |
| VE92 | C403-1 | R412-1 | RE9 | E402 | TB403-10 |
| VE92 | C403-1 | S412-B | VE93 | E407 | C408-2 |
| RE92 | C403-2 | R412-2 | VE92 | E407 | M402-1 (-) |
| VE92 | C404-1 | L404-2 | LE9 | E407 | V404-1 |
| VC95 | C404-1 | R431-2 | LE9 | E407 | V406-1 |
| LE9 | C408-1 | R410-1 | LE9 | E407 | V402-1 |
| VE93 | C408-2 | E407 | VB90 | E409 | R417-3 |
| VE92 | C408-2 | R411-2 | VB90 | E409 | S405-5 |
| RC93 | C427 | TB407-7 | VC93 | E410 | R418 |
| VB93 | C428 | TB407-2 | VC92 | E410 | TB407-3 |
| RC90 | C429 | R404-2 | VB93 | E411 | TB407-1 |
| RC92 | C430 | K402-4 | VB92 | E418 | R431-2 |
| VE2 | C431 | R405-2 | RC95 | J401-1 | K402-6 |
| VB91 | C432 | S405-13 | RC90 | J401-2 | K402-8 |
| LE9 | C447-1 | L403-2 | RD93 | K401-1 | CB401-6 |
| VG0 | CB401-3 | TB401-1 | RD90 | K401-2 | B401-1 |
| VG2 | CB401-2 | TB401-2 | RD92 | K401-3 | CB401-5 |
| VG95 | CB401-1 | TB401-3 | RD92 | K401-4 | B401-2 |
| RD95 | CB401-4 | K401-5 | RD95 | K401-5 | CB401-4 |
| RD95 | CB401-4 | T401-1 | RD95 | K401-6 | B401-3 |
| RC5 | CB401-4 | T402-2 | RC 93 | K401-7 | K403-4 |
| RE5 | CB401-4 | TB402-2 | RC9 | K401-8 | E401 |
| RD92 | CB401-5 | K402-3 | RC935 | K401-9 | TB403-8 |
| RE2 | CB401-5 | TB402-1 | RC935 | K401-9 | K402-9 |
| RD92 | CB401-5 | K401-3 | RC93 | K401-10 | S112-2 |
| RD93 | CB401-6 | K4015 | RC93 | K401-10 | TB403-6 |
| VG0 | CB402-1 | TB401-1 | RD93 | K402-1 | CB401-6 |
| VG2 | CB402-2 | TB401-2 | RC902 | K402-2 | C404-1 |
| VG4 | CB402-3 | TB401-3 | RD92 | K402-3 | CB401-5 |
| VGO | CB402-4 | K405-3 | RC92 | K402-4 | C430 |
| VG2 | CB402-5 | K405-5 | RC923 | K402-5 | T401-1 |
| VG4 | CB402-6 | K405-7 | RC95 | K402-6 | J401-1 |
| DAS0 | DC401-1 | S406-1 | RC95 | K402-6 | T406-2 |
| DAS9 | DC401-2 | S406-2 | RD90 | K402-7 | T401-5 |
| RC936 | DS401-1 | K402-9 | RC90 | K402-8 | XF402-1 |
| RE9 | E401 | E402 | RC90 | K402-8 | T406-1 |
| RC9 | E401 | K401-8 | RC90 | K402-8 | J401-2 |

TABLE 3-2. CABINET FROM-TO INFORMATION (CONT)

| WIRE CODE | FROM | TO | WIRE CODE | FROM | TO |
| :---: | :---: | :---: | :---: | :---: | :---: |
| RC935 | K402-9 | K401-9 | RC91 | K4U6-12 | XDS402-1 |
| RC936 | K402-9 | DS401-1 | RC91 | K406-12 | K403-13 |
| RC91 | K402-9 | K403-13 | VB95 | K407-1 | R408-2 |
| RC935 | K402-9 | TB404-5 | RC92 | K407-3 | TB403-7 |
| RC936 | K402-10 | S407-1 | RC923 | K407-4 | S408-1 |
| RC902 | K402-10 | TB407-12 | V C903 | L401-1 | TB406-5 |
| RD96 | K403-1 | XF401-1 | V C905 | L401-2 | C401-1 |
| RC6 | K403-1 | K406-7 | V C906 | L402-2 | C402-1 |
| RC6 | K403-3 | K406-8 | LE9 | L403-1 | T406-19 |
| RC91 | K403-3 | S412-2 | LE9 | L403-2 | C447-1 |
| RC93 | K403-4 | K401-7 | LE9 | L403-2 | TB407-11 |
| RC93 | K403-4 | S407-2 | LE9 | L403-2 | S413-A |
| RC9 | K403-6 | E401 | LE9 | L403-2 | R480-1 |
| RC923 | K403-7 | K407-3 | VE93 | L404-1 | TB412-9 |
| RC923 | K403-9 | TB403-2 | VE92 | L404-2 | C404-1 |
| RC925 | K403-10 | TB404-3 | VB92 | M401-1 (+) | S405-11 |
| RC926 | K403-12 | K405-11 | VB90 | M401-2 (-) | S405-2 |
| RC91 | K403-13 | K402-9 | VE92 | M402-1 (-) | E407 |
| RC91 | K403-13 | K406-12 | RE92 | M402-2 (+) | R407-1 |
| RC905 | K403-14 | S403-2 | VC905 | M403-1 (-) | TB407-15 |
| RC926 | K403-14 | TB404-4 | VC906 | M403-2 (+) | TB407-7 |
| RC7 | K404-2 | TB404-10 | RC91 | M404-1 (+) | R419-1 |
| RC6 | K404-3 | K406-8 | RC925 | M404-2 (-) | E401 |
| RC6 | K404-4 | K405-1 | VG0 | R401-1 | K406-1 |
| RC4 | K404-6 | TB403-5 | VGO | R401-2 | K406-2 |
| RC9 | K404-7 | E401 | VG2 | R402-1 | K406-3 |
| RC6 | K405-1 | K404-4 | VG2 | R402-2 | K406-4 |
| RC9 | K405-2 | E401 | VG4 | R403-1 | K406-5 |
| VG0 | K405-3 | CB402-4 | VF4 | R403-2 | K406-6 |
| VG0 | K405-4 | K406-1 | RC902 | R404-1 | K402-2 |
| VG2 | K405-5 | K402-5 | RC90 | R404-2 | C429 |
| VG2 | K405-6 | K406-3 | VE92 | R405-1 | C403-1 |
| VG4 | K405-7 | CB402-6 | VE2 | R405-2 | C431 |
| VG4 | K405-8 | K406-5 | VC92 | R405-2 | TB407-14 |
| RC926 | K405-11 | K403-12 | RE92 | R407-1 | M402-2 (+) |
| RC91 | K405-12 | K406-11 | VC95 | R407-1 | TB405-4 |
| VG0 | K406-1 | K405-4 | VC93 | R407-2 | TB405-3 |
| VGO | K406-1 | R401-1 | VB91 | R407-2 | S405-6 |
| VG0 | K406-2 | R401-2 | VB93 | R408-1 | R409-2 |
| VG0 | K406-2 | T403-1 | VB95 | R408-2 | K407-1 |
| RE90 | K406-2 | S411-2 | RE92 | R409-1 | TB412-4 |
| VG2 | K406-3 | K405-6 | VB93 | R409-2 | R408-1 |
| VG2 | K406-3 | R402-1 | LE9 | R410-1 | C408-1 |
| VG2 | K406-4 | R402-2 | VE92 | R411-2 | C408-2 |
| VG2 | K406-4 | T403-8 | VE92 | R412-1 | C403-1 |
| RE2 | K406-4 | S411-4 | RE92 | R412-2 | C403-2 |
| VG4 | K406-5 | K405-8 | VE905 | R413-1 | E401 |
| VG4 | K406-5 | R403-1 | V C92 | R417-1 | C402-1 |
| VG4 | K406-6 | T403-15 | VB90 | R17-3 | E409 |
| VG4 | K406-6 | R403-2 | VC93 | R418 | E410 |
| RE5 | K406-6 | S411-6 | RC91 | R419-1 | M404-1 (+) |
| RC6 | K406-7 | K403-1 | DAS90 | R419-2 | S406-3 |
| RC6 | K406-8 | K403-3 | VC95 | R431-2 | C404-1 |
| RC6 | K406-8 | K404-3 | VC91 | R431-2 | C402-2 |
| RC902 | K406-9 | TB407-5 | VB92 | R431-2 | E418 |
| RC 9 | K406-10 | E401 | LE9 | R480-1 | L403-2 |
| RC91 | K406-11 | K405-12 | RCl | S111-1 | TB404-7 |

TABLE 3-2. CABINET FROM-TO INFORMATION (CONT)

| WIRE CODE | FROM | TO | WIRE CODE | FROM | TO |
| :---: | :---: | :---: | :---: | :---: | :---: |
| RC1 | S111-2 | XDS403-1 | VG2 | T403-8 | K406-4 |
| RC1 | S111-2 | S412-1 | VG4 | T403-15 | K406-6 |
| RC9 | S112-1 | E401 | LE9 | T403-22 | T406-10 |
| RC93 | S112-2 | K401-10 | LE9 | T403-23 | T406-4 |
| RC925 | S113-1 | TB404-3 | LE9 | T403-24 | T406-7 |
| RC925 | S113-1 | S114-1 | RE90 | T404-1 | S411-1 |
| RC925 | S114-1 | S113-1 | RE2 | T404-8 | S411-3 |
| RC923 | S114-2 | S410-2 | RE5 | T404-15 | S411-5 |
| RC925 | S114-2 | TB404-3 | VD91 | T404-22 | TB412-1 |
| RC905 | S403-2 | K403-14 | VD95 | T404-23 | TB412-2 |
| VB90 | S405-2 | M401-2 (-) | VD96 | T404-24 | TB412-3 |
| VB91 | S405-6 | R407-2 | RC90 | T406-1 | K402-8 |
| VB95 | S405-7 | TB405-5 | RC95 | T406-2 | K402-6 |
| VB902 | S405-8 | TB405-7 | LE9 | T406-3 | XV402-F1 |
| VB90 | S405-5 | E409 | LE9 | T406-4 | V401-1 |
| VB92 | S405-11 | M401-1 (+) | LE9 | T406-4 | T403-23 |
| VB91 | S405-13 | C432 | LE9 | T406-5 | XV402-F2 |
| VB9 | S405-15 | E401 | LE9 | T406-6 | XV404-F1 |
| VB903 | S405-16 | TB405-6 | LE9 | T406-7 | V403-1 |
| VB96 | S405-17 | TB405-8 | LE9 | T406-7 | T403-24 |
| - DASO | S406-1 | DC401-1 | LE9 | T406-8 | XV404-F2 |
| DAS9 | S406-2 | DC401-2 | LE9 | T406-9 | XV406-F1 |
| DAS90 | S406-3 | R419-2 | LE9 | T406-10 | V405-1 |
| RC936 | S407-1 | K402-10 | LE9 | T406-10 | T403-22 |
| RC93 | S407-2 | K403-4 | LE9 | T406-11 | XV406-F2 |
| RC923 | S408-1 | K407-4 | LE9 | T406-12 | XV401-F1 |
| RC92 | S408-2 | S409-1 | LE9 | T406-14 | XV401-F2 |
| RC92 | S409-1 | S408-2 | LE9 | T406-15 | XV403-F1 |
| RC92 | S409-2 | S410-1 | LE9 | T406-17 | XV403-F2 |
| RC92 | S410-1 | S409-2 | LE9 | T406-18 | XV405-F1 |
| RC923 | S410-2 | S414-2 | LE9 | T406-19 | L403-1 |
| RE90 | S411-1 | T404-1 | LE9 | T406-20 | XV405-F2 |
| RE90 | S411-2 | K406-2 | VG0 | TB401-1 | CB402-1 |
| RE2 | S411-3 | T404-8 | VG0 | TB401-1 | CB401-3 |
| RE2 | S411-4 | K406-4 | VG2 | TB401-2 | CB402-2 |
| RE5 | S411-5 | T404-15 | VG2 | TB401-2 | CB401-2 |
| RE5 | S411-6 | K406-6 | VG4 | TB401-3 | CB402-3 |
| VE92 | S412-B | C403-1 | VG95 | TB401-3 | CB401-1 |
| RC1 | S412-1 | S112-2 | RE9 | TB401-4 | TB402-3 |
| RC91 | S412-2 | TB403-3 | RE9 | TB401-4 | E402 |
| LE9 | S413-A | L403-2 | RE2 | TB402-1 | CB401-5 |
| LE9 | S413-A | S414A | RE5 | TB402-2 | CB401-4 |
| VE903 | S413-B | S414-B | RE9 | TB402-3 | TB401-4 |
| VE903 | S414-B | S413-B | RD90 | TB403-1 | TB404-9 |
| LE9 | S414-A | S413-A | RC923 | TB403-2 | K403-9 |
| RC923 | T401-1 | K402-5 | RC91 | TB403-3 | S412-2 |
| RC95 | T401-1 | CB401-4 | RC4 | TB403-5 | K404-6 |
| RD93 | T401-5 | CB401-6 | RC93 | TB403-6 | TB404-6 |
| RD90 | T401-5 | K402-7 | RC93 | TB403-6 | K401-10 |
| RC6 | T401-8 | X401-2 | RC92 | TB403-7 | K407-3 |
| RC9 | T401-9 | E401 | RC935 | TB403-8 | K401-9 |
| RC90 | T402-1 | XF402-2 | RE9 | TB403-10 | E402 |
| RC5 | T402-2 | CB401-4 | RC925 | TB404-3 | K403-10 |
| RB90 | T402-3 | TB406-1 | RC925 | TB404-3 | S113-1 |
| RB91 | T402-4 | C401-2 | RC926 | TB404-4 | K403-14 |
| RB92 | T402-5 | TB406-7 | RC935 | TB404-5 | K402-9 |
| VG0 | T403-1 | K406-2 | RC93 | TB404-6 | TB403-6 |

TABLE 3-2. CABINET FROM-TO INFORMATION (CONT)



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TD-539
E830-1 5-Kw FM Power Amplifier

| ITEM | DESCRIPTION | COLLINS <br> paft number |
| :---: | :---: | :---: |
| Ratis | RESISTOR. FIXED. COMPOSITION: 3300 ohms, $: 20^{\circ}$. 30 v dc | 712-4224-00 |
| 8469 | RESISTOR, FIXED, COMPOSITION: 470 ohmis, $-10^{c i}$. 2 w | 745-5638-00 |
| R170 | RESISTOR, FIXED: 22 ohms, $\mathrm{r}^{20} 0^{\circ} \%, 18 \mathrm{w}$ | 712-0002-00 |
| R471 | RESISTOR, FLXED: 50 ohms ${ }^{\text {c }}$ - $20^{\circ} \mathrm{c}$, 60 w | 712-0070-00 |
| R. 472 | RESISTOR, FIXED: same as R470 | 712-0002-00 |
| thru R+74 |  |  |
| Ri75 | RESISTOR, FIXED, WIREWOUND. 100 ohms, : 5 \%. 5 w; Dale Proxlucts part no. RSm-5 | 747-5440-00 |
| 17476 | RESISTOR. FIXED, WIREWOUND: same as R475 | 747-5440-00 |
| R477 | NOT USED |  |
| R479 |  |  |
| R480 | RESISTOR. FDXED, WIRENOUND: 20 ohms. $5 \% .55 \mathrm{w}$ | 747-2713-00 |
| R481 | RESISTOR FIXED. FILM: 2870 ohms, $1 \mathrm{I}_{\mathrm{r}}^{\mathrm{C}}, 1 \mathrm{w}$ | 705-3272-00 |
| R482 | RESISTOR, FIXED, COMPOSITION: 10.000 ohnis. $10^{\circ \prime 2} 0,2 \mathrm{w}$ | 745-5691-00 |
| R483 | RESISTOR, FIXED, FILM. same as R481 | 705-3272-00 |
| S401 | SWITCH, PUSH: normally open and closed. bakelite: Arrow-Hart \& Hepeman part no. B-2 | 260-2020-00 |
| S402 | SWITCH, PUSH. same as Stol | 260-2020-00 |
| $\begin{aligned} & \text { thru } \\ & \text { S404 } \end{aligned}$ |  |  |
| S405 | SWITCLI. ROTARY: 6 positons. 2 moving. 12 fixed contacts, 115 vac. $7-1 / 2 \mathrm{amps}, 2$ poles, 5 throw | 259-0113-00 |
| S406 | SWITCH, TOGGLE: spdi; 125 v or 250 v dc or 250 v dc: 5 ample: Micro Swith parino. 6ATII-T | 200-1509-00 |
| S407 | SWITCH. AIRFLOW int ERLOCK: spdi: 5 amps at 250 rac; metal case | 266-8307-00 |
| S408A | CONTACT ASSEMBLY, ELECTRICAL- $5 / 8 \mathrm{in}$. by $11 / 16 \mathrm{in}$. by $1-7 / 8 \mathrm{in}$. uverall | 260-40 40-00 |
| S408B | CONTACT ASSEMBI I : door interlock switeh, female: Neptune Electronles part no. F7460330G-4 | 260-4050-00 |
| S409A | CONTACT ASSEMBLY. ELECTRICAL. same as S408A | 260-40-40-00 |
| S409 ${ }^{\text {B }}$ | CONTACT ASSEMbly same as St08B | 260-4050-00 |
| S410A | CONTACT ASSEMBLY, ELECTRICAL: Same as S408A | 260-4040-00 |
| S410B | CONTACT ASSEMDLY same as S408b | 260-4050-00 |
| S411 | SWITCH TOGGLE: 3 pst: lug type ter minals, 10 amps, $250 \mathrm{vac} ; 15$ amps at 125 v ac: CutlerHammer, Inc. Part no. 7611-K2 | 260-2274-00 |
| S412 | ELEMENT SECTION. ANTENNA NO. I: chronate dip: 9 in, ls overall | 549-2245-003 |
| R412A | SWITCH. SENSITIVE: sıdt: 6 amps 28.5 v dc; Micro Switch part no. Y3-1 | 260-0025-00 |
| R412BC | Pio S4 12 |  |
| T401 | TRANSFORMER. POWER, STEP-DOWN. 1 pri wanding. 250 v , tapped at $200 \mathrm{v}, 210 \mathrm{v}, 220 \mathrm{v}, 230$ vat 100 va. 240 v. 50 to $60 \mathrm{cps} ; 120$ rat 4.25 amp secougdary, open Irane; 5-1/8 in. by 5-1/8 in. by 5-3,8 in. : Electro Enfr. Works part no. El2322 | 662-0043-00 |
| T402 | TRANSFORMER. POWER, STEP-UP: $230 v, 60$ eps pri. 300 v de at 100 ma. center tapped secondary; metal case; 2.750 in . Jy 2.937 in . by $3.812 \mathrm{in}$. . Chicago Std. Transformer part no. 31260 | 662-0049-00 |
| T 403 | TRANSFORMER. POWER STEP-UP: $230 \% 3$ phase. 50 to 60 cps , tapped at 200. 210. 220. 240 , 250 v inputs; 5800 v de at 1.6 amps secondary; open frame: 10 in . by $13-1 / 4 \mathrm{in}$. by 16 mm -; Electro Engr. Works part no. El2211 | 664-0007-00 |
| T404 | TRANSFORMER, POWER, STEP-UP: $230 v, 3$ phase. 50 to 60 cps , tapped at 200. 210, 220, 240 , $250 v$ inputs: 650 r at 1.8 amp secondary: open frame: 6-1,2 in. by $8-3 / 4 \mathrm{in}$. by 9 in ., Electro Engincering Works part no. El2371 | 664-0004-00 |


| ITEM | DESCRIPTION | COLLINS <br> PAET NUMDER |
| :---: | :---: | :---: |
| T405 | TRANSFORMER. POWER, STEP-DONN: 1 primary winding 245 v. tapped at $235 \mathrm{v}, 225$ v, $215 \mathrm{v}, 205 \mathrm{v}, 501060 \mathrm{cps}, 7.8 \mathrm{v}$ at 75 amp , center tapped, secondary: open frame: 5-1/2 in. by 6-1/4 in. Dy $\mathbf{6 - 5 / 8} \mathrm{in}$. : Electro Engr. Works part no. E9186-A | 662-0213-00 |
| T406 | TRANSFORMER. POWER, STEP-DOWN: 1 primary winding, 230 v . 50 to 60 cps , 4 secondaries each w/ 5 v, center tapped. three at 7.5 amps, one at 22.5 amps: open frame: 4-3/4 in. by 8-5, 8 in . by $9-1 / 8 \mathrm{in}$. : Electro Engr. Works part no. E10240 | 662-0273-00 |
| TB401 | TERMINAL BOARD: ghenolic: 50 amps GOO vac or de; 3-1/4 in. $\mathrm{l}_{6}$ by 1-5/16 in. h; Square D Co. part no. TB-4 | 306-0778-00 |
| TB*02 | TERMINAL STRIP: latrier type: black phenolic connector strip: 4 terminals: $1-5 / 16$ in. by 37/32 in. !f: Howarll B. Jones part no. 4-142 | 367-5040-00 |
| TB403 | TERMINAL BOARD: phenolic: $13 / 32 \mathrm{~m}$. by 7/B in. by 5-11/04 m. . inels 12 screw type terminais: tarrier type | 367-0518-00 |
| TB404 | TERMINAL DOARD: same as TB403 | 367-0518-002 |
| TB405 | TERMINAL BOARD. black phemolic: $\mathbf{1 / 2} \mathbf{i n}$. by 1-1,8 in. by 7-3/4 in.: 16 lerminals incl | 367-4160-00 |
| TB406 | TERMINAL BOARD: type PBE-P plastic: $1 / 16$ in. thk, 3-1/2 in. w by 6-5/8 in. 15 | 549-2234-003 |
| TB407 | TERMINA L EOARD: type GEE laminated glass cloth; $1 / 8 \mathrm{in}$. 1 lk , 4 -1/4 in. w by 12 in . 1g | 549-2264-003 |
| TB408 | TERMINAL BOARD: 4 terminals brass nickel plated screws and eyclets; $7 / 8 \mathrm{in}$. why $2-5 / 32 \mathrm{in}$. $1_{\mathrm{f}}$, | 367-0002-00 |
| T8409 | TERMINAL BOARD: same as TB408 | 367-0002-00 |
| TB410 | NOT USED |  |
| TB411 | NOT USED |  |
| TB412 | RECTIFIER ASSEMBLY: $1-1 / 2 \mathrm{in}$. by $3-3 / 4 \mathrm{in}$. by $18-1 / 4 \mathrm{in}$. | 549-2274-00 |
| V401 | ELECTRON TUBE: glass envelope, rectifier: General Electric 872A; 872 | 256-0037-00 |
| V402 thru | ELECTRON TUBE: same as V401 | 256-0037-00 |
| V406 |  |  |
| V407 | ELECTRON TUBE: tetrode: 1ype 4Cx5000A | 256-0122-00 |
| $\underset{401 \mathrm{~A}}{\mathrm{XDS}}$ | LAMPHOLDER: pancl mounting for use will candelabra screw hase lamp | 262-0255-00 |
| $\underset{401 \mathrm{~B}}{\mathrm{XDS}}$ | LENS. INDICATOR LIGHT: glass, green, Iranslucent; [rosted back; chrome holder | 262-0258-00 |
| $\begin{gathered} \mathrm{XDS} \\ 402 \mathrm{~A} \end{gathered}$ | LAMPHOLDER: same as XDS 401 A | 262-0255-00 |
| $\begin{gathered} \times D S \\ 402 B \end{gathered}$ | LENS, indICATOR LIGHT: glass, red. translucent; Irosted back; chrome holder | 262-0259-00 |
| $\underset{403 \mathrm{~A}}{\mathrm{XDS}}$ | LAMPIOLDER: mountmg bracket for minature Latyonet tase pilot light bulbs | 262-1260-00 |
| $\underset{403 \mathrm{~B}}{\mathrm{XDS}}$ | LeNS: for use w/ miniature tayonet base type $\uparrow$ -3-1/4 lamp: amber, frosted back; Dialight Corp. part no. 857B-213 | 262-0922-00 |
| XF401 | FUSEHOLDER: w/ transparent knol for use with 3 AG luses; $0-20$ amps, $100-250 \mathrm{v}$ rating; Bussman Mfy. Co. part no. HKL-JRZ | 265-1040-00 |
| XF402 | FUSEHOLDER: same as XF401 | 265-1040-00 |
| XV401 | SOCKET, ELECTRON TUBE: jumbo 4 pin bayonet Lase tube socket: 20 amps ; E. F. Johnson Co. part no. 123-211-1 | 220-1460-00 |
| XV402 | SOCKET, Electron tube: same as XV401 | 220-1460-00 |
| XV407 | SOCRET | 220-1479-00 |



Figure 4-1. Cavity, Parts Location


Figure 4-2. Indicator Panel, Parts Location


Figure 4-3. Plate Power Supply, Parts Location


Figure 4-4. Control Panel. Parts Location


Figure 4-5. Filament Supply, Parts Location


Figure 4-G. Power Panel. Small Parts Location


Figure 4-7. Power Panel, Parts Location


# system instruction book 

This manual includes<br>SP-195 830F-1A 10-Kw FM Broadcast Transmitter<br>TD-536 A830-2 10 W Wide-Band FM Broadcast Exciter<br>TD-537 786M-1 Stereo Generator (optional)<br>TD-538 B830-1 250-Watt FM Power Amplifier<br>TD-580 F830-1 10-Kw FM Power Amplifier<br>©Collins Radio Company $196 ?$

# system instructions 



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unit instructions
TD No.Title
536 A830-2 10 W Wide-Band FM Broadcast Exciter538 B830-1 250-Watt FM Power Amplifier
580
F830-1 10-Kw Power Amplifier

### 1.1 Purpose of Instruction Book.

This instruction book is a guide for installing, adjusting, operating, and maintaining 830F-1A $10-\mathrm{Kw}$ FM Broadcast Transmitter.

### 1.2 Purpose of Equipment.

The 830F-1A 10-Kw FM Broadcast Transmitter is used for continuous monophonic or optional stereophonic FM broadcast service on a single frequency in the range from 88 to 108 megacycles with a maximum output power of 10,000 watts.

### 1.3 Description of Equipment.

### 1.3.1 PHYSICAL DESCRIPTION.

The 830F-1A 10-Kw FM Broadcast Transmitter, shown in figure 1-1, is contained in two inter connected cabsnets that, together, are 76 inches wide, 76 inches high, 27 inches bleep, and weigh approximately 1900 pounds. All transmitter operating controls are located behind the doors on the front of the cabinets. The filament and plate on-off controls and eight monitoring meters are located at the top front of the cabinets. The meters may be observed easily while operating the tuning controls. A monitoring meter is also provided with the


Figure 1-1. 830F-1A 10-Kw FM Broadcast Transmitter, Over-all View
wide-band exciter. The transmitter uses a maximum of 15 tubes ( $10-\mathrm{kw}$ power amplifier may use optional silicon diode rectifiers in the plate supply instead of mercury-vapor tubes) and 20 transistors, most of which are accessible from the front of the transmitter. The bottom front of the transmitter cabinets are removable to allow access to components on the bottom of the inside panels.

The large doors at the upper rear of the cabinets allow access to the upper part of the transmitter for servicing and maintenance. The lower rear half of the transmitter cabinets are covered by removable panels that contain ventilating fans and permanent air filters. Operating personnel are protected by both electrical and mechanical interlocks on the rear doors and panels. These interlocks ground the transmitter highvoltage circuits when the doors are opened or the panels are removed. The power amplifier plate-tuning resonator is located in an interlocked compartment at the front of the transmitter.

Inside the transmitter, heavy iron-core components are at the bottom of the cabinets. The exciter portion of the transmitter and the 250 -watt driver are contained in one cabinet. The other cabinet contains a 10,000 -watt power amplifier and harmonic filter.

Cooling air for the transmitter is drawn through permanent air filters at the rear of the cabinets by high-volume fans and exhausted through shielded openings in the tops of the cabinets. Individual blowers supply cooling air directly to the driver and power amplifier tubes.

Room is provided in the 250 -watt amplifier cabinet for mounting a stereo generator and SCA equipment if multiplex operation is desired.

### 1.3.2 ELECTRICAL DESCRIPTION.

The 830F-1A $10-\mathrm{Kw}$ FM Broadcast Transmitter is composed of three electrically-connected subunits: (1) a wide-band exciter that furnishes a 10 -watt $F M$ output to drive (2) a 250 -watt amplifier that, in turn,

TABLE 1-1
SUBASSEMBLY INSTRUCTION BOOKS

| PUBLICATION | INSTRUCTION BOOK <br> NUMBER |
| :--- | :---: |
| A830-2 10 W Wide-Band <br> FM Broadcast Exciter <br> B830-1 250-Watt FM <br> Power Amplifier <br> F830-1 10-Kw FM <br> Power Amplifier | TD-536 |

drives (3) a 10,000 -watt power amplifier. Instruction books covering the exciter and power amplifiers used in the transmitter are listed in table 1-1. These three books are supplied following section $V$ of this system instruction book. The unit instruction books contain detailed descriptions of the three transmitter subunits.

### 1.4. Equipment Supplied.

Table 1-2 lists equipment that is supplied as part of 830F-1A 10-Kw FM Broadcast Transmitter.

TABLE 1-2 EQUIPMENT SUPPLIED

| EQUIPMENT | COLLINS <br> PART NUMBER |
| :--- | :---: |
| A830-2 10 W Wide-Band FM <br> Broadcast Exciter <br> B830-1 250-Watt FM <br> Power Amplifier <br> F830-1 10-Kw FM Power <br> Antplifier <br> $5 / 10-K w$ Harmonic Filter | $549-2714-00$ |

### 1.5 Accessory Equipment.

Table 1-3 lists accessory equipment that is available for use with 830F-1A 10-Kw FM Broadcast Transmitter. Information on $786 \mathrm{M}-1$ Stereo Generator will be found in Unit Instructions, TD-537.

TABLE 1-3 ACCESSORY EQUIPMENT

| EQUIPMENT | COLLINS <br> PART NUMBER |
| :---: | :---: |
| $786 \mathrm{M}-1$ Stereo Generator | $522-2914-00$ |

### 1.6 Equipment Specifications.

### 1.6.1 MECHANICAL.

Weight . . . . . . . 1900 pounds maximum.
Size . 76 inches wide, 76 inches high, 27 inches deep.

Ventilation . . . . . . Two ventilating fans, two blowers.


| Carrier-frequency stability. | - Within $\pm 2000 \mathrm{cps}$ of specified carrier frequency over ambient temperature range from $+10^{\circ} \mathrm{C}\left(50^{\circ} \mathrm{F}\right)$ to $+45^{\circ} \mathrm{C}$ $\left(113^{\circ} \mathrm{F}\right)$ and line-voltage variations of $\pm 15$ percent. |
| :---: | :---: |
| Harmonic and spurious radiation |  |
|  | . Any emission appearing on a frequency removed from the carrier by between 120 kc and 240 kc , inclusive, is at least 30 db below the level of the unmodulated carrier. |
|  | Any emission appearing on a frequency removed from the carrier by more than 240 kc and up to and including 600 kc is at least 40 db below the level of the unmodulated carrier. |
|  | Any emission appearing on a frequency removed from the carrier by more than 600 kc is at least 80 db below the level of the unmodulated carrier. |
| Modulation characteristics: | . Direct frequency modulation. Standard audio preemphasis is incorporated in modulator. |
| Audio input impedance | 600 ohms balanced. |
| Audio input level . . . | . $+10 \mathrm{dbm} \pm 2 \mathrm{db}$. |
| Audio frequency distortion . . . . . | .50 to $15 \mathrm{kc}, 1.0$ percent maximum. |
| FM noise level. . . | . Not less than 65 db below 100 percent modulation ( $\pm 75 \mathrm{kc}$ ). |
| AM noise level (rms) . | . Not less than 55 db below equivalent 100 percent AM. |

### 2.1 Unpacking and Inspection.

Be careful when uncrating the transmitter and components to avoid damaging the equipment. Inspect the transmitter carefully for scratches, dents, or other physical damage. Check for loose screws and bolts. Inspect all controls, such as switches, for proper operation as far as can be determined without applying power to the transmitter. Examine cables and wiring, making sure that all connections are tight and clear of each other and the chassis. File any damage claims promptly with the transportation company. If such claims are to be filed, retain all packing material.

### 2.2 Transmitter Location

Plan transmitter and wiring placement carefully before starting installation work. Refer to figure 2-1. This
diagram shows the location of all wiring openings in the transmitter cabinets. As will be noted in figure $2-1$, several alternate wiring arrangements can be used. Select the combination that most nearly suits the station requirements.
Allow adequate clearance both in front and back of the transmitter. There should be a minimum clearance of $3-1 / 2$ feet behind the transmitter to provide sufficient room for service work.

If desired, an air duct may be placed over the exhaustair opening in the top of the $10-\mathrm{kw}$ amplifier cabinet to carry heat away from the transmitter.

### 2.3 Intercalinet Conncctions.

Place the two transmitter cabinets beside each other in their permanent location so that, when viewed from

TABLE 2-1. CABINET INTERCONNECTIONS

| WIRE CODE | 10-KW POWER AMPLIFIER TERMINAL BOARD CONNECTIONS (from) | 250-WATT DRIVER TERMINAL BOARD CONNECTIONS (to) |
| :---: | :---: | :---: |
| RE2 | TB402-1 | TB301-1 |
| RE5 | TB402-2 | TB301-2 |
| RE9 | TB402-3 | TB301-3 |
| RC90 | TB403-1 | TB304-1 |
| RC923 | TB403-2 | TB304-2 |
| RC91 | TB403-3 | TB304-3 |
| RC4 | TB403-5 | TB304-5 |
| RC93 | TB403-6 | TB304-6 |
| RC92 | TB403-7 | TB304-7 |
| RC935 | TB403-8 | TB304-8 |
| RCO | TB403-9 | TB304-9 |
| VE9 | TB403-10 | TB304-10 |
| RC91 | TB403-11 | TB304-11 |





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EXTERIOR KNOCKOUTS NE ED NOT BE UEED.


- the front, the 250 -watt driver cabinet is on the left and the $10-\mathrm{kw}$ amplifier cabinet is on the right. Then comect the two cabinets with the appropriate cables. The intercabinet cables consist of an r-f coaxial cable, and a laced control and power cable. Both cables are furnished with the transmitter.

Connect the coaxial cable from the r-f output of the 250 -watt driver to the $\mathrm{r}-\mathrm{f}$ input of the $10-\mathrm{kw}$ power amplificr. These connections are made on the tops of the cabinets. Run the control and power cablethrough the circular openings near the bottom rear of the cabinets on the sides where the cabinets adjoin. Table 2-1 shows the terminal locations of the laced control and power cable and the cable wire code. An explanation of the wire code is given inside the back cover of the instruction book. When the transmitter is received, one end of the laced control and power cable will be connected to the texminal boards of the $10-\mathrm{kw}$ power amplifier. The other end must be connected to the 250 -watt driver at the station site.

### 2.4. Replacement of Compouents Removed for Shipping.

Several of the transmitter components have been removed from the cabinets and packed separately for safety during shipping. These include the plate and screen transformers, the coaxial line resonator center conductor, and fragile units such as power amplifier tubes, mercury-vapor tubes, and crystals. These units should not be replaced in the cabinets until the transmitter is in its permanent location. Wires and cables that were disconnected before shipping have been tagged to facilitate reconnection. Refer to the photographs in section 4 of TD- 538 and TD- 580 for assistance in replacing these components in the correct locations.

## NOTE

When replacing the $10-\mathrm{kw}$ screentransformer, make sure the transformer clears the cabinet blower blades and the components located on the rear of the power panel by at least $3 / 4$ inch.

### 2.5 External Conncctions.

Refer to figure 2-1 for assistance in making the following external connections.
a. Connect the audio input to the transmitter. Bring the audio signal through the bottom (or any one of the optional cable input locations) of the 250 -watt driver cabinet on a shielded twisted pair. Connect the two audio leads to terminals 1 and 2 of TB305 if monaural operation is specified. If optional stereophonic operation is employed, the left audio leads are connected to TB305-1 and 2 and the right audio leads to TB305-4 and 5. TB305 is located about half way up the cabinet on the left side as viewed from the rear of the cabinet. Connect the shield to terminal 3 of TB305.
b. Connect the FM monitor to the monitor output on the top of the $10-\mathrm{kw}$ amplifier cabinet. Refer to figure $2-1$. Use type RG-58/U coaxial cable to make this connection.
c. Connect the antenna transmission line to the r-f output located on top of the $10-\mathrm{kw}$ amplifier cabinet. The r-f connection is for a standard EIA 3-1/8-inch flange.

## caution

Before making this antenna connection, be sure that the transmission line and antemna present a nominal impedance of 50 ohms and an swr of not more than 2:1 at the transmitter operating frequency. If the transmitter output is improperly matched, the transmitter will not operate properly and may be damaged. This is important. See figure $2-8$ for a chart showing swr as a function of power out versus reflected power.
d. Connect the power input cable to the transmitter. This power cable should be brought from an external fused cutout box rated for 100 amperes. Use type $R$ or T AWG \# 4 wire to make these connections. Connect the three wires to terminals 1,2 , and 3 of TB401, located at the bottom left side of the $10-\mathrm{kw}$ amplifier cabinet. The power cable may be brought into the transmitter through holes in either the top or bottom of the $10-\mathrm{kw}$ amplifier cabinet.


The 3-phase power input must be connected properly. To check for proper phasing of the 3 -phase power input, turn off the plate circuit breaker and supply 3 -phase power to the transmitter control circuits. Check the $10-\mathrm{kw}$ pA blower for clockwise rotation. If the blower rotation is not clockwise, reverse any two of the incoming power leads. Always shut off the transmitter 3 -phase power at the external fused cutout box before making any adjustments to the transmitter. Recheck the PA blower for clockwise rotation.

### 2.6 Internal Connections.

The 830F-1A 10-Kw FM Broadcast Transmitter plate, screen, filament, and control circuit transformers are fitted with adjustable taps to compensate for line variations from 200 to 250 volts in 10 -volt steps. To adjust transformers T301, T303, T401; T403, T404, and T405, perform the following steps.
a. Measure the line voltage at the transmitter fused cutout box for each of the three phases. The three voltage readings should be nearly equal.
b. In turn remove the wire from the tapped portion of each transformer and move to the transformer tap


Figure 2-2. Transformer Details
whose voltage rating most closely corresponds to the voltage measured in step a. Refer to figure $2-2$ for transformer terminal numbers and the input voltage which should be applied to each terminal. Do not move the solder lug from transformer terminal 5 of T301 or T401 as this terminal supplies 230 volts to the cabinet fans, the grid bias supply, and the high-voltage power supply filaments.
c. Tighten all transformer terminal connections.

## NOTE

The first three units of the $\mathrm{F} 830-1$ have power amplifier plate and screentransformers with connections numbered differently than in figure 2-2.

TABLE 2-2. PLATE SUPPLY TRANSFORMER CONNECTIONS FOR REDUCED POWER OPERATION

| $\begin{gathered} \text { LINE } \\ \text { VOLTAGE } \end{gathered}$ | AUTHORIZED TRANSMITTER OUTPUT POWER |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | 3-4 KILOWATTS |  |  | 4-5 KILOWATTS |  |  | 5-10 KILOWATTS |  |  |
|  | TYPE CONNECTION | CONNECT TO TRANSFORMER VOLTAGE TAP | plate VOLTS OUT | TYPE <br> CONNECTION | CONNECT TO TRANSFORMER VOLTAGE TAP | Plate VOLTS OUT | TYPE CONNECTION | CONNECT TO TRANSFORMER VOLTAGE TAP | PLATE VOLTS OUT |
| 200 | Delta | 250 | 4500 | Delta | 240 | 4800 | Delta | 200 | 5700 |
| 210 | Y | 165 | 4500 | Delta | 250 | 4800 | Delta | 210 | 5700 |
| 220 | Y | 165 | 4400 | Delta | 250 | 5000 | Delta | 220 | 5700 |
| 230 | Y | 165 | 4600 | Y | 165 | 4600 | Delta | 230 | 5700 |
| 240 | Y | 200 | 4000 | Y | 165 | 4800 | Delta | 240 | 5700 |
| 250 | Y | 200 | 4100 | Y | 165 | 5000 | Delta | 250 | 5700 |

TABLE 2-3. SCREEN GRID SUPPLY TRANSFORMER CONNECTIONS FOR REDUCED POWER OPERATION

| $\begin{gathered} \text { LINE } \\ \text { VOLTAGE } \end{gathered}$ | AUTHORIZED TRANSMITTER OUTYUT POWER |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | 3-7 KILOWATTS |  |  | 7-10 Kilowatts |  |  |
|  | $\begin{gathered} \text { TYPE } \\ \text { CONNECTION } \end{gathered}$ | CONNECT TO TRANSFORMER VOLTAGE TAP | SCREEN volts out | TYPE CONNECTION | CONNECT TO TRANSFORMER VOLTAGE TAP | SCREEN volts OUT |
| 200 | Delta | 240 | 625 | Delta | 200 | 750 |
| 210 | Delta | 250 | 630 | Delta | 210 | 750 |
| 220 | Delta | 250 | 660 | Delta | 220 | 750 |
| 230 | Y | 165 | 630 | Delta | 230 | 750 |
| 240 | Y | 165 | 650 | Delta | 240 | 750 |
| 250 | Y | 165 | 680 | Delta | 250 | 750 |

Check that a jumper wire is installed from TB404-7 to TB404-8 to ensure that 115 -volt power will be applied to the transmitter control circuitry.

If the optional stereo generator is installed any time after the initial $830 \mathrm{~F}-1 \mathrm{~A}$ installation, the $18-\mathrm{db}$ audio pad will have to be removed from the audio circuitry of the exciter. Refer to Unit Instructions TD-536 for the location of the $18-\mathrm{db}$ audio pad.

### 2.7 Reduced Power Operalion.

If the authorized transmitter power output is below 10 kilowatts, the plate and screen voltage and the power amplifier drive will have to be reduced accordingly. Refer to tables 2-2 and 2-3 for the plate and screen power supply transformer hookup for reduced output. The tables refer to the transformer primary windings in each case. Figures $2-3$ and $2-4$ show the recommended control grid and screen grid currents for output power ratings from 3 to 10 kilowatts of transmitter output power.

For example, if the authorized transmitter power is 4.5 kilowatts, and the measured line voltage is 230 volts, table 2-2 shows the plate supply transformer primary will be connected in a wye configuration. To obtain a wye connected primary, connect transformer terminals 1,8 , and 15 together and remove the adjustable jumpers. As the measured line voltage is 230 volts, connect the 3 -phase line to the 165 -volt terminals $X, Y$, and $Z$ as shown in table 2-2. (The 165 terminology is only for reference. The actual winding voltage is less than 165 volts.) This will give a plate supply output voltage of approximately 4600 volts. The actual transformer tap number for the given tap voltage may be obtained by referring to figure $2-2$. As the authorized power output is 4.5 kw and the
line voltage 230 volts, the screen supply transformer will also be connected in a wye configuration. (Refer to table 2-3.) The screen supply transformer wye primary is obtained by connecting together transformer terminals 1,8 , and 15 , and removing the adjustable jumpers. Connect the 3 -phase line input to 165 -volt terminals $X, Y$, and $Z$. This will give a screen voltage of approximately 630 volts.
With the plate and screen transformer now connected correctly for a 3 -phase input of 230 volts and a $4.5-\mathrm{kw}$ output, the transmitter may be tuned according to the tuning procedure given in paragraph 2.11. When tuning for reduced power operation, substitute drive and screen currents taken from figures 2-3 and 2-4 for those given in the tuning procedure.

## NOTE

If, when tuning the power amplifier, sufficient power output cannot be obtained by adjusting the POWER OUTPUT ADJUST potentiometer, connect the plate and screen transformers as shown in the next lower line voltage column of tables 2-2 and 2-3.

### 2.8 Remote Control.

Remote control of $830 \mathrm{~F}-1 \mathrm{~A}$ 10-Kw FM Broadcast Transmitter can easily be accomplished by connection to terminal boards TB302 and TB303, located in the 250 -watt power amplifier cabinet, and by placing a jumper between TB404-4 and TB404-9. Terminal board T'B404 is located within the $10-\mathrm{kw}$ power amplifier cabinet. Table 2-4 lists the terminal board connections and the remote functions of each pair of


Figure 2-3. Recommended Screen Grid Current for Reduced Power Operation


Figure 2-4. Recommended Control Grid Current for Reduced Power Operation
terminals. Remote "on" switches should be the normally open momentary type. Remote "off" switches should be the normally closed momentary type. For remote operation, the LOCAL-REMOTE switch within the 250 -watt power amplifier cabinet should be in the REMOTE position. When in the REMOTE position, it is possible to control the transmitter from the transmitter panel switches or from the remote point.

TABLE 2-4
REMOTE CONTROL CONNECTIONS

| FUNCTION | TERMINALS |  |
| :--- | :---: | :---: |
|  | TB302 | TB303 |
| FILAMENT ON |  | 1 and 2 |
| FILAMENT OFF | 8 and 9 |  |
| PLATE OFF |  | 2 and 4 |
| PLATE ON |  | 2 and 3 |

For simplified operation, the FILAMENT ON and PLATE OFF switches could be eliminated. The PLATE ON switch starts a sequence of operations which turns the filaments on and the plate voltage on after the filament time delay is accomplished. The FILAMENT OFF switch shuts down all transmitter functions.
Equipment is available that will completely control and monitor transmitter operation from a remote location through standard telephone pairs. When such remote control equipment is used, necessary installation and connection information will be supplied with the remote equipment.
If an optional stereo generator is employed in the 830F-1A, remote control of the stereo mode may be accomplished by a ground on TB302-7. If the ground is present, the transmitter will be inthe stereo mode. If the ground is removed, the transmitter will switch to the monaural mode. Local control of the stereo mode is also available at the transmitter.

### 2.9 Frequency Changc.

If the transmitter operating frequency is changed, five components will have to be changed or adjusted. The components are the exciter heterodyning crystal, Y426; the driver plate slider; the power amplifier grid tank slider; the power amplifier plate tank slider; and the neutralizing bars of the power amplifier.
Table 2-5 lists the channel frequency versus crystal frequency and the Collins part number for each crystal. Figure 2-5 shows the distance the driver plate tank slider should be positioned from the deck plate (tube socket mounting plate) for each operating frequency between 88 and 108 megacycles. Figure $2-6$ shows the distance the power amplifier plate tank and grid tank sliders should be positioned from the deck plate for the operating frequencics.

### 2.9.1 NE UTRALIZATION PROCEDURE

> WARAIAG

Voltages are present in this transmitter that are dangerous to life. Observe safety precautions when performing any inspection or work within the cabinet or plate resonator. Do not depend entirely on the interlocks. Always shut down the transmitter before doing any work inside the transmitter cabinet. Remember that 115 volts is present on one side of the door interlocks.

Power amplifier neutralization must be completed after any change in frequency to provide the proper degree of amplifier stability. Neutralization of the power amplifier is accomplished as follows: Set the neutralization sliding contacts to the position corresponding to the assigned station frequency as shown in figure 2-7. Turn on the exciter and driver, and tune the exciter and driver according to the procedure given in paragraph 2.11. Tune the power amplifier gricl and plate tank. Turn the POWER OUTPUT ADJUST control to its maximum counterclockwise position. Turn off the exciter. Turn off the driver platecircuit breaker. Turn on the power amplifier plate voltage and remove the power amplifier control grid bias fuse while observing the WATTMETER for an indication. The WATTMETER switch must be in the FORWARD position. If no indication is noted on the WATTMETER, the power amplifier is properly neutralized.

## CAUYION

Do not allow the power amplifier plate current to exceed 2 amperes during the test. Do not run the test for more than 5 seconds without a 1 -minute cooling off period between tests.

If a WAT TMETER indication is noted with the removal of the control grid bias fuse, readjust the neutralizing sliding contacts and repeat the test.
The neutralization can be checked further by noting the grid reaction as the plate is tuned through resonance. Minimum or no grid current reaction indicates proper neutralization. Neutralization of the $10-\mathrm{kw}$ power amplifier is normally not critical and should be easily adjusted.

### 2.10 Final Installation Procedure.

a. Check that all tubes, both in the exciter and power amplifiers, are securely in place.
b. Check all exciter transistors for proper place-- ment. The transistor locating tab must be pointing to the transistor socket locating mark placed adjacent to the transistor socket.
c. Check fuses on both the exciter and power amplifiers.

TABLE 2-5. CRYSTAL PART NUMBERS

| $\begin{gathered} \text { CHANNEL } \\ \text { FREQ. } \\ \text { (me) } \end{gathered}$ | $\begin{aligned} & \text { CRYSTAL } \\ & \text { FREQ } \\ & \text { (mc) } \end{aligned}$ | COLLINS <br> PART NUMBER | $\begin{aligned} & \text { CHANNEL } \\ & \text { FREQ } \\ & \text { (mc) } \end{aligned}$ | $\begin{aligned} & \text { CRYSTAL } \\ & \text { FREQ } \\ & \text { (mc) } \end{aligned}$ | COLLINS <br> PART NUMBER |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 88.1 | 74.10000 | 289-2744-00 | 98.1 | 84.10000 | 289-2704-00 |
| 88.3 | 74.30000 | 289-2745-00 | 98.3 | 84.30000 | 289-2795-00 |
| 88.5 | 74.50000 | 289-2746-00 | 98.5 | 84.50000 | 289-2796-00 |
| 88.7 | 74.70000 | 289-2747-00 | 98.7 | 84.70000 | 289-2797-00 |
| 88.9 | 74.90000 | 289-2748-00 | 98.9 | 84.90000 | 289-2798-00 |
| 89.1 | 75.10000 | 289-2749-00 | 99.1 | 85.10000 | 289-2799-00 |
| 89.3 | 75.30000 | 289-2750-00 | 99.3 | 85.30000 | 289-2800-00 |
| 89.5 | 75.50000 | 289-2751-00 | 99.5 | 85.50000 | 289-2801-00 |
| 89.7 | 75.70000 | 289-2752-00 | 99.7 | 85.70000 | 289-2802-00 |
| 89.9 | 75.90000 | 289-2753-00 | 99.9 | 85.90000 | 289-2803-00 |
| 90.1 | 76.10000 | 289-2754-00 | 100.1 | 86.10000 | 289-2804-00 |
| 90.3 | 76.30000 | 289-2755-00 | 100.3 | 86.30000 | 289-2805-00 |
| 90.5 | 76.50000 | 289-2756-00 | 100.5 | 86.50000 | 289-2806-00 |
| 90.7 | 76.70000 | 289-2757-00 | 100.7 | 86.70000 | 289-2807-00 |
| 90.9 | 76.90000 | 289-2758-00 | 100.9 | 86.90000 | 289-2808-00 |
| 91.1 | 77.10000 - | 289-2759-00 | 101.1 | 87.10000 | 289-2809-00 |
| 91.3 | 77.30000 | 289-2760-00 | . 101.3 | 87.30000 | 289-2810-00 |
| 91.5 | 77.50000 | 289-2761-00 | 101.5 | 87.50000 | 289-2811-00 |
| 91.7 | 77.70000 | 289-2762-00 | 101.7 | 87.70000 | 289-2812-00 |
| 91.9 | 77.90000 | 289-2763-00 | 101.9 | 87.90000 | 289-2813-00 |
| 92.1 | 78.10000 | 289-2764-00 | 102.1 | 88.10000 | 289-2814-00 |
| 92.3 | 78.30000 | 289-2765-00 | 102.3 | 88.30000 | 289-2815-00 |
| 92.5 | 78.50000 | 289-2766-00 | 102.5 | 88.50000 | 289-2816-00 |
| 92.7 | 78.70000 | 289-2767-00 | 102.7 | 88.70000 | 289-2817-00 |
| 92.9 | 78.90000 | 289-2768-00 | 102.9 | 88.90000 | 289-2818-00 |
| 93.1 | 79.10000 | 289-2769-00 | 103.1 | 89.10000 | 289-2819-00 |
| 93.3 | 79.30000 | 289-2770-00 | 103.3 | 89.30000 | 289-2820-00 |
| 93.5 | 79.50000 | 289-2771-00 | 103.5 | 89.50000 | 289-2821-00 |
| 93.7 | 79.70000 | 289-2772-00 | 103.7 | 89.70000 | 289-2822-00 |
| 93.9 | 79.90000 | 289-2773-00 | 103.9 | 89.90000 | 289-2823-00 |
| 94.1 | 80.10000 | 289-2774-00 | 104.1 | 90.10000 | 289-2824-00 |
| 94.3 | 80.30000 | 289-2775-00 | 104.3 | 90.30000 | 289-2825-00 |
| 94.5 | 80.50000 | 289-2776-00 | 104.5 | 90.50000 | 289-2826-00 |
| 94.7 | 80.70000 | 289-2777-00 | 104.7 | 90.70000 | 289-2827-00 |
| 94.9 | 80.90000 | 289-2778-00 | 104.9 | 90.90000 | 289-2828-00 |
| 95.1 | 81.10000 | 289-2779-00 | 105.1 | 91.10000 | 289-2829-00 |
| 95.3 | 81.30000 | 289-2780-00 | 105.3 | 91.30000 | 289-2830-00 |
| 95.5 | 81.50000 | 289-2781-00 | 105.5 | 91.50000 | 289-2831-00 |
| 95.7 | 81.70000 | 289-2782-00 | 105.7 | 91.70000 | 289-2832-00 |
| 95.9 | 81.90000 | 289-2783-00 | 105.9 | 91.90000 | 289-2833-00 |
| 96.1 | 82.10000 | 289-2784-00 | 106.1 | 92.10000 | 289-2834-00 |
| 96.3 | 82.30000 | 289-2785-00 | 106.3 | 92.30000 | 289-2835-00 |
| 96.5 | 82.50000 | 289-2786-00 | 106.5 | 92.50000 | 289-2836-00 |
| 96.7 | 82.70000 | 289-2787-00 | 106.7 | 92.70000 | 289-2837-00 |
| 96.9 | 82.90000 | 289-2788-00 | 106.9 | 92.90000 | 289-2838-00 |
| 97.1 | 83.10000 | 289-2789-00. | 107.1 | 93.10000 | 289-2839-00 |
| 97.3 | 83.30000 | 289-2790-00 | 107.3 | 93.30000 | 289-2840-00 |
| 97.5 | 83.50000 | 289-2791-00 | 107.5 | 93.50000 | 289-2841-00 |
| 97.7 | 83.70000 | 289-2792-00 | 107.7 | 93.70000 | 289-2842-00 |
| 97.9 | 83.90000 | 289-2793-00 | 107.9 | 93.90000 | 289-2843-00 |




Figure 2-5. Driver Plate Cavity Tuning Chart


Figure 2-6. Power Amplifier Plate and Grid Tuning Chart


Figure 2-7. Neutralization Tuning Chart
d. Insert the crystals into the exciter crystal sockets. The 14 -me crystal should be placed into Y501 and the heterodyning crystal into Y426.
e. Carefully inspect all rear door and panel interlocks to be certain that they are operating correctly. Do this by pressing the contact block, located on the doors and the lower rear panels, until the spring is completely compressed, then releasing the block. If the block does not spring back to itsoriginal position, adjust until it operates properly.
f. Recheck all cables and wiring in the transmitter to be sure that all connections are made securely and properly.
g. Check that the sliders in the tuning resonators of both the driver and the power amplifier are in proper
position for the assigned station frequency and that the sliders are tight against the outer resonator walls.
$h$. Check the final power amplifier tube for proper seating. This tube is pressed into place and may have jarred loose during installation.
i. Replace the lower rear panels on the transmitter cabinets. Insert the ventilating fan plugs into the sockets provided.

### 2.11 Starting the Transmitter in a New Installation.

Before starting the transmitter for the first time, read section 3 of the subunit instruction books to become familiar with the location and function of the various transmitter controls. Then, perform the following procedures.


Voltages are present in this transmitter that are dangerous to life. Observe safety precautions when making any transmitter adjustments. Do not reach inside the rear of the transmitter cabinets whenever high voltages are applied. Do not depend entirely on door interlocks. Always shut off transmitter power at the external cutout box and ground all capacitors in the transmitter cabinets with the shorting stick before doing any work inside the rear of the cabinets. When working in the power amplifier resonators, remember that 115 volts $\mathrm{a}-\mathrm{c}$ is present on one side of the resonator compartment interlocks. Keep metal tools and all parts of the body away from transistor cases.
a. Complete the entire transmitter installation procedure as directed in earlier paragraphs of this instruction book.
b. Close the doors at the rear of the cabinets. Open the doors at the front of the cabinets and remove the lower front panels so the entire inside panel is exposed. Check to be sure that the covers on the resonator compartments of both the driver and power amplifier are securely closed and latched.
c. Set the PLATE circuit breaker in the driver and the PLATE and FILAMENT breakers in the power amplifier to ON.
d. Press the FILAMENT ON switch on either cabinet. The green indicator lamps at the top left of both cabinets should light. This means that all transmitter tube filaments and cooling air blowers are operating.
e. Set the right cabinet MULTIMETER switch to FIL $V$ 8V FS. Adjust the FILAMENT VOLTAGE ADJUST control until the right cabinet MULTIMETER indicates 7.5 volts. If this meter indication cannot be reached, readjust the taps on the filament transformer, T403, to the next lower voltage tap.

## NOTE

If mercury-vapor rectifier tubes are used in the $10-\mathrm{kw}$ amplifier (right cabinet), run the tubes with only the filaments lighted for at least $1 / 2$ hour before applying plate voltage to the amplifier. This is done to remove any mercury coating that may be on tube elements. This operation is necessary only for new tubes or used tubes that have been inverted or agitated. While this aging process is being completed, the exciter and 250 -watt amplifier may be tuned by performing the following steps.
f. Set the POWER switch on the exciter to ON. Allow the exciter oscillator to warm up for 15 minutes.
g. Turn S 101 to the 14 MC REF B position and check M101 for an indication in the $B$ meter range. Turn

S101 to the AFC KEY B position and check M101 for an indication in the $B$ meter range.

## NOTE

When S101 is in the AFC KEY B position, the meter pointer will not hold steady but will pulse at approximately 5 cps , the keying generator rate. This pulse is an indication of normal operation.

Turn S101 to the MOD OUTPUT B position and check M101 for an indication in the B meter range. If all meter indications for the three S101 positions fall within the $B$ meter range, proceed with the following tuning procedures. If any of the meter indications fall outside of the $\mathbf{B}$ meter range, the modulator and afc discriminator are out of adjustment and will have to be adjusted according to the maintenance section of this instruction book.
h. Switch S101 to the MIXER GRID A position. Set POWER OUT resistor R454 to its midposition. Using a nonmetallic screwdriver type tuning tool, adjust the BUF TUNE control for a peak indication on meter M101.

## NOTE

The MIX BAL control, R428, should be placed in its midrange position. No further adjustment of this control is necessary unless the transmitter frequency falls within the range of 97 to 100 mc . See step n if the transmitter frequency falls within this range.
i. Switch S101 to the V428B position. Adjust L429 and L430 for a maximum indication on M101.
j. Switch S101 to the V429B position. Adjust L431 and L432 for a maximum indication on M101.
k. Switch S101 to the V430B position. Adjust L433 and L434 for a maximum indication on M101.

1. Switch S 101 to the V430C position. Adjust the PA PLATE control for a minimum indication on M101. m . Set the driver MULTIMETER switch to GRID FS 40 MA . Adjust first the exciter PA MATCH control, then the power amplifier GRID TUNING control for a peak MULTIMETER indication.
n . If the transmitter frequency falls between 97 and 100 mc , the following additional step will have to be completed. Place a grid dip meter tuned to 98 mc near the exciter output. Adjust the MIX BAL control for a minimum output as indicated on the grid dip meter.
o. Set the driver POWER OUTPUT ADJUST control fully counterclockwise. Connect a 50 -ohm, 250 -watt dumimy load to the r-f output connector of the driver.
p. Press the driver PLATE ON switch. The red indicator lamp at the top right of the cabinet should light, and the P. A. PLATE VOLTAGE meter should indicate $2100 \pm 100$ volts.
q. Set the driver MULTIMETER switch to SCREEN FS 400 VDC. The MULTIMETER should indicate 300 $\pm 35$ volts.
r. Adjust the driver PLATE TUNING control for a dip in the P.A. PLATE CURRENT meter indication.
s. Set the driverWATTMETER switch to FORWARD. Adjust the OUTPUT COUPLING control for approximately 5 ma of screen curreat.
t. Turn the driver POWER OUTPUT ADJUST control approximately two-thirds of its maximum clockwise rotation.
u. Increase the driver coupling a small amount by turning the OUTPUT COUPLING control clockwise until the PA screen current is reduced to approximately 10 ma . Adjust the PLATE TUNING control for a dip in the P.A. PLATE CURRENT meter indication. (The plate tuning capacitor should be near its center position when the dip in the driver plate current occurs. If the capacitor is not in this position, move the plate slider in the appropriate direction and repeat step u.)
v. Rotate the POWER OUTPUT ADJUST control clockwise a small amount.
w. Repeat steps $u$ and $v$ until the P.A. PLATE CURRENT meter indicates a minimum inclication and the WATTMETER indicates 250 watts. At this time the driver screen current should be not less than 5 ma nor more than 20 ma .
$x$. Turn off the driver plate power and disconnect the dummy load from the driver. Reconnect the coaxial cable from the driver to the power amplifier. Set the driver POWEROUTPUT ADJUST potentiometer. fully counterclockwise.
y. Set the driver WATTMETER switch to REFLECTED. Adjust the power amplifier GRID TUNING control for minimum reflected power.

## NOTE

If a low value of reflected power cannot be obtained, change the setting of the power amplifier (right cabinet) grid slider slightly. Adjust the position of the $r-f$ input tap and repeat step y.
2. Turn the driver POWEROUTPUT ADJUST control until the proper PA grid current is flowing (proper value approximately 50 ma obtained from figure 2-4).
aa. Set the power amplifier POWER OUTPUT ADJUST controi in the maximum counterclockwise direction. Check the power amplifier filament voltage and. if necessary readjust the FILAMENT VOLTAGE ADJUST control for 7.5 volts as indicated on the power amplifier MULTIMETER. If the transmitter filaments have been energized for at least $1 / 2$ hour (if mercury-vapor rectifiers are used), press the power amplifier PLATE ON switch. When this switch is pressed, the red indicator lamp at the top right of the power amplifier cabinet should light and the P.A. PLATE VOLTAGE should indicate $6450 \pm 200$ volts. ab. Set the WATTMETER switch to FORWARD. Adjust the power amplifier PLATE TUNING control for maximum WATTMETER indication. (The power amplifier plate tuning capacitor should be near its center position when the maximum WATTMETER indication occurs. If the capacitor is not in this
position, move the power amplifier plate slider in the appropriate direction and repeat step ab.)


The power amplifier P.A. PLATE CURRENT meter indication should never exceed 2.5 amperes (power supply rating).
When tuning, be sure that the plate dissipation does not exceed 6 kw at any time. Plate dissipation is:

$$
\text { Plate dissipation }=E_{p} \mathbf{I}_{p}-\text { power out }
$$

ac. Set the power amplifier WATTMETER switch to REFLECTED. The swr must be less than 2:1 at all times (see figure 2-8). If the indication of reflected power is excessive, check the antenna and its associated $r$-f cable for a possible mismatch.
ad. Set the power amplifier WATTMETER switch to FORWARD. Adjust the OUTPUT COUPLING control for approximately 100 ma of screen current.
ae. Turn the power amplifier POWER OUTPUT ADJUST control to approximately its midpoint position. Recheck the driver power output for proper PA grid current.

## NOTE

The values of screen current given in the steps are only approximate. The tube dissipation will allow for greater variations in screen current without adversely affecting tube operation.
af. Increase the power amplifier coupling a small amount by turning the OUTPUT COUPLING control clockwise until the PA screen current is reduced to approximately 200 ma (or value obtained from figure 2.3). Adjust the PLATE TUNING control for maximum WATTMETER indication.
ag. Rotate the power amplifier POWER OUTPUT ADJUST control clockwise a small amount.
ah. Repeat steps af and ag until the power amplifier P. A. PLATE CURRENT meter indicates the transmitter output is 10 kw (or authorized power) as measured by the indirect method. At this time, the PA screen current should be approximately 200 ma (or reduced current for reduced power applications obtained from figure 2-3). The indirect method of measuring power output is:

$$
\text { Power output }=\mathrm{I}_{\mathrm{p}} \mathrm{E}_{\mathrm{p}} \mathrm{~K}
$$

K or efficiency for determining the transmitter power output is obtained from the production test data supplied with the transmitter. If the final does not tune up properly, the fault may lie in improper neutralization. If faulty neutralization is suspected, neutralize the final according to the procedure given in paragraph 2.9.1.


Figure 2-8. Directional Wattmeter SWR Determination Table
ai. Check to be sure that the FM monitor that is connected to the transmitter is properly calibrated. Then, if necessary, adjust the exciter VHFOSC FREQ ADJ control until the monitor indicates that the transmitter operating frequency is within specified operating limits.
aj. Apply 50 -cps audio tone to the transmitter input. The input level should be such that the voltage at the transmitter audio input terminals is +10 dbm .
ak. Adjust the exciter MOD GAIN control until the monitor indicates 100 percent modulation.
al. Replace the lower front panel on the transmitter cabinet and close the front doors. The transmitter is now ready for standard broadcast use.

## NOTE

At this point it is suggested that a record be made of all meter readings for future maintenance and trouble shooting. These meter readings maybe recorded in table 5-3.

### 3.1 Gencral.

Refer to the unit instruction books to become familiar with the operation and function of controls on the power amplifier, the driver, and the exciter.

After the transmitter has been placed in operation it will be necessary to check meter indications from time to time to be sure the transmitter is operating properly and to occasionally touch-up the power amplifier loading and tuning.

### 3.2 Starting the Transmitter in Normal Operation.

The transmitter may be put into operation by two different methods, depending upon the circumstances. For normal operation, press either FILAMENT ON switch to start the driver and power amplifier filaments and to warm up the exciter (the exciter POWER switch should be left in the ON position at all times). Check the driver grid current to be sure the exciter is presenting sufficient drive to the driver before applying plate power. Approximately 30 secondsafter filament power is applied, press the driver PLATE ON switch.

Check the operation of the driver and, if it is operating properly, press the power amplifier PLATEON switch, starting the transmitter.

The alternate method of starting the transmitter consists of pressing the power amplifier PLATE ON switch only. The power amplifier filament, the driver filament, and the exciter will immediately start to warm up. As soon as the driver time delay relay has completed its cycle, the driver and power amplifier plate voltages will come on automatically, starting the transmitter.

To shut down the transmitter it is recommended, but not necessary, to press the driver PLATE OFF switch, wait a few seconds, and then shut off the filament and exciter power by pressing either FILAMENT OFF switch. It is also possible to press either FILAMENT OFF switch which immediately removes plate, filament, and exciter power. Pressing the driver PLATE OFF switch first allows the plate power supply voltages to discharge through the driver and power amplifier while the filament is at normal operating temperature and, in addition, cools the power amplifier components.

## seciion <br> 

principles of operation

### 4.1 Gencral.

Refer to figure 4-1, a block diagram of 830F-1A $10-\mathrm{Kw}$ FM Broadcast Transmitter. The transmitter can be broken down into four main subassemblies; an exciter, a driver, a power amplifier, and harmonic filter. Refer to section 2 of the unit instruction books for a complete explanation of the exciter, the driver, and the power amplifier.

### 4.2 A830-2 10 W Wide-Band FM Broadcast Exciter.

Refer to figure 4-1, a block diagram of the 830F-1A. The baseband audio is coupled to the A830-2 through a pre-emphasis network, and through an 18-db pad to J604 and two baseband amplifiers. The baseband output is coupled to a voltage-sensitive capacitor, C654. C654 is a diode which varies in capacity in proportion to the voltage across it. The FM oscillator is tuned to 14 mc . The capacity of C 654 varies in proportion to the baseband audio and therefore the output is a $14-\mathrm{mc}$ signal frequency modulated by the baseband audio. The deviation of the $14-\mathrm{mc}$ signal is $\pm 75 \mathrm{kc}$ for 100 percent modulation. The osciallator output is coupled through two limiters to remove any amplitude modulation. The limited $14-\mathrm{mc}$ signal is then amplified and a portion of the output coupled to the rate correction frequency discriminator and to the output amplifier. The output of the frequency discriminator is simply the baseband audio detected from the modulated $14-\mathrm{mc}$ signal. This detected audio is coupled back to the baseband input to correct for any nonlinearity in C654.

A portion of the limiter output is also coupled to the afc buffer stage. The afc buffer output, the modulated 14 -me signal, is coupled to the reference oscillator and afc limiters through a diode switch. The output of the 14 -mc reference oscillator is also coupled to the reference oscillator and afc limiters through a diode switch. The diode switch is operated by a 5 -cps keying generator. The $5-\mathrm{cps}$ generator is a unijunction transistor operating as a relaxation oscillator keying a multivibrator.

The diode switch alternately connects the modulated $14-\mathrm{mc}$ signal (afc buffer output) and the 14 -me reference signal to the afc discriminator. The afc discriminator dctects the difference between the $14-\mathrm{mic}$ reference signal and the modulated $14-\mathrm{mc}$ signal. The modulated $14-\mathrm{mc}$ signal will cause a baseband audio output at the discriminator. This is not an error in frequency, so a portion of the baseband audio input is amplified by the baseband canceling amplifier and fed
into the output of the frequency discriminator through a diode switch. This diode switch iskeyed by the same 5 -cps signal which switched the reference oscillator and afc limiter input. When the modulated 14 -mc signal is connected to the referenceoscillator and afclimiter input, the baseband canceling signal is switched into the output of the frequency discriminator to cancel the baseband output from the discriminator.

The input signal to the four error signal amplifiers is a 5 -cps square wave. The amplitude of this square wave is proportional to the frequency error in the FM oscillator. The exror signal amplifier square wave output is converted to a d-c control signal in the synchronous detector. The synchronous detector is also keyed by the 5 -cpskeying signal. Thed-c error signal is coupled to C654 to correct the frequency modulation oscillator.

A portion of the limiter output is fed to the output amplifier. The modulated $14-\mathrm{mc}$ signal from the output amplifier is then heterodyned up to the operating frequency in a balanced mixer. The crystal frequency is 14 mc below the customer's operating frequency. The crystal oscillator output is coupled to a buffer stage and is mixed with the modulated $14-\mathrm{mc}$ signal in the balanced mixer. The balanced mixer output is limited and amplified to the 10 -watt $\mathbf{r}$-f output level. The output impedance of the A830-2 is between 50 and 70 ohms.

The power supply for the A830-2 is of conventional design and supplies operating voltages for the vacuum tubes and transistors in the A830-2.

### 4.3 Driver and Final.

The driver consists of a single ceramic-type tetrode tube. The tube is operated as a class $C$ amplifier with a tuned coaxial line resonator plate circuit. The output from the driver is fed to a $10-\mathrm{kw}$ final amplifier.

The final power amplifier is also made up of a single ceramic-type tetrode tube. The tube is operated as a grounded screen, class C amplifier, using screen grid neutralization. The plate works into a tuned coaxial line resonator with a similar tuned coaxial line resonator for the control grid tank. The plate output is fed through a harmonic filter which reduces all output harmonics into the antenna.

The harmonic filter consists of two series-resonant M-derived low-pass end sections and two constant-K,


T center sections. The harmonic filter starts to attenuate above 110 mc and reaches maximum attenuation at the carrier second harmonic. The attenuation pattern then tapers off slowly as the frequency rises. The over-all result of the harmonic filter is in keeping the harmonics attenuated at least 80 db below the carrier frequency.

### 4.3.1 CONTROL CIRCUITS.

One phase of the 230 -volt, 3 -plase power is stepped down to 115 volts a-c by transformers T301 and T401. This lower voltage is used to activate relays in the transmitter control circuits and is also fed to the exciter as its primary power source. The control circuits allow power to be applied to the transmitter only in the proper sequence to prevent damage to the driver and final amplifier. These circuits also contain protective devices to prevent damage to components from accidental overloads.

### 4.3.2 PLATE CONTACTORS AND POWER SUPPLIES.

The driver plate contactor consists of a heavy-duty relay which controls the 230 -volta-c primary power to
the plate power supply. The driver plate contactor is actuated by the clriver PLATE ON switch througl the control circuitry.

The driver plate power supply is a conventional fullwave power supply capable of delivering 2100 volts d-c at 250 ma to the driver.

The power amplifier plate contactors are heavy-duty relays which control the 3 -phase power supply and are activated by the power amplifier PLATE ON switch. The two plate contactors are energized in a step-start sequence (switches resistances momentarily into the line) to reduce the generation of transients in the power supply when power is first applied.

The power amplifier plate power supply is a conventional 3-phase full-wave power supply capable of providing 5700 volts at 2.5 amperes. The screen power supply is a full-wave 3 -phase silicon diode power supply capable of providing 750 volts at 2.6 amperes. The control grid bias supply places the control grid at about cutoff and is set at the factory.

### 5.1 General.

The following paragraphs contain information concerning maintenance of $830 \mathrm{~F}-1 \mathrm{~A}$ 10-Kw FM Broadcast Transmitter.


Voltages are present in this transmitter that are dangerous to life. Observe safety precautions when performing any maintenance. Do not reach inside the transmitter cabinets whenever high voltage is applied. Do not depend on door interlocks. Always shut down the transmitter before doing any work inside the transmitter cabinets. Immediately upon opening the rear cabinet doors, short out the power supply capacitors with the shorting sticks provided with the transmitter.

Refer to the applicable unit instructions for specific maintenance procedures for each subassembly.

### 5.2 Normal Tuning Procedures.

The following are tuning procedures which should be observed after the transmitter has been installed and tuned according to the installation procedures given in section 2. Table 5-1 presents abbreviated tuning instructions to be used with the following procedures. Table 5-1 can be detached from this instruction book, placed within one of the cabinet doors or adjacent to the transmitter, and used as a reference when the cperator becomes familiar with the transmitter.
a. Open the doors at the front of the cabinets.
b. Press either FILAMENT ON switch and allow the exciter to warm up for at least 15 minutes.
c. Set S101 on the exciter to MIXERGRID A. Adjust BUF TUNE control for a peak on M101.
d. Set S101 to V428 B and adjust L429 and L430 for a peak on M101.
e. Set S101 on V429 B and adjust LA31 and L432 for a peak on M101.
f. Set S101 to V430 B and adjust L433 and L434 for a peak on M101.
g. Set S101 to V430C B and adjust the PA PLATE control for a minimum indication on M101.

## NOTE

Convenient marked ranges are available on the meter which correspond to switch positions. These ranges give an approximate requirement for min-max readings for each switch position.
h. Tune the exciter coupling by setting the MULTIMETER switch to GRID FS 40 MA and adjusting the exciter PA MATCH control for maximum coupling. Tune the driver GRID TUNING control for a peak MULTIMETER indication.
i. Set the driver WATTMETER to REFLECTED.
j. Press the driver PLATE ON switch. Adjust the power amplifier GRID TUNING for a dip in the R.F. WATTMETER indication. Set the driver to FORWARD. In turn, adjust the driver PLATE TUNING control for minimum indication on the driver P.A. PLATE CURRENT meter and increase the OUTPUT COUPLING control and the POWER OUTPUT ADJUST control, by small amounts, until the proper value of PA grid current is flowing (function of power level; see figure 2-4). The driver screen current may be 0 but should not be more than 20 ma when the driver is properly tuned.
k. Adjust the power amplifier FILAMENT VOLTAGE ADJUST control for 7.5 volts as shown on the MULTIMETER.

1. Press the power amplifier PLATE ON switch.
m. Adjust the power amplifier GRID TUNING control for a maximum indication on the MULTIMETER. (This should be near the minimum reflected power on the driver wattmeter.). Place the power amplifier WATTMETER switch in the FORWARD position. n. In turn, adjust the power amplifier PLATE TUNING control for a maximum WATTMETER indication and increase the OUTPUT COUPLING control and the POWER OUTPUT ADJUST control, by small amounts, until the transmitter is producing authorized power as measured by the indirect method.

$$
\text { Power output }=I_{p} E_{p} K
$$

The power amplifier screen current should be at the value specified in figure 2-3 for the transmitter output power level when the power amplifier is properly tuned. Normal operation is achieved when the R.F. WATTMETER indication is maximum. Efficiency may be improved by detuning plate circuit slightly from maximum power to the high frequency side of resonance (less capacity).
o. Adjust the VHF OSC FREQ ADJ control until the FM monitor inclicates the transmitter is operating within the specified operating limits.

### 5.3 Modulator and AlיC Discriminator Adjustment Procedures.

The broadband exciter is designed to be exceptionally stable and will require few adjustments over a long period of time. The following adjustment procedures should only be followed if the exciter is not operating within limits upon installation, or if any of the transistors (Q503, Q511, Q601, and Q604) are replaced. Replacement of other components should not normally require a change in the adjustments given in this section. The transmitter will have to be energized when performing steps $c$ and $e$. Use an a-c vtvm such as a Ballantine Model 310A when making adjustments. A d-c vtvm such as a Heath type may be used in step $f$.

## NOTE

When adjusting the modulator and afc section of the exciter, use a nonmetallic hex type adjusting tool. Keep all metallic tools, the hands, and other parts of the body away from transistor cases. When disabling the afc with the complete transmitter in operation, check the station monitor to be sure the center frequency stays within the FCC requirements.
a. Remove transistor Q509 and place a vtym from TP501 to ground. Tune L505 and L504 for a peak indication on the vtvm. Be sure to tune the inductances in the order given to minimize the limiting effect. Check that each stage is limiting when making these adjustments. Limiting will show up as a broad flat peak on the vtvm when tuning. Set the controls midway between the limiter fall-off points shown on the vtvm. Do not replace Q509 at this time.
b. Tune the afc discriminator by placing a vtum from TP501 to ground and tune the DISCR PRI control, C515, for maximum indication. Place the vtvm from TP502
to ground and adjust the DISCR SEC control, C518, for 0 on the vtum.
c. Disable the afc by pressing the AFC DISABLE switch and adjust the OSC FREQ control until the station monitor indicates the exciter is on frequency. Replace Q509.
d. Remove Q607. Place a vtvm between TP602 and ground. Tune L606 and L603 for a maximum indication on the vtum. Tune the inductors in the order shown to minimize the effects of the limiter.
e. Remove afc by holding down the AFC DISABLE switch. With the vtvm from TP602 and ground, tune the DISCR PRI control, C639, for a maximum indication on the vtvm. Move the vtvm to TP601 and tune the DISCR SEC control, C644, for a 0 indication when the modulation monitor indicates the exciter is approximately on frequency. Repeat the tuning of the DISCR PRI and DISCR SEC controls. Replace Q607.
f. Place a d-c vtvm between TP603 and ground. Set the AMP BIAS control for a 7.5 -volt indication on the vtym.
g. Place a vtvm between TP504 and ground. Remove Q510. Tune L611 and L608 for a maximum indication on the vtvm. Replace Q510.
h. With a vtvm on TP504, adjust the REF LEVEL control for an equal indication on the vivm with first Q510 removed and then Q509 removed. (This equalizes the modulator oscillator voltage and the $14-\mathrm{mc}$ reference voltage.) Replace the transistors.
i. Place an oscilloscope between TP504 and ground. Apply a. $150-\mathrm{cps}$ audio signal on J601. Adjust the MOD BAL control for a minimum $150-\mathrm{cps}$ indication as shown on the oscilloscope.

## NOTE

The MOD BAL control must be adjusted slowly to allow the error signal amplifiers to stabilize between adjustments.


Figure 5-1. Distortion Test Setup

TABLE 5-1. ABBREVIATED TUNING PROCEDURES

|  | CONTROL | POSITION | ADJUSTMENT | INDICATING METER | INDICATION | NOTES <br> Allow transmitter to warm up at least 15 minutes before tuning. |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{aligned} & \text { cy } \\ & \text { y } \\ & \text { y } \\ & \text { x } \end{aligned}$ | S101 | MIXER GRID A | BUF TUNE | M101 | Maximum |  |
|  | S101 | V428 B | *L429, L430 | M101 | Maximum |  |
|  | S101 | V429 B | *L431, L432 | M101 | Maximum |  |
|  | S101 | V430B | *L433, L434 | M101 | Minimum |  |
|  | S101 | V430C B | PA (exciter) PLATE | M101 | Minimum |  |
|  | MULTIMETER | GRID FS 40 MA | PA MATCH GRID TUNING | MULTIMETER | Maximum |  |
| $\begin{aligned} & \text { 品 } \\ & \text { 足 } \\ & \text { ( } \end{aligned}$ | WATTMETER | FORWARD | PLATE TUNING <br> output COUPLING <br> POWER OUTPUT ADJUST | P.A. PLATE CURRENT R.F. WATTMETER R.F. WATTMETER | Minimum <br> Maximum <br> 250 watts or less depending on drive requirements | Repeat the adjustment of PLATE TUNING, OUTPUT COUPLING, and POWER OUTPUT ADJUST controls until proper power amplifier drive is obtained. |
|  | WATTMETER | FORWARD | PLATE TUNING <br> OUTPUT <br> COUPLING <br> POWER OUTPUT ADJUST | R.F. WATTMETER <br> R.F. WATTMETER <br> R.F. WATTMETER | Maximum <br> Authorized power indication <br> Authorized power indication | Repeat the adjustment of PLATE TUNING, OUTPUT COUPLING, and POWER OUTPUT ADJUST controls until authorized power is achieved by the indirect power measuring method. $P=I_{p} E_{p} K$ |
| *Use slotted nonmetallic screwdriver on these adjustments. |  |  |  |  |  |  |

TABLE 5-2. DISTORTION CHECKS

| FREQUENCY | DISTORTION IN PERCENT |  |  |
| :---: | :---: | :---: | :---: |
|  | $25 \%$ MODULATION | $50 \%$ MODULATION | $100 \%$ MODULATION |
| 50 |  |  | . |
| 100 |  |  |  |
| 400 |  |  |  |
| 1000 |  |  |  |
| 5000 |  |  |  |
| 7500 |  |  |  |
| 10,000 |  |  |  |
| 15,000 |  |  |  |

For a complete exciter alignment procedure, refer to the maintenance section in TD-536.

### 5.4. Distortion Testing Procedure.

a. Refer to figure 5-1. Connect an audio frequency signal generator, such as a Hewlett-Packard Model 600 D to the exciter audio input, terminals 1 and 2 of TB305. (Disconnect the station console audio input leads when making this connection.) Connect a distortion and noise meter, such as a Hewlett-Packard Model 330D, to the broadcast monitor. Connect a 50 -ohm artificial load to the r-f output connector located on top of the transmitter cabinet. Turn on the transmitter.
b. Apply a $50-\mathrm{cps}$ audio tone to the transmitter input. The input level should be such that the voltage at the transmitter audio input terminals is $+10 \pm 2 \mathrm{dbm}$.
c. Adjust the exciter MOD GAIN control until the monitor inclicates 100 percent modulation ( $\pm 75-\mathrm{kc}$ deviation).
d. Measure the distortion at the frequencies and modulation levels given in table 5-2. The distortion shall be less than 1.0 percent for frequencies between 50 cps and 15 kc .

### 5.5 Audio Frequency Response Measurements.

a. Refer to figure 5-2. Connect an audio frequency signal generator, such as a Hewlett-Packard Model


Figure 5-2. Audio Frequency Response, Test Setup


Figure 5-3. Audio Frequency Response Limits

600D, to terminals 1 and 2 of terminal board TB305. (Disconnect the station console audio input leads when making these measurements.) Connect a vacuum-tube voltmeter, such as a Ballantine Model 310A, to the audio output terminals of the audio frequency generator. Connect a 50 -ohm artificial load to the r-f output connector located on top of the transmitter cabinet. Turn on the transmitter.
b. Check the audio frequency response of the transmitter by modulating the transmitter at $50,100,400$, $1000,5000,7500,10,000$, and 15,000 cps for 25 percent, 50 percent, and 100 percent modulation. Audio frequency response is measured by keeping the percentage of modulation constant and measuring the magnitude of audio, at each frequency given, to give the desired percentage of modulation. The audio frequency response must fall within the limits given in figure 5-3.

## NOTE

When taking audio frequency response measurements, a broadcast monitor, such as Hewlett-Packard Model 335B, should be used. Do not use an instrument where audio deemphasis might give a false indication of peak modulation.

### 5.6 FM Noise Measurements.

a. Refer to figure 5-4. Connect an audio frequency signal generator, such as a Hewlett-Packard Model

600D, to terminals 1 and 2 of terminal board TB305. (Disconnect the station console audio input leads when making this measurement.) Connect a vacuum-tube voltmeter to the output terminals of the broadcast monitor. Connect an artificial load to the r-f output connector located on top of the power amplifier cabinet. Turn on the transmitter.
b. Modulate the transmitter 100 percent ( $\pm 75-\mathrm{kc}$ deviation) with 400 cps of audio.
c. Remove the modulating 400 cps and read the residual $F M$ noise on the vacuum-tube voltmeter. The residual $F M$ noise shall be less than -65 db below 100 percent modulation.

### 5.7 AM Noise Mcasurements.

a. Refer to figure 5-5. Short out terminals 1 and 2 of terminal board TB305. Connect a vacuum-tube voltmeter to connector J3 of the Hewlett-Packard Model 335B broadcast monitor. Connect a 50 -ohm artificial load to the r-f output connector located on top of the power amplifier cabinet. Turn on the transmitter.
b. Switch the broadcast monitor to measure carrier level.
c. Measure the $A M$ noise in db at J 3 of the broadcast monitor in the following manner. Set modulation monitor to carrier level and measure the d-clevel on the modulation meter ( 100 percent on scale equals 10 volts). Connect the vacuum-tube voltmeter to J3 and terminate J3 with a 2 -megohm resistor. Measure the a-c level on the vacuum-tube voltmeter. (The


Figure 5-4. FM Noise Test Setup


Figure 5-5. AM Noise Test Setup
input to the vacuum-tube voltmeter should be a shielded cable having less than 100 uuf distributed capacitance.) The AM noise is related to the ratio of the d-c reading and the a-c level. The AM noise shall be not less than -55 db below voltage or $\mathrm{d}-\mathrm{c}$ carrier level.

$$
\text { AM noise in } \mathrm{db}=20 \log \frac{\mathrm{~d}-\mathrm{c} \text { reading }}{\mathrm{a}-\mathrm{c} \text { reading }}
$$

### 5.8 Tronble Shooting.

Standard trouble-shooting procedures should be used in finding malfunctions in the transmitter. As suggested
in TD-536, TD-538, and TD-580, meter indications for all functions should be recorded when the transmitter is installed and operating properly. Table 5-3 is supplied for recording these readings. If some malfunction should occur after the normal meter readings are recorded, it is a simple matter to compare the meter readings of the malfunctioning equipment with the normal meter readings. When trouble-shooting and comparing the meter readings it is advisable to start with the final stage and proceed backwards until normal readings are encountered. The malfunctioning stage will then be the one immediately ahead of normal meter indications.

As most cases of trouble will be traced to tubes or transistors, it is advisable to first of all replace the tube (or transistor) in the stage in which the trouble is suspected. If the trouble does not clear with tube or transistor replacement, it will become necessary to take resistance or voltage measurements within the suspected circuit to determine which component has failed.

When tracing trouble within the power amplifier it will be helpful to use the "from-to"' information given in unit instructions TD-538 and TD-580. The "fromto" information gives the actual location of individual wires within the transmitter cabinets. When
used in conjunction with the schematics, the "fromto" information can be very helpful.

If the transmitter center frequency shifts excessively with modulation, the trouble may be isolated to either afc circuitry or the modulator circuitry of the exciter by disabling the afc and noting if the carrier shifts more than 1.8 kc with a change in modulation from 0 to 100 percent. If the modulator oscillator shifts more than 1.8 kc with the afc disabled, the trouble will be within the modulator oscillator circuits. The afc circuitry cannot shift the modulator oscillator frequency more than 1.8 kc . If the carrier shift is under 1.8 kc , the trouble will be in the afc circuitry.

TABLE 5-3. NORMAL TRANSMITTER METER INDICATIONS

|  | CONTROL | POSITION |  |
| :--- | :--- | :--- | :--- |

0

# unit instructions 

## F830-1 <br> 10-Kw FM Power Amplifier

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## general description

### 1.1 Purpose of Instruction Book.

This unit instructions provides information about F830-1 $10-\mathrm{Kw}$ FM Power Amplifier. Information which is furnished covers a general description of the equipment, principles of operation, maintenance procedures, and a parts list.


Figure 1-1. F830-1 10-Kw FM Power Amplifier, Over-all View

### 1.2 Purpose of Equipment.

The F830-1 10-Kw FM Power Amplifier is used for continuous monaural or stereophonic FM broadcast service on a single frequency in the range from 88 to 108 megacycles with an output power of 10,000 watts.

## I. 3 Description of Equipment.

### 1.3.1 PHYSICAL DESCRIPTION.

The F830-1 10-Kw FM Power Amplifier, shown in figure $1-1$, is contained in a single cabinet that is 38 inches wide, 76 inches high, 27 inches deep, and weighs approximately 1240 pounds. All power amplifier operating controls are located behind the doors on the front of the cabinet. The filament and plate on-off controls and four monitoring meters are located at the top front of the cabinet. The screen circuit breaker is located inside the rear of the cabinet. The monitoring meters may be observed easily whileoperating the tuning controls. The power amplifier uses one r-famplifier tube and six rectifier tubes. Silicon diode rectifiers are an optional item used in place of the six rectifier tubes. The r-f amplifier tube is accessible from the front of the power amplifier. The bottom front of the power amplifier cabinet is removable to allow access to components on the bottom of the inside panel.
Large doors at the upper rear of the cabinet (see figure $1-2$ ) allow access to the upper part of the power amplifier for servicing and maintenance. The lower rear half of the power amplifier cabinet is covered by a removable panel containing a ventilating fan and a permanent air filter. Operating personnel are protected by both electrical and mechanical interlocks on the rear doors and panel. These interlocks remove the plate voltage and ground the high-voltage circuits when the doors are opened or disable the high voltage when the panel is removed. The power amplifier plate tuning and grid tuning resonators are located in an interlocked compartment at the front of the transmitter.

Inside the F830-1, heavy iron-core components are at the bottom of the cabinet. A harmonic filter, attached to $\mathrm{F} 830-1$, is located in the rear cabinet compartment.

Cooling air for the power amplifier is drawn through a permanent air filter at rear of cabinet by a highvolume fan and exhausted through a shielded opening at top of cabinet. A single high-volume blower supplies cooling air directly to the power amplifier tube.

### 1.3.2 ELECTRICAL DESCRIPTION.

The F830-1 $10-\mathrm{Kw}$ FM Power Amplifier consists of a single, air-cooled power amplifier tube and all


Figure 1-2. F830-1 Rear View with Doors Gpen and Lower Panel Removed
associated power supply and control circuitry. The F830-1 input impedance is 50 ohms nominal, unbalanced. The F830-1 output power is at least 10,000 watts over the frequency range of 88 to 108 megacycles into a 50 -ohm load with an swr not exceeding 2:1.

Line power input required is 60 cycle, 3 phase with primary taps onall power transformers to compensate for local line voltage variations from 200 to 250 volts. Other taps are available for reduced power operation. Circuit breakers in the input side of the plate, screen, and control circuits are provided for primary current overload protection. The control circuit auxiliary power supply and the control grid bias supply are fused. Time delay circuitry for protection of the power amplifier during warmup is provided, with the actual time delay control received from the driver time delay relay. Circuits are provided for remote control tie-in with the driver remote control circuits, with the actual remote control available from one source. Instruction
books covering the exciter and driver used in conjunction with the F830-1 $10-\mathrm{Kw}$ FM Power Amplifier are listed in table 1-1.

TABLE 1-1
ASSOCIATED EQUIPMENT INSTRUCTION BOOKS

| ASSOCIATED EQUIPMENT | INSTRUCTION BOOK <br> PART NO. |
| :--- | :---: |
| A830-2 10 W Wide-Band <br> FM Broadcast Exciter | $523-0755303$ |
| $786 \mathrm{M}-1$ Stereo Generator | $523-0755304$ |
| B830-1 250-Watt FM | $523-0755596$ |
| Power Amplifier |  |
| D830-1 1000-Watt FM | $523-0755334$ |
| Power Amplifier |  |

### 1.4 Equipment Supplied.

Table 1-2 lists equipment supplied as part of the F830-1 10-Kw FM Power Amplifier.

TABLE 1-2
EQUIPMENT SUPPLIED

| EQUIPMENT | COLLINS <br> PART NUMBER |
| :---: | :---: |
| F830-1 10-Kw FM Power <br> Amplifier <br> $5 / 10-K w ~ H a r m o n i c ~ F i l t e r ~$ | $522-2981-00$ |

### 1.5 Equipment Required but not Supplied.

Table 1-3 lists equipment required for operation of F830-1 10-Kw FM Power Amplifier but not supplied as part of the power amplifier. The 250 -watt poweramplifier is used in place of the 1000 -watt power amplifier for reduced power applications.

TABLE 1-3
EQUIPMENT REQUIRED BUT NOT SUPPLIED

| EQUIPMENT | COLLINS <br> PART NUMBER |
| :---: | :---: |
| A830-2 10 W Wide-Band <br> FM Broadcast Exciter <br> B830-1 250-Watt FM <br> Power Amplifier <br> or | $522-2714-00$ |
| D830-1 1000-Watt FM <br> Power Amplifier | $549-2008-00$ |

### 1.6 Equipment Specifications.

### 1.6.1 MECHANICAL.



Ambient humidity
range . . . . . . . . 0 to 95 percent relative humidity.

Altitude . . . . . . 0 to 6000 feet.

### 1.6.2 ELECTRICAL.

Power source. . . . . 200 to 250 volts, 60 cycle, 3 phase.
Maximum 60-cps
power requirements . . 20 kilowatts.
R-f input power . . . . 250 watts nominal.
Power output . . . . . 10,000 watts nominal or at optional reduced power.
Output impedance . . . 50 ohms, nominal unbalanced.

Frequency range . . . 88 to 108 megacycles. Exact operating frequency determined by frequency of exciter.

### 1.7 Tube and Semiconductor Complement.

Table 1-4 lists the tube and semiconductor complement supplied as part of the F830-1 10-Kw FM Power Amplifier.

TABLE 1-4. TUBE AND SEMICONDUCTOR COMPLEMENT

| QUANTITY | FUNCTION |  |
| :---: | :--- | :--- |
| 6 | TYPE | Plate voltage rectifiers or |
| 72 | $66-6726$ | Plate voltage silicon rectifiers |
| 1 | $4 C X 5000 \mathrm{~A}$ | Power amplifier |
| 8 | $1 N 540$ | Control grid bias rectifiers |
| 4 | $1 N 540$ | Multimeter rectifiers |
| 12 | $66-6726$ | Screen grid voltage rectifiers |
| 4 | $1 N 3044 A$ | Zener remote control line protection |
| 2 | $1 N 3016 \mathrm{~B}$ | Zener remote control line protection |

## section

### 2.1 General.

The F830-1 10-Kw FM Power Amplifier contains a power amplifier and associated circuitry for the amplification of approximately 250 watts of r-f drive up to 10 kilowatts of $\mathrm{r}-\mathrm{f}$ power. The $\mathrm{F} 830-1$ operates in the frequency range of 88 to 108 megacycles. The F830-1 output can be fed through a harmonic filter, for the attenuation of spurious radiations, to an antenna or to a higher power amplifier.

Refer to figure 2-1, a block diagram of the F830-1 $10-\mathrm{Kw}$ FM Power Amplifier. The 230 -volt, 60 -cps, 3 -phase line input is fed to the plate and screen stepstart contactors where plate and screen voltage on-off functions are controlled by the control circuits. The 230-volt, 3-phase line input is also fed to an auxiliary power supply, T401, where part of the 230 -volt input is reduced to 115 volts single phase. The single phase output of T401 is then fed to the control circuits. The control circuits turn the plate, screen, filament, and control grid bias supply on and off. The control circuits also feed 230 -volt, $60-\mathrm{cps}$, single phase power to the driver. Provisions are made within the control
circuitry to connect to the filament on-off and plate onoff functions of a higher power amplifier, if one is used. The higher power amplifier could then control the filament on-off and plate on-off functions.

The power amplifier consists of a single, ceramic type, forced air-cooled, grounded screentube working into a resonant, quarter wavelength, coaxial line. The control grid circuit consists of a coaxial line resonator with a swamping resistor in parallel to provide a low impedance broadband load to the controlgrid. Metering circuits are included to measure filament voltage, control grid bias voltage, screen grid voltage, control grid current, and screen current.

The output power of the F830-1 may be reduced by lowering the plate, screen, and control grid voltages.

### 2.1.1 CONTROL CIRCUITS.

Refer to figure 2-2, a simplified schematic diagram of the control circuits of F830-1 10-Kw FM Power Amplifier. The control circuits apply or remove filament, plate, screen, and grid bias voltages, turn the cabinet fan and power amplifier blower on andoff, and provide


Figure 2-1, F830-1 10-Kw FM Power Amplifier, Block Diagram


overload protection for all power amplifier circuits. These functions are discussed in the following paragraphs.

The 230 -volt 3 -phase power enters the cabinet on TB401 and is fed to two circuit breakers, CB401 and CB402. The PLATE circuit breaker, CB402, protects only the plate and screen supplies, while the filament circuit breaker, CB401, protects the remaining power amplifier circuitry. The 3 -phase output from CB402 is fed to two contactors to provide a voltage stepping action when plate and screen power is applied. This stepping action reduces the transients generated in the plate and screen power supplies when the $\mathrm{F} 830-1$ is turned on or off. Circuit breaker CB403 is placed in the screen supply 3-phase input to protect the screen power supply and to provide an aid whentrouble shooting the transmitter.

The 3-phase output from filament circuit breaker CB401 is split into single phase components and fed to the filament contactor, the blower relay, the auxiliary power supply T 401 , and the driver.

Power to operate the filament and blower relays is obtained from the 115 -volt secondary of T401. A CONTROL indicator, DS403, is lighted when the 115volt power is present at the secondary of T401. This power is fed through the FILAMENTOFF switch to the 250 -watt driver control circuitry at TB403-3. The driver control circuitry can interrupt power at this point if filament-off functions are to be controlled from the driver. Filament and blower control power is routed through the driver and re-enters the F830-1 at TB403-8. The power thenis fed directly to the filament and blower control relays.

If FILAMENT ON switch S112 is pressed on either the driver or the F830-1, a ground is placed on blower relay K401 causing the relay to close. This action starts the power amplifier blower, B401, and closes contact K401-7 and K401-8. Closing of these contacts holds a ground on K401 when the momentary FILAMENT ON switch is released. When PA blower B401 comes up to speed, air pressure activated switch S 407 closes, causing filament relay K 402 to energize. The green filament-on light. DS401, will come on; the cabinet blower will start; and filament power will be fed to the tubes. The control grid bias supply will also be energized from the primary of transformer T401.

Power for operation of the plate circuits is obtained from the same source as the filament power transformer, T401. To start the plate-on sequence, PLATE ON switch S 113 is pressed actuating plate hold relay K403. This closes contacts 10 and 12 to hold K403 energized when the PLATE ON switch is released. The ground is obtained from the driver control circuitry to allow the driver to interrupt the ground for the plate-off function. Contacts 7 and 9 will also close, turning on the driver plate power if driver plate power has not been turned on previously. With K 403 energized, contacts 1 and 3 will make, transferring the
control voltage to time delay transfer relay K404. K404 will actuate when the driver time delay relay has completed the cycle allowing the filaments to warm up to operating temperature. With K404 energized, 115 volts will be transferred to the plate-on contactor, K405, causing K405 to close and apply power to the plate and screen power supplies. A fraction of a second later (as soon as K405-11 and 12 close), the step-start contactor will actuate and apply 230 volts 3 phase to the plate and screen power supplies. The red plate-on indicator, DS402, will light, indicating that plate and screen power is applied to the power amplifier. This stepping action limits transients that may be introduced into the power supply by the initial application of power.

If the PLATE OFF switch, S114, is pressed, plate and screen voltage will be removed from the power amplifier only and will not normally affect the driver. Pressing the PLATE OFF switch, S114, opens the ground to plate-hold relay K403, releasing K403. Stepstart contactor K 406 will then release removing direct 3 -phase power from the plate and screen power supplies and momentarily throwing dropping resistors R401, R402, and R403 into the plate-screen circuits. Moments after step-start contactor K 406 opens, plateon contactor K405 will open, removing all power from the plate and screen power supplies. This stepping action limits transients that would normally be introduced into the power supply by the sudden removal of power.

If FILAMENT OFF switch S111 is pressed, power will be removed momentarily from all circuits within the driver and power amplifier, causing the holding relays to drop out and shut off the power amplifier and driver.

### 2.1.2 POWER AMPLIFIER CIRCUITS.

Refer to figure 2-3, a simplified schematic of the power amplifier circuitry of F830-1 10-Kw FM Power Amplifier. The power amplifier consists of a single ceramic type, forced air-cooled tube working into a tuned coaxial line resonator over the standard frequency modulated broadcast band of 88 to 108 megacycles.

Power amplifier V407 is a fixed bias, class C, grounded screen tetrode. The input from the driver is fed through blocking capacitor C452 into a tuned coaxial
line resonator (a foreshortened $\frac{\lambda}{4}$ transmission line) and fed to the control grid. Resistor R468 in the grid circuit swamps out much of the driver power which is fed into the tuned coaxial line resonator and, in addition, presents a low impedance broadband load to the control grid. Initial tuning of the grid coaxial line resonator is accomplished by a slider which physically lengthens or shortens the grid coaxial line resonator. Fine grid tuning is accomplished by adjusting variable capacitor C436. Fixed capacitors C437, C438, C439, C440, C441, and C442 in parallel with C 436 couple the input tuned circuit to the grid. As the power amplifier is a grounded screen amplifier, the filament must be below ground potential. The grid bias supply is floating and is connected in series with the screen grid
power supply giving negative control grid bias with respect to the filament. L405 blocks r-ffrom the grid bias supply.

The plate works into a tuned coaxial line resonator (foreshortened $\frac{\lambda}{4}$ transmission line) similar to the grid resonator. The plate resonator consists of a short piece of coaxial transmission line which resonates with the plate capacity of V407 and PLATE TUNING capacitor C445. Initial tuning is similar to the grid resonator, and is accomplished by a shorting slider on the transmission line which physically lengthens or shortens the transmission line. Parasitic suppressors located within the resonator (C449, R474, L410, R484, and L411) suppress resonances around 200 and 400 megacycles. The parasitic suppressor capacity is formed by the distributed capacity of resistors R747 and R484 and the resonator walls.
R-f output coupling is achieved by a movable plate within the resonator forming a capacitor between the center conductor of the coaxial transmission line and the movable metal plate. This capacitor is C444. R-f output from C444 is then fed to a directional coupler if a higher power amplifieris driven, or to a harmonic filter and through the directional coupler if an antenna is fed directly. Suppressor L408, R471, and C446 effectively damp the higher order resonances of the plate tank circuit.

As stated previously, power amplifier V 407 operates as a grounded screen amplifier. Actually, the screen is slightly above r-f ground to provide screen neutralization of the tube. Neutralization is accomplished by balancing the capacitance bridge formed by the plate-to-control grid capacitance, the plate-to-screen grid capacitance, and the screen-to-control grid capacitance. This is accomplished by adding a small variable inductance, L406, to form an r-f voltage from the screen-to-control grid 180 degrees out of phase with the plate-to-control grid r-f. This additional impedance balances the bridge to neutralize the tube for a fixed operating frequency.
Plate voltage is obtained from a conventional 3-phase power supply, which, when connected inseries with the screen power supply, forms the plate voltage. The power supply transformer is connected in a delta primary and a wye secondary configuration. Six mercury-vapor rectifier tubes, V401 through V406, are arranged in a 3 -phase full-wave bridge rectifier circuit. Silicon diode rectifiers are optional. A P.A. PLATE CURRENT meter, M402, is placed in the ground side of the plate supply to indicate only plate current. As the screen supply is inseries with the plate supply, the P.A. PLATE VOLTAGE meter, M403, is placed across both the screen and plate supplies to indicate the cathode-to-plate voltage. The plate supply output is approximately 5700 volts d-c at 2.5 amperes. When the plate supply voltage is added to the screen supply voltage, the total, 6450 volts, equals the total plate voltage.
Screen voltage is obtained from a conventional 3-phase semiconductor power supply. A transient suppressor,

R506, R507, and C411, reduces transients formed when the power supply is turned on or off. The screen supply output is approximately 750 volts at 2.6 amperes.
The control grid bias supply is a single-phase, semiconductor power supply. A resistor and capacitor placed in parallel with each diode equalize currents and minimize transients in each diode. A transient suppressor, C412 and R430, reduces transients when power is turned on or off. The positive end of the grid bias supply is connected to the fixed side of POWER OUTPUT ADJUST potentiometer R405. This permits R405 to control both the bias voltage and screen voltage when R405 is adjusted. The negative end of the bias supply is tied into the control grid through a meter shunt resistor and a filtering network. The control grid bias supply output is approximately -300 volts at 300 ma . Bias voltage is made adjustable by R417.
The filament of power amplifier $V 407$ is below d-c ground because of the grounded screen configuration. The filament-to-ground potential will then be the screen voltage. Capacitors placed across the filament, and from each side of the filament to ground, provide r-f grounding.
Metering circuits are provided to measure the filament voltage, control grid bias voltage, control grid current, screen current, and screen grid voltage. In addition, an external metering position is available for connection to the multimeter. The customer may employ this extra meter position for any use that he may desire. The external meter movement input is $0-1$ ma at 100 ohms.
The filament voltage metering circuit employs a fullwave bridge rectifier to change the a-c filament voltage to d-c. The bridge output is filtered by R418 and C410, and the complete circuit is calibrated by R415. The grid bias metering circuit is a voltage measuring device which measures the control grid bias voltage directly through the use of meter multiplier resistor R416. Grid current is metered by shunt resistors R428 and R492. Screen current is measured by placing the MULTIMETER across shunt resistors R413 and R414. As the screen is grounded, only screen current will flow through these resistors. Screen voltage is measured directly by the use of meter multiplying resistor R495. External meter readings may be made by connecting to the properterminals onTB405 (not shown on simplified schematic diagram, figure 2-3). (See figure 5-1.) The MULTIMETER is shunted with capacitor C450 to prevent any stray r-f from damaging the meter movement.

### 2.2 Control Functions.

The following paragraphs describe the functions of all controls in F830-1 10-Kw FM Power Amplifier. Refer to figure 2-4 for control locations.

The controls located on the front panel directly under the meters include the FILAMENT ON, FILAMENT OFF, PLATE OFF, and PLATE ON controls. The green filament on indicator and red plate on indicator are placed in line with the above-mentioned controls.


The FILAMENT ON switch, S112, energizes the filaments, the blowers, and the bias supply and will turn on the driver filament and the exciter if connected correctly. The FILAMENT OFF switch, S111, deenergizes all transmitter circuits. The PLATE ON switch, S113, energizes the plate and screen power supplies and the driver plate supply. The PLATEOFF switch, S114, removes plate and screen voltage. The green filament indicator light, DS401, lights when the FILAMENT ON switch is pressed and the PA blower has activated the PA blower interlock. DS 401 indicates that voltage is available to the filament control contactor. The filament control contactor starts the cabinet blower and the bias supply and supplies the necessary voltage to the $\mathrm{F} 830-1$ filaments. The red plate-on indicator light, DS402, indicates plate voltage has been applied to the power amplifier. The following controls are located directly under the right front door on the power amplifier panel. The POWER OUTPUT ADJUST potentiometer, R405, adjusts the power amplifier screen potential thus changing the output power. The FILAMENT VOLTAGE ADJUST potentiometer, R404, adjusts the filament transformer input voltage thus changing the filament voltage of V407. The WATTMETER switch, S406, connects the R. F. WATT METER to either the reflected power or forward power section of the directional coupler. The WATTMETER switch is normally left in the FORWARD position. In addition, the WATTMETER switch has a third position for connecting both forward and reflected output power voltages from the directional coupler to remote lines. The MULTIMETER switch, S405, selects either filament voltage, bias voltage, control grid current, screen current, or screen voltage. In addition, an external position is available for use by the customer. Table 3-1 lists the MULTIMETER switch positions and typical indications for each of the four meter positions.
The following controls are located directly behind the left front door on the power amplifier resonator. The PLATE TUNING control, C445, tunes the plate to resonance and is set near the minimum indication on the P.A. PLATE CURRENT meter, M402. At this
point, the power output should be at the peak as indicated on R.F. WATTMETER M404. The OUTPUT COUPLING control, C444, adjusts the coupling of the load to the plate coaxial line resonator center conductor. The GRID TUNING control, C436, tunes the grid resonator.

The following controls are located on the power panel directly behind the bottom front panel of the power amplifier cabinet. The PLATE circuit breaker. CB402, is a protective device which monitors the plate and screen supply transformer primary currents. The circuit breaker will open when the current exceeds 50 amperes. The CONTROL circuit breaker, CB401, monitors the total control circuit current. This current includes the power fed to the driver and exciter. The circuit breaker will open when the control circuit current exceeds 8 amperes. The 5 -ampere CONTROL CIRCUIT fuse, F401, protects the control circuits from overloads. The $1 / 4$-ampere BIAS VOLTAGE fuse, F402, protects the bias supply from overloads. Blower fuses, F403, F404, and F405 protect PA blower B401.

The following adjustments are located directly below the P.A. PLATE CURRENT and R.F. WATTMETER indicators when the upper switch and meter panel is raised. The left potentiometer, located behind the panel, is filament voltmeter calibrating potentiometer R415. The filament metering circuit calibrating adjustment is set at the factory and normally will not require adjustment. The center potentiometer is the overload adjusting potentiometer, R408. Theoverload adjustment normally is set for a plate current of 2.6 amperes. The right potentiometer is wattmeter adjusting resistor R419. This potentiometer, set at the factory, should not require adjustment. The control grid adjustable resistor, R417, is located in the left bottom rear of the power amplifier cabinet. The resistor is set for approximately 500 ma maximum of plate current with no drive to the power amplifier. The screen circuit breaker CB403, is located behind the left rear door, and is attached to the rectifier panel. The circuit breaker protects the screen power supply from overloads above 10 amperes.

## maintenance

### 3.1 General.

This section contains information concerning the maintenance of F830-1 10-Kw FM Power Amplifier.

## WARNING <br> 

Voltages present in this equipment aredangerous to life. Observe safety precautions when performing any maintenance. Do not reach inside the F830-1 cabinet when high voltage is applied. Do not depend entirely on door interlocks. Always shut down the F830-1 before doing any work inside the F830-1 cabinet. Im mediately upon opening the rear doors, short out all high-voltage points using the shorting stick located inside the left rear door.

### 3.2 Preventive Maintenance.

Most service interruptions in equipment of this type are caused by dirt and corrosion. Corrosion is accelerated by the presence of moisture and dust. Dust should be removed periodically with a soft brush or a dry, oil-free air jet. Remove dust as often as a perceptible quantity accumulates at any point in the power amplifier.

When the $\mathrm{F} 830-1$ is operated near salt water or in other corrosive atmospheres, inspect and clean interlock switches, cable connectors, tube prongs, and other metal parts more frequently to keep the equipment in top operating condition.

### 3.2.1 AIR FILTER CLEANING.

At least once each month, or more often if needed, clean the air filter according to the following procedure:
a. Remove the air filter from the F830-1 cabinet by loosening the two thumb screws located above the air filter. Slide the air filter to the extreme right, and pull the left side of the air filter out as soon as the filter clears the panel. Slide the air filter to the left and remove.
b. Mark with an arrow the direction of the airflow.
c. Wash by passing a fine spray of hot water through the filter in the direction opposite that of the airflow. Gently shake the water out of the filter.
d. Dip the filter in water-soluble oil, such as Filterkote M (Collins part number 005-0609-00), available from Collins Radio Company Service Parts Department, Cedar Rapids, Iowa.
e. Remove the filter from the oil; lay the filter face down until oil ceases to drip from the filter.
f. Replace the filter into the lower rear panel with the airflow arrows (marked when the filter was removed) pointing in the direction of the airflow. Tighten the two thumb screws.
g. Replacement filters are Collins part number 009-1069-00.

### 3.2.2 PA TUBE CLEANING.

The power amplifier tube depends upon a stream of air passing through the fins to cool the anode. When these fins become dirty, the airflow is reduced, and the tube life is shortened. The radiator fins should be cleaned as follows:
a. Remove the r-f amplifier tube as described in paragraph 3.2.2.1.

## CAUTION

Special care must be used in removing or installing the power amplifier tube.
b. Direct a low-pressure ( 50 psi ) air stream through the fins in the direction opposite to the normal airflow until all dust is removed.
c. Replace the r-f amplifier tube as described in paragraph 3.2.2.1.d.

### 3.2.2.1 PA TUBE REMOVAL.

## WARNING

Voltages present within the plate compartment are dangerous to life. Shut down the F830-1 before doing any work inside the compartment.

The power tube may be removed as follows:
a. Open the power amplifier resonator door.
b. Loosen the slider clamp, and raise the center conductor straight up until it clears the tube completely. Make sure the slider stays in its original position while raising the center conductor.
c. While holding up the center conductor, grasp the tube handle and remove the tube from its socket. It may be necessary to push the tube partially into the center conductor so the tube will clear the tube socket shield before complete tube removal.
d. Tube replacement is the reverse of removal. After tube replacement, check the plate slider for proper distance from the deck plate for the station frequency. See System Instructions, Frequency Change, for this approximate distance. Be sure the tube is seated firmly in the tube socket before replacing the center transmission line.

### 3.2.3 INSPECTION.

Once a week, check and clean the three interlock switches at the rear of the F830-1 cabinet to be sure they are in good working order. Once each month, check all connections in the F830-1. Tighten any nuts, bolts, or screws that may be loose. Check cable connections to see that they are clean and mechanically secure. Check moving parts, such as tuning controls, for excessive wear. Check the plate and grid sliders for oxidation around ground springs.
Silicon rectifier diodes should be checked for shorts approximately every six months. To check for diode shorts, place a reverse voltage from an ohmmeter across each diode. If any diode reads shorted, replace the diode. Normally, a high resistance will be noticed across each diode caused by the diode parallel resistor or by diode leakage.

### 3.2.4 LUBRICATION.

The PA blower has bearings that are lubricated for the life of the equipment. Lubricate each of the cabinet
fan bearings every 6 months with 3 drops of SAE \#20 nondetergent oil. The rear cabinet fan bearing oil hole may be reached through the air filter opening, with the air filter removed.

### 3.2.5 TUBE MAINTENANCE.

Power amplifier V401 should be inspected (tube in place) once each week to ensure that an accumulation of dust does not build up on the radiator fins. If dust is present, clean as described in paragraph 3.2.2. When tuning the $\mathrm{F} 830-1$, care should be taken not to exceed the maximum plate current shown in table 3-1.

If mercury-vapor tubes are used in the plate power supply, spare, preaged mercury-vapor rectifier tubes should be available for immediate replacement. To ready these tubes for emergency use, place them in the power amplifier during off-the-air hours, and run them for twenty minutes with only the filaments lighted. This will remove the mercury coating from the tube elements. Then carefully remove the tubes from the F830-1, and store them in an upright position where they will not be inverted or agitated. Whenthese preaged tubes are placed in the $\mathrm{F} 830-1$, handle them carefully to avoid the twenty-minute warmup period that will be required if mercury comes in contact with the tube elements. Never apply plate voltageto mercuryvapor rectifier tubes that have not been aged long enough to remove all mercury from the tube elements.

### 3.3 Trouble Shooting.

The most common cause of trouble will probably be traced to tube failure. If a tube is suspected of failure,

TABLE 3-1. TYPICAL METER INDICATIONS

| METER | METER SWITCH POSITION | INDICATION |
| :--- | :--- | :--- |
| MULTIMETER | FIL V 8V FS | 7.5 volts |
| MULTIMETER | BIAS V 400V FS | $180-220$ volts |
| MULTIMETER | GRID 80MA FS | $45-60$ ma |
| MULTIMETER | SCREEN 400MA FS | $150-225$ ma |
| MULTIMETER | SCREEN 800V FS | 750 volts |
| PA PLATE VOLTAGE |  | $6400-6500$ volts |
| PA PLATE CURRENT | FORWARD 15KW | Approx. 2.25 A |
| R.F. WATTMETER | REFLECTED 5KW | 10 kw |
| R.F. WATTMETER |  | Less than 500 watts |

replace it with a tube of known quality, and note any change in performance. A small loss in emission of V407 can be compensated by a change in the setting of the POWER OUTPUT ADJUST potentiometer.

Four meters are located on the F830-1 front panel to assist in locating any trouble which may occur. Table 3-1 contains typical meter indications. These average indications are obtained from several production power amplifiers. The indications of certain $\mathrm{F} 830-1$ may vary slightly outside the given limits without affecting power amplifier performance. A list of panel meter indications for each individual power amplifier should be taken when the F830-1 is operating properly in its particular installation. Any abnormal deviation from
these values will be apparent during a check of meter indications.

### 3.4 Cable Chart.

Table 3-2 contains from-to information for cables installed in F830-1 10-Kw FM Power Amplifier. The table is useful in locating point-to-point wiring within the F830-1 cabinet. The from column is listed in alphabetical and numericalorder. To find a particular wire, establish the point on the F830-1 from which wire tracing is to be initiated. Find this point in the from column of table 3-2, and the to column will give the location of the other end of that particular wire. The wire information given in table 3-2 pertains only to those wires located in the main cable of the F830-1.

TABLE 3-2. CABINET FROM-TO INFORMATION

| FROM | TO | WIRE CODE | FROM | TO | WIRE CODE |
| :---: | :---: | :---: | :---: | :---: | :---: |
| B401-1 | K401-2 | RD90 | CB401-4 | T402-2 | RC 5 |
| B401-2 | K401-4 | RD92 | CB401-4 | TB402-2 | RE5 |
| B401-3 | K401-6 | R.D95 | CB401-4 | XF403-1 (top) | RD95 |
| C401-1 | L401-2 | VC905 | CB401-5 | K402-3 | RD92 |
| C401-2 | T402-4 | RB91 | CB401-5 | TB402-1 | RE2 |
| C401-2 | TB406-12 | VC902 | CB401-5 | XF404-1 (top) | RD92 |
| C402-1 | L402-2 | VC906 | CB401-6 | K402-1 | RD93 |
| C402-1 | R417-1 | VC92 | CB401-6 | T401-5 | RD93 |
| C402-2 | R431-2 | VC91 | CB401-6 | XF405-1 (top) | RD93 |
| C403-1 | R405-1 | VE92 | CB402-1 | CB401-3 | H9 |
| C403-1 | R412-1 | VE92 | CB402-1 | TB401-1 | н9 |
| C403-1 | S412-B | VE92 | CB402-2 | CB401-2 | H9 |
| C403-2 | R412-2 | RE92 | CB402-2 | TB401-2 | H9 |
| C404-1 | L404-2 | VE92 | CB402-3 | CB401-1 | H9 |
| C404-1 | R431-2 | VC95 | CB402-3 | TB401-3 | H9 |
| C408-2 | E407 | VE93 | CB402-4 | K405-3 | H9 |
| C408-2 | R411-2 | VE92 | CB402-5 | K405-5 | H9 |
| C427 | R415-3 | RC93 | CB402-6 | K405-7 | H9 |
| C428 | TB407-2 | VB93 | CB403-1 | K406-6 | RE5 |
| C429 | R404-2 | RC90 | CB403-2 | K406-4 | RE2 |
| C430 | K402-4 | RC92 | CB403-3 | K406-2 | RE90 |
| C431 | R405-2 | JE2 | CB403-4 | T404-15 | RE5 |
| C432 | S405-13 | VB91 | CB403-5 | T404-8 | RE2 |
| CB401-1 | CB402-3 | H9 | CB403-6 | T404-1 | RE90 |
| CB401-2 | CB402-2 | H9 | DC401 (red) | S406-11 | DAS2 |
| CB401-3 | CB402-1 | H9 | DC401 (yellow) | S406-4 | DAS9 |
| CB401-4 | T401-1 | RD95 | E401 (grd) | K401-8 | RC9 |

TABLE 3-2. CABINET FROM-TO INFORMATION (Cont)

| FROM | TO | WIRE CODE | FROM | TO | WIRE CODE |
| :---: | :---: | :---: | :---: | :---: | :---: |
| E401 (grd) | E402 (grd) | RE9 | K402-5 | T401-1 | RC923 |
| E401 (grd) | K403-6 | RC9 | K402-6 | J401-1 | RC95 |
| E401 (grd) | K405-2 | RC9 | K402-6 | T406-2 | RC95 |
| E401 (grd) | K406-10 | RC9 | K402-7 | T401-5 | RD90 |
| E401 (grd) | M404-1 (+) | RC9 | K402-8 | J401-2 | RC90 |
| E401 (grd) | S112-1 | RC9 | K402-8 | T406-1 | RC90 |
| E401 (grd) | S405-15 | VB9 | K402-8 | XF402-1 | RC90 |
| E401 (grd) | R413-2 | VE9 | K402-9 | K401-9 | RC935 |
| E401 (grd) | T401-9 | RC9 | K402-9 | K403-13 | RC91 |
| E402 (grd) | E401 (grd) | RE9 | K402-9 | TB404-5 | RC935 |
| E402 (grd) | TB402-4 | RE9 | K402-9 | XDS 401-1 | RC936 |
| E402 (grd) | TB403-10 | RE9 | K402-10 | S407-1 | RC936 |
| E407 | C408-2 | VE93 | K402-10 | TB407-12 | RC902 |
| E407 | M402-2(-) | VE92 | K403-1 | K406-7 | RC6 |
| E410 | TB407-3 | VC92 | K403-1 | XF401-1 | RD96 |
| E411 | S405-12 | VB93 | K403-3 | K406-8 | RC6 |
| E411 | TB407-1 | VB93 | K403-4 | K401-7 | RC93 |
| E418 | R405-2 | VB92 | K403-4 | S407-2 | RC93 |
| E422 (grd) | TB407-20 | RC9 | K403-6 | E401 (gra) | RC9 |
| J401-1 | K402-6 | RC95 | K403-7 | K404-7 | RC923 |
| J401-2 | K402-8 | RC90 | K403-7 | K407-3 | RC923 |
| K401-1 | XF405-2 (side) | RD93 | K403-9 | TB403-2 | RC923 |
| K401-2 | B401-1 | RD90 | K403-10 | TB404-3 | RC925 |
| K401-3 | XF404-2 (side) | RD92 | K403-12 | K403-14 | RC926 |
| K401-4 | B401-2 | RD92 | K403-12 | K405-11 | RC926 |
| K401-5 | XF403-2 (side) | RD95 | K403-13 | K402-9 | RC91 |
| K401-6 | B401-3 | RD95 | K403-13 | K406-12 | RC91 |
| K401-7 | K401-10 | RC93 | K403-14 | K403-12 | RC926 |
| K401-7 | K403-4 | RC93 | K403-14 | S113-2 | RC905 |
| K401-8 | E401 (grd) | RC9 | K403-14 | TB404-4 | RC926 |
| K401-9 | K402-9 | RC935 | K404-1 | K404-6 | RC4 |
| K401-9 | TB403-8 | RC935 | K404-2 | TB404-10 | RC7 |
| K401-10 | K401-7 | RC93 | K404-3 | K406-8 | RC6 |
| K401-10 | S112-2 | RC93 | K404-4 | K405-1 | RC6 |
| K401-10 | TB403-6 | RC93 | K404-6 | K404-1 | RC4 |
| K402-1 | CB401-6 | RD93 | K404-6 | TB403-5 | RC4 |
| K402-2 | R404-1 | RC902 | K404-7 | K403-7 | RC923 |
| K402-3 | CB401-5 | RD92 | K405-1 | K404-4 | RC6 |
| K402-4 | C430 | RC92 | K405-2 | E401 (grd) | RC9 |

TABLE 3-2. CABINET FROM-TO INFORMATION (Cont)

| FROM | TO | WIRE CODE | FROM | TO | WIRE CODE |
| :---: | :---: | :---: | :---: | :---: | :---: |
| K405-3 | CB402-4 | H9 | L402-1 | L401-2 | VC6 |
| K405-4 | K406-1 | H9 | L402-2 | C402-1 | VC906 |
| K405-5 | CB402-5 | H9 | L404-1 | Z404-5 | VE93 |
| K405-6 | K406-3 | H9 | L404-2 | C404-1 | VE92 |
| K405-7 | CB402-6 | H9 | M401-1 (+) | S405-11 | VB92 |
| K405-8 | K406-5 | H9 | M401-2 (-) | S405-2 | VB90 |
| K405-11 | K403-12 | RC926 | M402-1 (+) | R407-1 | RE92 |
| K405-12 | K406-11 | RC91 | M402-2 (-) | E407 | VE92 |
| K406-1 | K405-4 | H9 | M403-1 (+) | TB407-7 | VC906 |
| K406-1 | R401-1 | H9 | M403-2 (-) | TB407-15 | VC905 |
| K406-2 | CB403-3 | RE90 | M404-1 ( + | E401 (grd) | RC9 |
| K406-2 | R401-2 | H9 | M404-2 (-) | R419-3 | RC91 |
| K406-2 | T403-1 | H9 | R401-1 | K406-1 | H9 |
| K406-3 | K405-6 | H9 | R401-2 | K406-2 | H9 |
| K406-3 | R402-1 | H9 | R402-1 | K406-3 | H9 |
| K406-4 | CB403-2 | RE2 | R402-2 | K406-4 | H9 |
| K406-4 | R402-2 | H9 | R403-1 | K406-5 | н9 |
| K406-4 | T403-8 | H9 | R403-2 | K406-6 | H9 |
| K406-5 | K405-8 | H9 | R404-1 | K402-2 | RC902 |
| K406-5 | R403-1 | H9 | R404-2 | C429 | RC90 |
| K406-6 | CB403-1 | RE5 | R405-1 | C403-1 | VE92 |
| K406-6 | R403-2 | H9 | R405-2 | C431 | VE2 |
| K406-6 | T403-15 | H9 | R405-2 | E418 | VB92 |
| K406-7 | K403-1 | RC6 | R405-2 | TB407-14 | VC92 |
| K406-8 | K403-3 | RC6 | R407-1 | M402-1 (+) | RE92 |
| K406-8 | K404-3 | RC6 | R407-1 | TB415-G | VC92 |
| K406-9 | TB407-5 | RC902 | R407-2 | TB415-L | VC91 |
| K4-6-10 | E401 (grd) | RC9 | R408-1 | R413-1 | VB93 |
| K406-11 | K405-12 | RC91 | R408-2 | K407-1 | VB95 |
| K406-12 | K403-13 | RC91 | R409-1 | K407-2 | VB92 |
| K406-12 | XDS402-1 | RC91 | R409-1 | Z404-4 | RE92 |
| K407-1 | R408-2 | VB95 | R411-2 | C408-2 | VE92 |
| K407-2 | R409-1 | VC92 | R412-1 | C403-1 | VE92 |
| K407-3 | K403-7 | RC923 | R412-2 | C403-2 | RE92 |
| K407-3 | TB403-7 | RC92 | R413-1 | R408-1 | VB93 |
| K407-4 | S408-1 | RC923 | R413-1 | S405-6 | VB91 |
| L401-1 | TB406-5 | VC903 | R413-2 | E401 (grd) | VE9 |
| L 401-2 | C401-1 | VC905 | R414-2 | TB415-K | VC9 |
| L401-2 | L402-1 | VC6 | R415-3 | C427 | RC93 |

TABLE 3-2. CABINET FROM-TO INFORMATION (Cont)

| FROM | то | WIRE CODE | FROM | то | WIRE CODE |
| :---: | :---: | :---: | :---: | :---: | :---: |
| R417-1 | C402-1 | vC92 | S410-2 | S114-4 | RC923 |
| R417-3 | S405-5 | vb90 | S412-B | C403-1 | VE92 |
| R419-2 | S406-2 | DAS3 | S412-1 | S111-2 | RC 1 |
| R419-3 | M404-2 (-) | RC91 | S412-2 | TB403-3 | RC91 |
| R431-2 | C402-2 | vC91 | S413-B | S414-B | VE903 |
| R431-2 | C404-1 | VC95 | S414-B | S413-B | VE903 |
| S111-3 | TB404-7 | RC1 | T401-1 | CB401-4 | RD95 |
| S111-4 | S412-1 | RC1 | T 401-1 | K402-5 | RC923 |
| S111-4 | XDS 403-1 | RC1 | T 401-5 | CB401-6 | RD93 |
| S112-1 | E401 (grd) | RC9 | T 401-5 | K402-7 | RD90 |
| S112-2 | K401-10 | RC93 | T 401-8 | XF401-2 | RC6 |
| S113-1 | S114-1 | RC925 | T401-9 | E401 (grd) | RC9 |
| S113-1 | TB404-3 | RC925 | T 402-1 | XF402-2 | RC90 |
| S113-2 | K403-14 | RC905 | T 402-2 | CB401-4 | RC5 |
| S114-3 | S113-1 | RC925 | T 402-3 | TB406-1 | RB90 |
| S114-4 | S410-2 | RC923 | T 402-4 | C401-2 | RB91 |
| S405-2 | M401-2 (-) | vb90 | T402-5 | TB406-7 | RB92 |
| S405-5 | R417-3 | vb90 | T 403-1 | K406-2 | н9 |
| S405-6 | R413-1 | VB91 | T403-8 | K406-4 | н9 |
| S405-7 | S405-15 | RC9 | T 403-15 | K406-6 | н9 |
| S405-8 | TB405-7 | VB902 | T 404 - 1 | CB403-6 | RE90 |
| S405-11 | M401-1 (+) | vB92 | T 404-8 | CB403-5 | RE2 |
| S405-12 | E411 | VB93 | T404-15 | CB403-4 | RE5 |
| S405-13 | C432 | VB91 | T404-22 | Z404-3 | VD91 |
| S405-15 | E401 (grd) | VB9 | T 404-23 | z404-1 | VD95 |
| S405-15 | S405-7 | RC9 | T 404-24 | Z404-2 | VD96 |
| S405-17 | TB405-8 | VB96 | T 406-1 | K402-8 | RC90 |
| S406-2 | R419-2 | DAS3 | T 406-2 | K402-6 | RC95 |
| S406-4 | DC401 (yellow) | DAS9 | TB401-1 | CB402-1 | н9 |
| S406-5 | TB410-16 | RA95 | TB401-2 | CB402-2 | н9 |
| S406-6 | TB410-15 | RA90 | TB401-3 | CB402-3 | н9 |
| S406-11 | DC401 (red) | DAS2 | TB401-4 | E402 (grd) | RE9 |
| S407-1 | K402-10 | RC936 | TB401-4 | TB402-3 | RE9 |
| S407-2 | K403-4 | RC93 | TB402-1 | CB401-5 | RE2 |
| S408-1 | K407-4 | RC923 | TB402-2 | CB401-4 | RE5 |
| S408-2 | S409-1 | RC92 | TB402-3 | TB401-4 | RE9 |
| S409-1 | S408-2 | RC92 | TB403-1 | TB404-9 | RB90 |
| S409-2 | S410-1 | RC92 | TB403-2 | K403-9 | RC923 |
| S410-1 | S409-2 | RC92 | тB403-3 | S412-2 | RC91 |

TABLE 3-2. CABINET FROM-TO INFORMATION (Cont)

| FROM | TO | WIRE CODE | FROM | TO | WIRE CODE |
| :---: | :---: | :---: | :---: | :---: | :---: |
| TB403-5 | K404-6 | RC4 | TB407-4 | XDS402-2 | RC902 |
| TB403-6 | K401-10 | RC93 | TB407-5 | K406-9 | RC902 |
| TB403-6 | TB404-6 | RC93 | TB407-6 | TB405-2 | VC91 |
| TB403-7 | K407-3 | RC92 | TB407-7 | M403-1 (+) | VC906 |
| TB403-8 | K401-9 | RC935 | TB407-12 | K402-10 | RC902 |
| TB403-9 | TB404-2 | RC1 | TB407-13 | XDS401-2 | RC902 |
| TB403-10 | E402 (grd) | RE9 | TB407-14 | R405-2 | VC92 |
| TB403-11 | TB410-1 | RA91 | TB407-15 | M403-2 (-) | VC905 |
| TB404-2 | TB403-9 | RC1 | TB407-15 | TB405-1 | VC92 |
| TB404-3 | K403-10 | RC925 | TB407-20 | E422 (grd) | RC9 |
| TB404-3 | S113-1 | RC925 | TB408-1* | TB405-12 | VB96 |
| TB404-4 | K403-14 | RCR26 | TB408-2* | TB405-13 | VB93 |
| TB404-5 | K402-9 | RC935 | TB409-1* | TB405-10 | VB92 |
| TB404-6 | TB403-6 | RC93 | TB409-2* | TB405-9 | VB91 |
| TB404-7 | S111-3 | RC1 | TB410-1 | TB403-11 | RA91 |
| TB404-8 | TB405-11 | VB95 | TB410-1 | TB411-1 | RA91 |
| TB404-8 | XF401-1 | RD96 | TB410-2 | TB411-2 | RA91 |
| TB404-9 | TB403-1 | RB90 | TB410-3 | TB411-3 | RA93 |
| TB404-10 | K404-2 | RC7 | TB410-4 | TB411-4 | RA95 |
| TB405-1 | TB407-15 | VC92 | TB410-5 | TB411-5 | RA96 |
| TB405-2 | TB407-6 | VC91 | TB410-15 | S406-6 | RA90 |
| TB405-3 | TB415-B | VC93 | TB410-16 | S406-5 | RA95 |
| TB405-4 | TB415-H | VC95 | TB411-1 | TB410-1 | RA91 |
| TB405-7 | S405-8 | VB902 | TB411-2 | TB410-2 | RA91 |
| TB405-8 | S405-17 | VB96 | TB411-3 | TB410-3 | RA93 |
| TB405-9 | TB409-2* | VB91 | TB411-4 | TB410-4 | RA95 |
| TB405-10 | TB409-1* | VB92 | TB411-5 | TB410-5 | RA96 |
| TB405-11 | TB404-8 | VB95 | TB415-B | TB405-3 | VC93 |
| TB405-12 | TB408-1* | VB96 | TB415-G | R407-1 | VC92 |
| TB405-13 | TB408-2* | VB93 | TB415-H | TB405-4 | VC95 |
| TB406-1 | T402-3 | RB90 | TB415-K | R414-2 | VC9 |
| TB406-5 | L401-1 | VC903 | TB415-L | R407-2 | VC91 |
| TB406-7 | T402-5 | RB92 | XDS 401-1 | K402-9 | RC936 |
| TB406-12 | C401-2 | VC902 | XDS 401-2 | TB407-13 | RC902 |
| TB407-1 | E411 | VB93 | XDS 402-1 | K406-12 | RC91 |
| TB407-2 | C428 | VB93 | XDS402-2 | TB407-4 | RC902 |
| TB407-3 | E410 | VC92 | XDS403-1 | S111-4 | RC1 |
| *For remote control only. Otherwise, tie in cable. |  |  |  |  |  |

TABLE 3-2. CABINET FROM-TO INFORMATION (Cont)

| FROM | TO | WIRE CODE | FROM | TO | WIRE CODE |
| :--- | :--- | :--- | :--- | :--- | :--- |
| XF401-1 | K403-1 | RD96 | XF404-2 (side) | K401-3 | RD92 |
| XF401-1 | TB404-8 | RD96 | XF405-1 (top) | CB401-6 | RD93 |
| XF401-2 | T401-8 | RC6 | XF405-2 (side) | K401-1 | RD93 |
| XF402-1 | K402-8 | RC90 | Z404-1 | T404-23 | VD95 |
| XF402-2 | T402-1 | R404-2 | T404-24 | VD96 |  |
| XF403-1 (top) | CB401-4 | RD95 | RD95 | Z404-3 | T404-22 |
| XF403-2 (side) | K401-5 | RD92 | Z404-5 | R409-1 | VD91 |
| XF404-1 (top) | CB401-5 |  | L404-1 | RE92 |  |

parts list

This section contains a list of all replaceable electrical, electronic, and critical mechanical parts for F830-1 10-Kw Power Amplifier 522-2981-000. The manufacturers' codes appearing in the MFR CODE column of the parts list are listed in numerical order at the end
of the parts list. The code list provides manufacfacturers' names and addresses as shown in the Federal Supply Code for Manufacturers, Handbook H4-1. Manufacturers not listed in Handbook H4-1 are assigned a five-letter code and will appear first in the code list.
F830-1 Illustrations ..... 20
F830-1 Parts List ..... 35
Rectifier Illustrations ..... 46
Rectifier Parts List ..... 47/48


Figure 4-1. F830-1 10-Kw FM Power Amplifier (Sheet 1 of 15)


Figure 4-1. F830-1 10-Kw FM Power Amplifier (Sheet 2 of 15)



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Figure 4-1. F830-1 10-Kw FM Power Amplifier (Sheet 4 of 15)

$$
+\sqrt{40^{2}} 20^{50^{c / 25}} 1^{2.0^{0}}
$$



DiamB
TOP VIEW

Figure 4-1. F830-1 10-Kw FM Power Amplifier (Sheet 5 of 15)


Figure 4-1. F830-1 10-Kw FM Power Amplifier (Sheet 6 of 15)


Figure 4-1. F830-1 10-Kw FM Power Amplifier (Sheet 7 of 15)


Figure 4-1. F830-1 10-Kw FM Power Amplifier (Sheet 8 of 15)


Figure 4-1. F830-1 10-Kw FM Power Amplifier (Sheet 9 of 15)


Figure 4-1. F830-1 10-Kw FM Power Amplifier (Sheet 10 of 15)


Figure 4-1. F830-1 10-Kw FM Power Amplifier (Sheet 11 of 15)


Figure 4-1. F830-1 10-Kw FM Power Amplifier (Sheet 12 of 15)


Figure 4-1. F830-1 10-Kw FM Power Amplifier (Sheet 13 of 15)


Figure 4-1. F830-1 10-Kw FM Power Amplifier (Sheet 14 of 15)


DETAILET


Figure 4-1. F830-1 10-Kw FM Power Amplifier (Sheet 15 of 15)


| SYMBOL | DESCRIPTION | MANUFACTURER'S PART NUMBER | $\begin{gathered} \mathrm{MFR} \\ \mathrm{CODE} \end{gathered}$ | COLLINS <br> PART NUMBER |
| :---: | :---: | :---: | :---: | :---: |
| C447 | CADACITOR. FXD. CERAMIC 500 UUF, 50x TOL, 20.000 vDC |  |  | 913-1101-000 |
| C448 | SAME AS C447 |  |  |  |
| Cas9 | NOT USED |  |  |  |
| C450 | SAME AS C423 |  |  |  |
| C451 | CAPACITOR, FXD, CERAMIC 75 UUF, 5\% TOL. 3500 VDC | $8505752 \mathrm{PORM5PCT}$ | 71590 | 913-0830-000 |
| C.53 | CAPACITOR. FXD. CERAMIC <br> 100 UUF. $10 \times$ TOL. 5000 VDC | 850S100N | 71590 | 913-0822-000 |
| C453 | same as cas |  |  |  |
| C454 | SAME AS C452 |  |  |  |
| C455 | NOT USED |  |  |  |
| C456 | CAPACITOR, FXD. CERAMIC 10.000 UUF, 20\% TOL, 500 vDc | CK63AW 103 M | 81340 | 913-1188-000 |
| C457 | same as cas6 |  |  |  |
| Cass | Not USED |  |  |  |
| Cas9 | SAME AG Ca56 |  |  |  |
| C460 | NOT USED |  |  |  |
| C461 | NOT USED |  |  |  |
| C4.62 | CAPACITOR, VAR, AIR single section. 7 UUF to 100 UUF |  |  | 922-0025-000 |
| C8401 | CIRCUIT BREAKER <br> 10 AMP CURRENT RATING | 33635 | 74193 | 260-0407-000 |
| $\mathrm{CB4O} 2$ | CIRCUIT EREAKER 50.0 amp Current rating |  |  | 260-1569-000 |
| C8403 | same as ceat |  |  |  |
| COSO | NOT USED |  |  |  |
| $\begin{aligned} & \text { CRAO2 } \\ & \text { CR403 } \end{aligned}$ | SEmICONDUCTOR DEVICE, DIODE |  |  | 353-1546-000 |
| through | SAME AS CR402 |  |  |  |
| CR409 |  |  |  |  |
| CR4 10 | semiconductor device. RECTIFIER | 66-6794 | 81483 | 353-6259-000 |
| CRA11 |  |  |  |  |
| through | not used |  |  |  |
| CR421 CR422 | SEmICONOUCTOR DEVICE, DIODE | MR326R | 04713 | 353-1453-000 |
| CR423 |  |  |  |  |
| through CR426 | SAME AS CRaz2 |  |  |  |
| CR427 |  |  |  |  |
| through CRA33 | NOT USED |  |  |  |
| CR434 | SEmiconductor Device, diode |  |  | 353-1546-000 |
| CR435 | SAME AS CRA34 |  |  |  |
| CR436 | SAME AS CRa34 |  |  |  |
| CR437 | SAME AS CR434 | \} |  |  |
| CRA38 | SEMICONDUCTOR DEVICE, DIODE | 1N3044A | 07688 | 353-1339-000 |
| CR439 | SAME AS CRa 38 |  |  |  |
| CR440 | SEMICONDUCTOR DEVICE, DIODE | 1N3016日 | 07688 | 353-3121-000 |
| CR442 Through | NOT USEO |  |  |  |
| CR444 |  |  |  |  |
| Crats | SEmiconductior device, diode | 4M6RS2isajh3 AD | 03508 | 353-0289-000 |
| CR446 | SAme As Crab |  |  |  |
| CR447 | SAME AS CRA3E |  |  |  |
| DC401 | COUPLER, DIRECTIONAL double coupler with center CONDUCTOR. 12.000 WATTS | 442 E 3 | 16973 | 277-0183-000 |
| DS401 | LAMP, INCANDESCENT PILOT LIGHT 日ULE | 3s6-5 | 24446 | 262-3310-000 |
| DS402 | SAME AS DS401 |  |  |  |
| DS403 | LAMP, GLOW <br> $1 / 25$ WATT, 65 VAC | NE51 | 24455 | 262-0021-000 |


| SYMBOL | DESCRIPTION | MANUFACTURER'S PART NUMBER | $\begin{aligned} & \text { MFR } \\ & \text { CODE } \end{aligned}$ | COLLINS PART NUMBER |
| :---: | :---: | :---: | :---: | :---: |
| E401 | LENS, INDICATOR LIGHT GLASS. GREEN TRANSLUCENT | 75A101GRN | 72765 | 262-0258-000 |
| E402 | LENS. INDICATOR LIGHT GLASS. RED TRANSLUCENT | 75A101RED | 72765 | 262-0259-000 |
| E403A | DISCHARGER, ELECTROSTATIC CONSISTING OF ONE PLAIN CARBON BLOCK | P1385 | 77554 $\times$ Cexa | $\begin{aligned} & 975-0008-000 \\ & 975-0009-000 \end{aligned}$ |
| E403B E403C | SAME AS EAO3A ARPESTOR | B |  | HOCDER |
| E404 | TERMINAL LUG <br> PREINSULATED SOLDERLES <br> RING TONGUE | MS25036 | 00779 | 549-2453-002 $304-0253-000$ |
| E405 | SPRING, DUAL INTERLOCK. |  |  | 549-2315-003 |
| E406 | SAME AS 405 |  |  |  |
| F401 | FUSE, CARTRIDGE <br> 5.0 AMPS. 0.03 OHMS, 250 vDC | F02A250U5AS | 81349 | 264-4090-000 |
| F402 | FUSE CARTRIDGE <br> 0.250 AMPS, 250 VDC | F028250V1-4AS | 81349 | 264-4240-000 |
| F403 | FUSE CARTRIDGE <br> 2 AMPS | F028125V2AS | 81349 | 264-0008-000 |
| F404 | SAME AS F403 |  |  |  |
| F405 | SAME AS F403 |  |  |  |
| J401 | CONNECTOR, ELECTRICAL 3 CONTACTS | 7484 | 74545 | 368-0014-000 |
| $J 402$ | CONNECTOR, ELECTRICAL 50 OHMS, 500 VOLTS | UG1187U | 80058 | 357-9476-000 |
| J403 | CONNECTOR, ELECTRICAL COPPER CONTACTS | 000675 | 94375 | 357-9248-000 |
| $J 404$ | CONNECTOR, ELECTRICAL 1 CONTACT | UGS8AU | 80058 | 357-0003-000 |
| $J 405$ | CONNECTOR, ELECTRICAL <br> 1 CONTACT | UG 10940 | 80058 | 357-9183-000 |
| K401 | RELAY. ARMATURE 4C CONTACT ARRANGEMENT | 83-4544 | 04221 | 970-1933-000 |
| $K 402$ | RELAY, POWER <br> 10 AMPS, 600 VAC, 50 CPS | 700B400A110V60 | 01121 | 405-0428-000 |
| K403 | SAME AS K401 |  |  |  |
| K404 | RELAY, ARMATURE 2C CONTACT ARRANGEMENT | 83-3598- | 04221 | 970-1934-000 |
| K405 | RELAY, POWER <br> 50 AMPS, $110 \mathrm{VAC}, 60 \mathrm{CPS}$ | 8502 | 81487 | 405-0298-000 |
| $K 406$ | SAME AS KAOS |  |  |  |
| K407 | RELAY, ARMATURE <br> IC CONTACT ARRANGEMENT | 95062 | 78277 | 408-1114-000 |
| L401 | ```REACTOR 100 MA DC, 280 OHMS. 1000 VOLTS``` | 18892 | 97965 | 678-0584-000 |
| 4.402 | SAME AS L401 |  |  |  |
| L403 | $\begin{aligned} & \text { REACTOR } \\ & \quad 3.0 \text { AMPS. } 7 \text { OHMS } \end{aligned}$ | E11868 | 80008 | 668-0089-000 |
| L404 | $\begin{aligned} & \text { REACTOR } \\ & 2.6 \text { AMPS. } 10 \text { OHMS } \end{aligned}$ | E12631 | 80008 | 668-0032-000 |
| L405 | COIL. RADIO FREQUENCY 0.50 OHMS. 3.9 UH .1145 MA CURRENT | LT7K193 | 81349 | 240-0177-000 |
| 4406 | NOT USED |  |  |  |
| L407 | CHOKE | LT7K189 | 81349 | 240-0173-000 |
| $L 408$ $L 409$ | NOT USED COIL. RADIO FREQUENCY SOLDERED, 1/2 INCH BY 11/日 INCHES LONG BY 5/8 INCHES |  |  | 549-2297-003 |
| L410 | NOT USED |  |  |  |
| L411 | NOT USED |  |  |  |
| L412 | SAME AS L403 |  |  |  |
| M401 | METER, ELECTRICAL <br> 1 MA DC METER RANGE, 100 OHMS |  |  | 548-0649-000 |
| M402 | AMMETER | 56-0824-0000 | 80145 | 458-0658-000 |



| SYMBOL | DESCRIPTION | MANUFACTURER＇S PART NUMBER | MFR CODE | COLLINS <br> PART NUMBER |
| :---: | :---: | :---: | :---: | :---: |
| MP432 | INCHES LONG <br> MOUNT．RESILIENT <br> CADMIUM ISCLATOR． 0.169 <br> I NCHES WIDE EY 0.169 <br> LONG | 100PH377－2 | $76005$ | 200－1957－000 |
| MP433 | MOUNT，RESILENT <br> CADMIUM ISOLATOR．0． 169 <br> INCHES WIDE EY 0.169 <br> INCHES LONG | 100PH377－8 | 76005 | 200－1961－000 |
| MP434 | NOT USED |  |  |  |
| MP435 | SAME AS MP40日 |  |  |  |
| MP436 | LATCH．MAGNETIC <br> ALUMINUM，SUPPLIED WITH OR WITHOUT STRIKER PLATE | ADPL $100 D C S T 1$ | 84792 | 015－0899－000 |
| P401 | CONNECTOR，ELECTRICAL <br> 3．WIRE MIDGET，TWIST LOCK． 10 AMPS． 250 VOLTS |  |  | 368－00：3－000 |
| P402 | CONNECTOR．ELECTRICAL BRASS．3／4 INCH DIAMETER BY $1-1 / 2$ INCHES LONG | UG $1185 \Delta \mathrm{~S}$ | 81349 | 357－9326－000 |
| P403 | CONNECTOR，ELECTRICAL <br> 1 CONTACT | MS 35168 －88E | 96906 | 357－9292－000 |
| P404 | SAME AS PAOZ |  |  |  |
| P405 | SAME AS P402 |  |  |  |
| R401 | RESISTOR，FXD，WIRE WOUND 3.4 OHMS， $10 \%$ TOL， 8.3 AMPS | 41162－2 | 44655 | 714－1612－000 |
| R4O2 | SAMF AS R4O1 |  |  |  |
| R403 | SAME AS RAOI |  |  |  |
| R404 | RESISTOR，FXD．WIRE WOUND 7.5 OHMS， $10 \%$ TOL． 100 watts | $\begin{aligned} & \text { R100W7-5-1 OPCT } 7- \\ & 8 \end{aligned}$ | 94310 | 738－0025－000 |
| R405 | RHEOSTAT <br> 50 OHMS， $10 \%$ TOL． 500 WATTS | 50446 | 44655 | 735－5300－000 |
| R406 | RESISTOR．FXD．COMPOSITION 270 OHMS， $10 \%$ TOL， $1 / 2$ WATT | RC20GF271K | 81349 | 745－1328－000 |
| R407 | RESISTOR．FXD，WIRE WOUND 2.5 OHMS， $3 x$ TOL， 50 WATTS | RH50－2R500G | 91637 | 747－8697－000 |
| R408 | RESISTOR，VAR．WIRE WOUND <br> 50 OHMS．10\％TOL， 2 wATTS |  |  | 377－0619－000 |
| R409 | RESISTOR，FXD．WIRE WOUND 0.75 OHMS． $5 \%$ TOL． 50 watTS | M50w－75－5 | 00213 | 747－9566－000 |
| R410 | RESISTOR，FXD，WIRE WOUND 100K OHMS，5\％TOL， 210 WATTS | RW47V104 | 81349 | 746－6737－000 |
| R411 | SAME AS R410 |  |  |  |
| 8412 | RESISTOR，FXD，WIRE WOUND 4K OHMS，5\％TOL， 210 WATTS | HL225－40000」 | 91637 | 746－6709－000 |
| R413 | RESISTOR，FXD，WIRE WOUND 0.5 OHMS，5\％TOL， 50 WATTS | MSOWO－5－5 | 00213 | 747－9564－000 |
| R414 | SAME AS R413 |  |  |  |
| R415 | RESISTOR．FXD．FILM <br> 1／4 WATT， $1 \%$ TOL |  |  | 705－0519－000 |
| R416 | ```RESISTOR, FXD, FILM 402K OHMS. 1% TOL, 1 WATT``` | RN75日4023F | 81349 | 705－3287－000 |
| R417 | RESISTOR．ADJUSTAELE， <br> WIRE WOUND <br> 1500 OHMS． $10 \%$ TOL． 50 <br> WATTS |  |  | 716－0026－000 |
| R41日 | RESISTOR．FXD，COMPOSITION 3900 OHMS， $10 \%$ TOL， 1 WATT | RC32GF 392K | 81349 | 745－3377－000 |
| R419 | RESISTOR，VAR，COMPOSITION IOK OHMS．20x TOL． 2 WATTS | RVALAYSA1038 | 81349 | 380－2757－000 |
| R420 | $\begin{aligned} & \text { RESISTOR, FXD, FILM } \\ & \text { IK OHMS, } 1 \% \text { TOL, } 2 \text { WATTS } \end{aligned}$ | RN80日 $1004 F$ | 81349 | $705-4254-000$ |
| R421 <br> THROUGH | SAME AS R420 |  |  |  |


| SYMBOL | DESCRIPTION | MANUFACTURER'S PART NUMBER | $\begin{aligned} & \text { MFR } \\ & \text { CODE } \end{aligned}$ | COLLINS PART NUMBER |
| :---: | :---: | :---: | :---: | :---: |
| R425 |  |  |  |  |
| R426 | RESISTOR, FXD. FILM 750K OHMS, $1 \%$ TOL, 2 WATTS | RN8087503K | 81349 | 705-4251-000 |
| R427 | GAME AS R426 |  |  |  |
| R428 | RESISTOR, FXD. WIRE WOUND 2.5 OHM.S. 3\% TOL. 5 WATTS | 155052-5-3PCT | 00213 | 747-9748-000 |
| R429 | NOT USED |  |  |  |
| R430 | RESISTOR, FXD, COMPOSITION 390 OHMS. $10 \%$ TOL. 2 WATTS | RC42GF391K | 81349 | 745-5635-000 |
| R431 | RESISTOR, FXD, WIRE WOUND 1600 OHMS. 5\% TOL, 55 WATTS | RW35V162 | 81349 | 747-2751-000 |
| R432 | RESISTOR, FXD, COMPOSITION 27K OHMS. $10 \%$ TOL. 1 WATT | RC32GF 273 K | 81349 | 745-3412-000 |
| R433 | RESISTOR. FXD. COMPOSITION $1 K$ OHMS, $10 \%$ TOL. 2 WATTS | RC42GF102K | 81349 | 745-5652-000 |
| R4.34 | SAME AG R433 |  |  |  |
| R435 | RESISTOR, FXD, COMPOSITION 100K OHMS, $10 \%$ TOL, 1 WATT | RC32GF104K | 81349 | 745-3436-000 |
| R4.36 | SAME AS R435 |  |  |  |
| R437 | SAME AS R435 |  |  |  |
| R438 | SAME AS R435 |  |  |  |
| R439 |  |  |  |  |
| THROUGH | SAME AS R435 |  |  |  |
| R4423 |  |  |  |  |
| THROUGH | NOT USED |  |  |  |
| R4R6 |  |  |  |  |
| R467 | RESISTOR, FXD, WIRE WOUND 220 OHMS. 5\% TOL. 6.5 !VATTS | RC428F184K | 81349 | 747-5447-000 |
| R468 | RESISTOR, FXD, COMPOS!TION 3300 OHMS, $1 \%$ TOL, 6.5 WATTS | 886AS3300 | 10646 | 712-4224-000 |
| R469 | RESISTOR. FXD. COMPOSITION 470 OHMS. $10 \%$ TOL. 2 WATTS | RC42GF471K | 81349 | 745-5638-000 |
| 8470 |  | $7725 F 2$ | 10646 | 712-0002-000 |
| R471 | ```RESISTOR, FXD, CERMET 50 OHMS, 20% TOL, 60 WATTS``` | 218SP9 | 10646 | 712-0070-000 |
| 0472 | SAMF $\triangle$ S R470 |  |  |  |
| 8473 | SANE AS R467 |  |  |  |
| R474 |  |  |  |  |
| R475 | RESISTOR, FXD, WIRE WOUND 100 OHMS, $5 \%$ TOL, 5 WATTS | RW67V101 | 81349 | 747-5440-000 |
| R476 | SAME AS R475 |  |  |  |
| R477 | RESISTOR, FXD, COMPOSITION 39 OHMS, $10 \%$ TOL, 2 WATTS | RC42GF390K | 81349 | 745-5593-000 |
| R478 | SAME AS R477 |  |  |  |
| R479 | NOT USED |  |  |  |
| R480 | RFSISTOR, FXD. WIRE WOUND 2n OHMS. 5\% TOL. 55 WATTS | RW35V200 | 81349 | 747-2713-000 |
| R48 1 | RESISTOR, FXD, FILM <br> 2870 OHMS. $1 \%$ TOL. 1 WATT | RN7582871F | 81349 | 705-3272-000 |
| R482 | RESISTOR, FXD, COMPOSITION 1 OK OHMS. $10 \%$ TOL, 2 WATTS | RC42GF103K | 81349 | 745-5694-000 |
| R483 | SAME AS R481 |  |  |  |
| R489 | SAME AS R471 |  |  |  |
| R485 | ```RESISTOR. FXD, FILM 1200 OHMS. 5% TOL, 1 WATT``` | RL325122J | 81349 | 745-3946-000 |
| R486 | SAME AS R4BE |  |  |  |
| R487 | ```RESISTOR, FXD, FILM 3600 OHMS, 5% TOL, 1 WATT``` | RL325362J | 81349 | 745-3974-000 |
| R488 | NOT USED |  |  |  |
| R4R9 | NOT USFD |  |  |  |
| R490 | SAME AS RAO7 |  |  |  |
| R491 | SAME AS R4O7 |  |  |  |
| R492 | RESISTOR. FXD, WIRE WOUNO 2.5 OHMS, $3 \%$ TOL. 5 WATTS | 1550552-5-3PCT | 00213 | 747-9748-000 |
| R493 | SAME AS R4E7 |  |  |  |



F830－1 10－Kw FM Power Amplifier

| SYMBOL | DESCRIPTION | MANUFACTURER＇S PART NUMBER | $\begin{aligned} & \text { MFR } \\ & \text { CODE } \end{aligned}$ | COLLINS <br> PART NUMBER |
| :---: | :---: | :---: | :---: | :---: |
| TB408 | TERMINAL BOARD <br> a TERMINALS，FOR USE WITH OPTIONAL REMOTE CONTROL EQUIPMENT | 600－4 | 75382 | 367－0002－000 |
| TR409 | SAME AS TRAOB |  |  |  |
| т日ム：0 | SAME AS TE405 |  |  |  |
| TB411 | TERMINAL STRIP 10 TERMINALS | 10－141 | 71785 | 367－4100－000 |
| TB4 12 | TERMINAL EOARD <br> PLASTIC， $1 / 16$ INCHES THICK BY 2－1／8 INCHES BY 4－9／16 INCHES LONG |  |  | 533－5766－003 |
| TB4 13 | SAME AS TB412 |  |  |  |
| TB414 | NOT USED |  |  |  |
| TB415 | ```TERMINAL EOARD TWO RINGS OF 6 TERMINALS EACH``` | 6 H 12 | 00534 | 306－0909－000 |
| $V 401$ | ELECTRON tUbe GLASS ENVELOPE，RECTIFIER | 8724872 | 72092 | 256－0037－000 |
| V402 |  |  |  |  |
| $\begin{aligned} & \text { THROUGH } \\ & \text { VAOG } \end{aligned}$ | SAme as vali |  |  |  |
| V407 | ELECTRON TU日E TETRODE | $1 \times 2 A B$ | 49671 | 256－0122－000 |
| XDS401A | LAMPHOLDER <br> PANEL MOUNTING，USE WITH CANDELABRA SCREW BASE LAMP | 75LESSLENS | 72765 | 262－0255－000 |
| XDS402A | SAME AS XDS4O1A |  |  |  |
| XDS403A | LAMPHOLDER MOUNTING BRACKET | 50 | 72765 | 262－1260－000 |
| XF401 | FUSEHOLDER | HKLJRWZZ | 71400 | 265－1040－000 |
| XF402 | SAME AS XF401 |  |  |  |
| XF403 | SAME AS XF401 |  |  |  |
| XF404 | SAME AS XF401 |  |  |  |
| XF405 | SAME AS XF401 |  |  |  |
| XV401 | SOCKET．ELECTRON TUBE 4 PIN EAYONET BASE TUBE SOCKET． 20 AMP | 123－211－30 | 74970 | 220－1460－000 |
| XVAO2 |  |  |  |  |
| THROUGH xv406 | SAME AS XVAOI |  |  |  |
| XV407 | SOCKET．ELECTRON TUBE AIR SYSTEM SOCKET | Y291 | 06980 | 220－1491－000 |
| 2401 | NOT USED |  |  |  |
| Z402 | NOT USED |  |  |  |
| 2403 | NOT USED |  |  |  |
| Z404 | ```SEMICONDUCTOR DEVICE. RECTIFIER THREE PHASE BRIDGE, 400 CPS``` | 67－7304 | 81483 | 353－6273－000 |
| 2405 | RECIFIER ASSEMBLY BIAS 0.062 INCHES BY 3.500 INCHES EY 6.625 INCHES |  |  | 549－2259－004 |
| Z406 | SEMICONDUCTOR DEVICE， RECTIFIER silicon，three phase full WAVE BRIDGE VOLTAGE DOU日LER．OPTIONAL ASSEMBLY． SEE BREAKDOWN ON PAGE 47／48 | 67－7303 | 81483 | 756－8563－000 |
| MANUFACTURERS CODES |  |  |  |  |
| CODE | MANUFACTURER |  |  |  |
| EIMAC | EIMAC |  |  |  |
| 00213 | SAGE ELECTRONICS CORP |  |  |  |
|  | P．O．BOX 3926 ROCHESTER，N．Y． |  |  |  |



| SYMBOL | DESCRIPTION | MANUFACTURER＇S PART NUMBER | $\begin{aligned} & \text { MFR } \\ & \text { CODE } \end{aligned}$ | COLLINS <br> PART NUMBER |
| :---: | :---: | :---: | :---: | :---: |
| 72982 | ERIE TECHNOLOGICAL PRODUCTS INC． <br> 604 WEST $12 T H$ StREET <br> ERIE，PA． |  |  |  |
| 74193 | heinemann electric co． 2619 日RUNSWICK PIKE TRENTON．N．J． |  |  |  |
| 74545 | hubeell harvey inc． BRIDGEPORT，CONN． |  |  |  |
| 74970 | E．F．Johnson CO． 297 TENTH AVE．S．W． WASECA，MINN． |  |  |  |
| 75382 | KULKA ELECTRIC CORP． <br> MT．VERNON．N．Y． |  |  |  |
| 76005 | ```LORD MFG. CO. 1635 WEST 12TH sTREET FRIF. PA``` |  |  |  |
| 76854 | OAK MFG．CO． gouth main CRYSTAL LAKE，ILL． |  |  |  |
| 78277 | sigma instruments inc． 170 PEARL ST． SOUTH ERAINTREE，MASS． |  |  |  |
| 80008 | electro engineering works OAKLAND，CALIF． |  |  |  |
| 80058 | JOINT ELECTRONIC TYPE DESIGNATION SYSTEM |  |  |  |
| 80145 | ASSEMBLY PRODUCTS INC． 7100 WILSON MILLS ROAD ChESTERLAND，OHIO |  |  |  |
| 80147 | biggs steel foundry ano fabricating co． <br> AKRON．OHIO |  |  |  |
| 81349 | MILITARY SPECIFICATIONS promulgated by standardization division directorate of logistic SERVICES DSA |  |  |  |
| 81483 | international rectifier CODR． <br> 1523 EACT GRAND AVE． <br> EL SEGUNDO，CALIF． |  |  |  |
| 81487 | SQuARE D CO． <br> industrial controller <br> DIVISION <br> 4041 NORTH RICHARDS ST． <br> MILWAUKEE，wIS． |  |  |  |
| 82877 | ROTRON MFG．CO．INC． 7－9 hasgrouck LaNE WOODSTOCK，N．Y． |  |  |  |
| 84792 | heppner mfg．CO． P．O．Box 0 ROUNO LAKE，ILL． |  |  |  |
| 85107 | neptune electronics co． 30 WEST 15TH．STREET NEW YORK．N．Y． |  |  |  |
| 90211 | SQUARE D CO． 9405 RIVER CHICAGO，ILL |  |  |  |
| 91637 | Dale Electronics inc． columbus，ne日r |  |  |  |
| 92702 | IMC MAGNETICS CORP． EASTERN DIVISION 570 MAIN STREET WESTBURY LONG ISLAND，N．Y． |  |  |  |
| 94310 | TRU－OHM PRODUCTS <br> MEMCOR COMPONETS DIVISION <br> P．O．日OX 890 <br> huntington．Ind． |  |  |  |



Figure 4-2. Semiconductor Device, Rectifier



## https://bh.hallikainen.org


[^0]:    NOTE: "TYPICAL OPERATION" data ara oblained by calculation from publishad characteristic eurves and confirmed by direct tests. Adjustment of the r-f grid drivo to obtain the specified plate current at the specified grid bias, screen voltage, and plate voltage is assumed. If this procedure is followed, there will be little variation in output power when tubes are changed. even thaugh there may be soma variations in grid and screan currents. The grid and screen currents which result when the desired plata current is obtained are incidental and vary from tube to tube. These eurrent variations eauso no difficulty so lang as the circuif maintains the correct voltage in the presence of the variations in current. If grid bias is obtained principally by means of a grid rosistor, the rosistor must be adjustable to obtain the required bias voltago when the correct r-f driving voltage is appliad.

